

EECS 16B

Designing Information Devices and Systems II

Lecture 4

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Transient Response

- Outline
 - R-L-C Circuits
 - Phasors

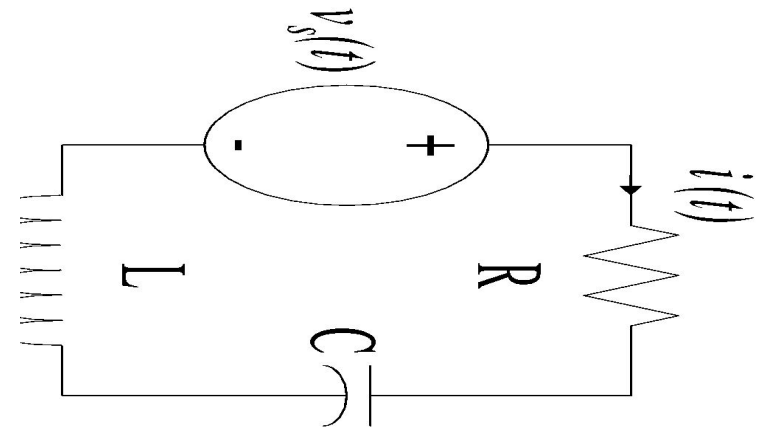
- Reading- Hambley text sections 4.5, 5.1,5.2slides

R-L-C circuits: Response in time

$$v_s = iR + v_c + v_L$$

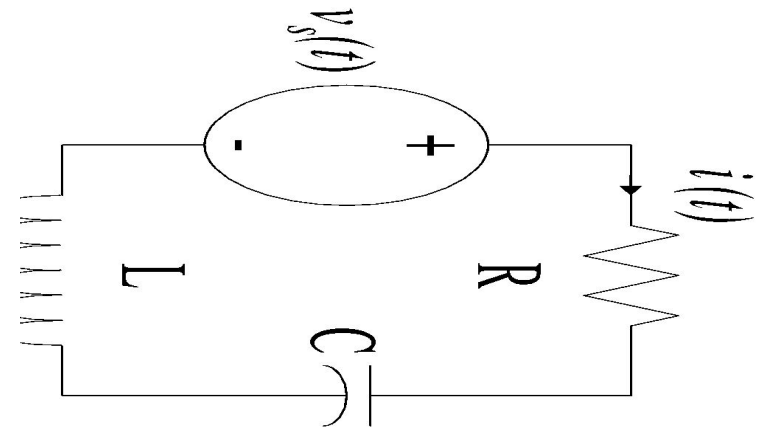
$$v_s = RC \frac{dv_c}{dt} + v_c + L \frac{di}{dt}$$

$$i = C \frac{dv_c}{dt}$$



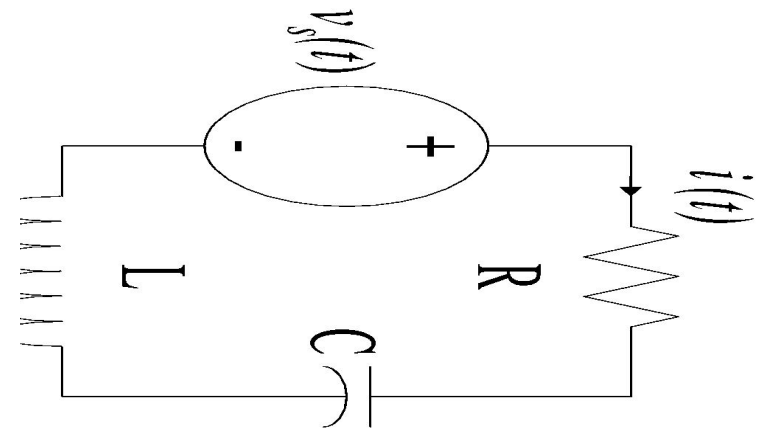
Filters, antennas, resonances

R-L-C circuits: Response in time



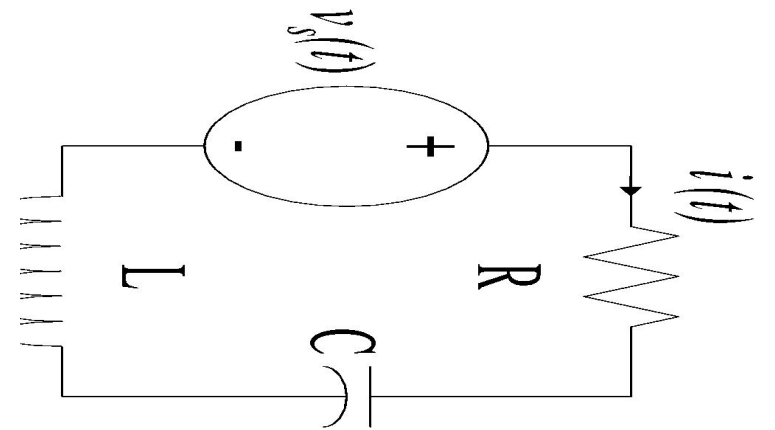
Filters, antennas, resonances

R-L-C circuits: Response in time



Filters, antennas, resonances

R-L-C circuits: Response in time

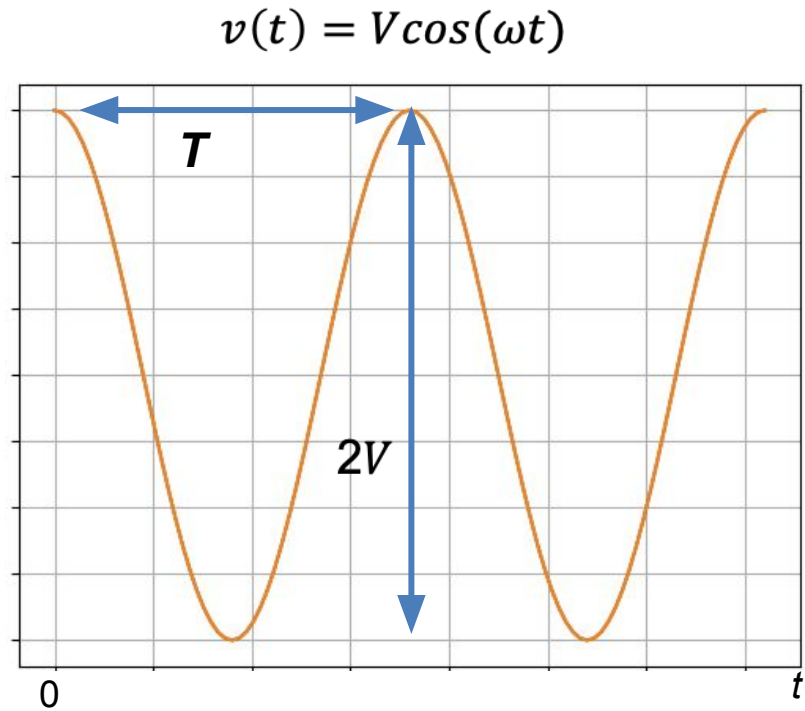


Filters, antennas, resonances

Summary

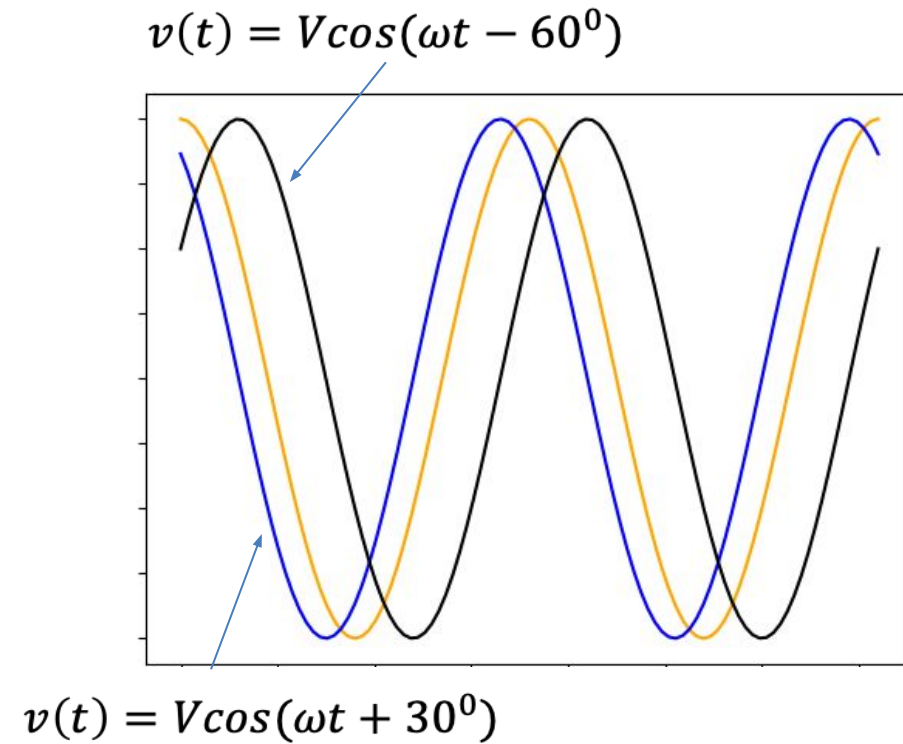
- Overdamped – real unequal roots
- Critically damped – Real Equal roots
- Underdamped – Complex roots

Sinusoidal voltages

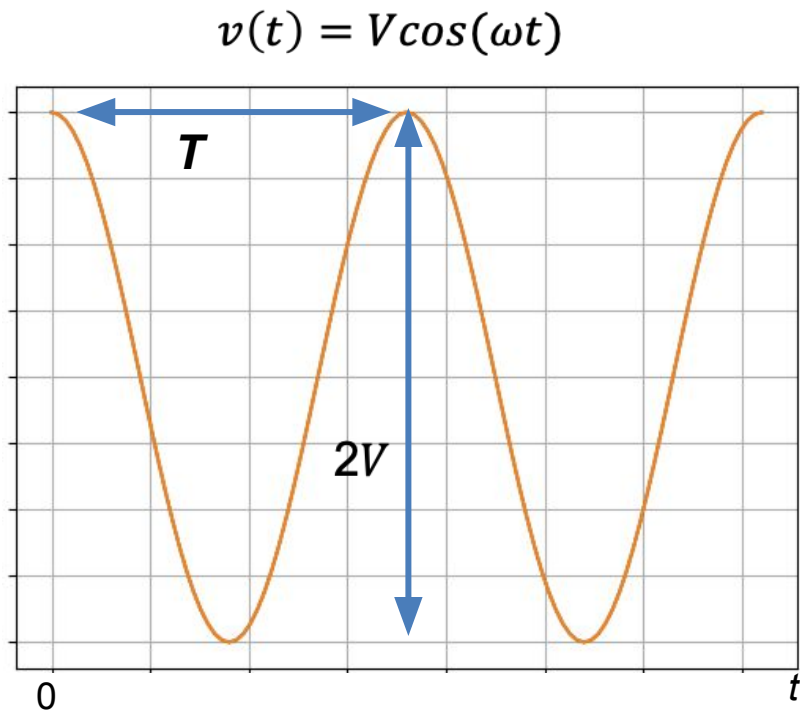


T : Period

$$\omega = \frac{2\pi}{T}$$



Root Mean Square Values



T : Period

$$\omega = \frac{2\pi}{T}$$

Average Power over one period:

$$P = \frac{1}{T} \int_0^T \frac{v^2}{R} dt$$

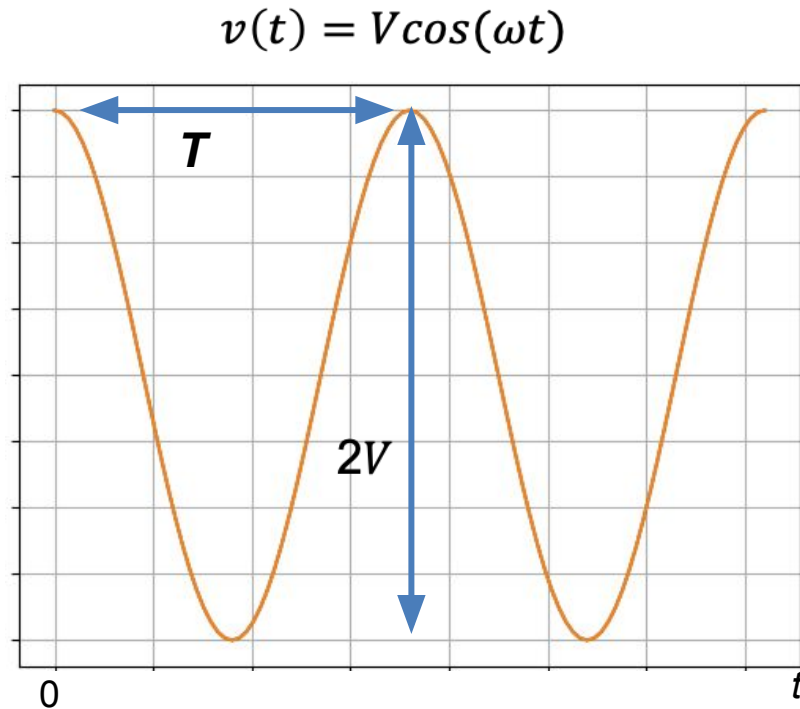
$$P = \frac{\left[\sqrt{\frac{1}{T} \int_0^T v^2 dt} \right]^2}{R}$$

Comparing with conventional equation: $P = \text{voltage}^2/R$

A new quantity is defined for time-varying voltages known as the root-mean-square voltage

$$v_{rms} = \sqrt{\frac{1}{T} \int_0^T v^2 dt}$$

Root Mean Square Value for Sinusoidal Voltage



T : Period

$$\omega = \frac{2\pi}{T}$$

Average Power over one period:

$$v_{rms}^2 = \frac{1}{T} \int_0^T v^2 dt = \frac{1}{T} \int_0^T V^2 \cos^2(\omega t) dt = \frac{1}{2T} V^2 \int_0^T (1 + \cos 2\omega t) dt$$

$$v_{rms}^2 = \frac{V^2}{2T} (t + 2\sin 2\omega t) \Big|_0^T = \frac{V^2}{2T} [T - 0 + \sin 2\omega T - 0] = \frac{V^2}{2}$$

$$v_{rms} = \frac{V}{\sqrt{2}}$$

How do we add arbitrary sinusoids?

$$v(t) = 10\cos\omega t + 5\sin\omega t - 5\cos(\omega t - 30^\circ)$$

$$v(t) = 10\cos\omega t + 5\cos(\omega t - 90^\circ) - 5\cos(\omega t - 30^\circ)$$

Remember? $\cos(a + b) = \cos a \cos b - \sin a \sin b$

Lets do it it differently $e^{j\theta} = \cos\theta + j\sin\theta$

$$e^{-j\theta} = \cos\theta - j\sin\theta$$

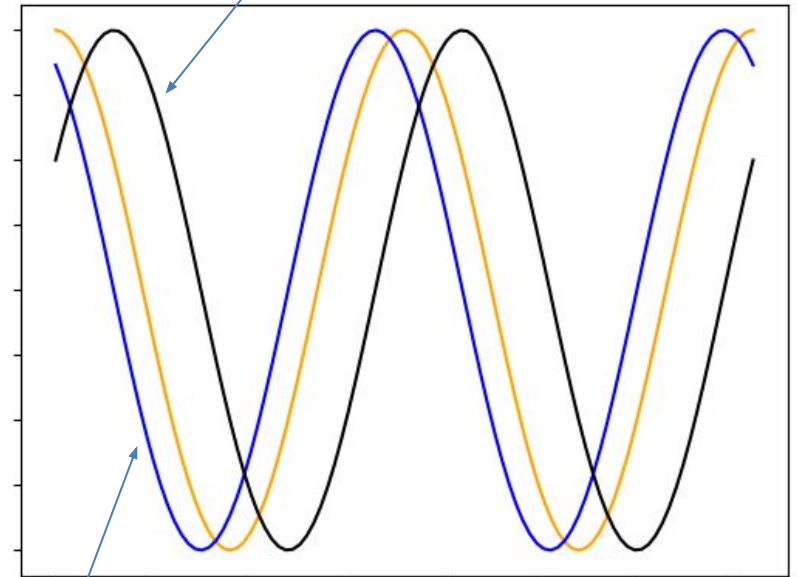
Then

$$\cos\theta = \frac{1}{2}(e^{j\theta} + e^{-j\theta})$$

$$\sin\theta = \frac{1}{2j}(e^{j\theta} - e^{-j\theta})$$

$$v(t) = 10\cos\omega t + 5\cos(\omega t - 90^\circ) - 5\cos(\omega t - 30^\circ) = 1/2e^{j\theta}[10 + 5e^{-j\theta}]$$

$$v(t) = V\cos(\omega t - 60^\circ)$$



$$v(t) = V\cos(\omega t + 30^\circ)$$

How do we add arbitrary sinusoids?

$$\begin{aligned}v(t) &= 10\cos\omega t + 5\cos(\omega t - 90^\circ) - 5\cos(\omega t - 30^\circ) \\&= \frac{1}{2}e^{j\omega t}[10] + \frac{1}{2}e^{j(\omega t - 90^\circ)}[5] - \frac{1}{2}e^{j(\omega t - 30^\circ)}[5] \\&\quad + \frac{1}{2}e^{-j\omega t}[10] + \frac{1}{2}e^{-j(\omega t - 90^\circ)}[5] - \frac{1}{2}e^{-j(\omega t - 30^\circ)} \\&= \frac{1}{2}e^{j\omega t}[10 + 5e^{-j90} - 5e^{-j30}] + \frac{1}{2}e^{-j\omega t}[10 + 5e^{+j90} - 5e^{+j30}] \\&= \frac{1}{2}e^{j\omega t}[10 + 5\cos 90 - j5\sin 90 - 5\cos 30 + j5\sin 30] + cc \\&= \frac{1}{2}e^{j\omega t}\left[10 + 0 - j5 - 5\frac{\sqrt{3}}{2} + \frac{j5}{2}\right] + cc \\&= \frac{1}{2}e^{j\omega t}[5.66 - j2.5] + cc \\&= \frac{1}{2}6.18e^{j(\omega t - 23^\circ)} + cc \\&= 6.18\cos(\omega t - 23^\circ)\end{aligned}$$

$$\begin{aligned}\cos\theta &= \frac{1}{2}(e^{j\theta} + e^{-j\theta}) \\ \sin\theta &= \frac{1}{2}(e^{j\theta} - e^{-j\theta})\end{aligned}$$

$$Ae^{-j\theta} = 5.66 - j2.5$$

$$Ae^{j\theta} = 5.66 + j2.5$$

$$A^2 = 5.66^2 - j^2 2.5^2$$

$$A^2 = 5.66^2 + 2.5^2$$

$$A = \sqrt{5.66^2 + 2.5^2} = 6.18$$

$$\cos\theta = 5.66/6.18; \sin\theta = 2.5/6.18$$

$$\tan\theta = \frac{2.5}{5.66} = 0.44$$

$$\theta = 0.41 = 23^\circ$$

Some Observations

$$\begin{aligned}v(t) &= \frac{1}{2}e^{j\omega t}[5.66 - j2.5] + cc \\&= \frac{1}{2}6.18e^{j(\omega t - 23^\circ)} + \frac{1}{2}6.18e^{-j(\omega t - 23^\circ)} \\&= \frac{1}{2}6.18[\cos(\omega t - 23^\circ) + j\sin(\omega t - 23^\circ)\cos(\omega t - 23^\circ) - j\sin(\omega t - 23^\circ)] \\&= \text{Real} [6.18e^{j(\omega t - 23^\circ)}]\end{aligned}$$

Phasors



In short hand, it is represented as $6.18 \angle -23^\circ$

Some Observations

$$A(t) = 5\cos(\omega t) = \text{Real}[5e^{j(\omega t)}]$$

$$B(t) = 5\cos(\omega t - 90^\circ) = \text{Real}[5e^{j(\omega t - 90^\circ)}]$$

$$C(t) = 5\cos(\omega t + 90^\circ) = \text{Real}[5e^{j(\omega t + 90^\circ)}]$$

At any given time t , $B(t)$ is trailing or lagging behind $A(t)$ by 90° while $C(t)$ is leading $A(t)$ by the same amount

Let us now look at the phasors at $t=0$

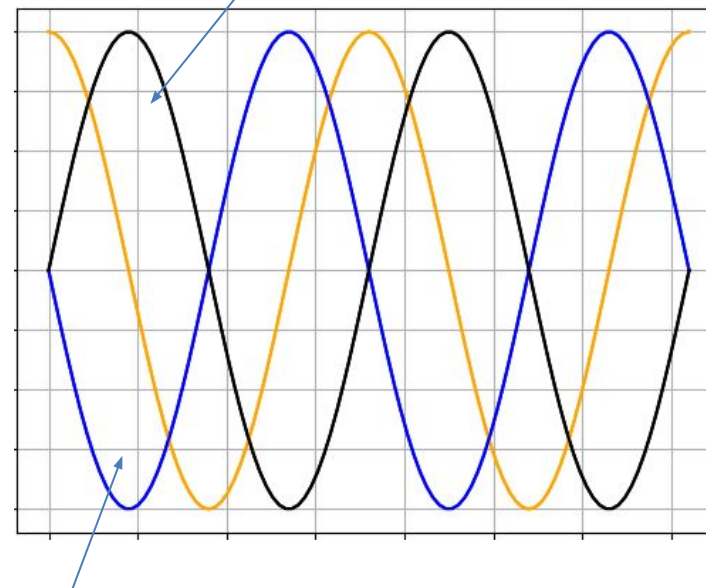
$$A(0) = 5$$

$$B(0) = 5(\cos 90^\circ - j\sin 90^\circ) = -j5$$

$$C(0) = 5(\cos 90^\circ + j\sin 90^\circ) = +j5$$

Therefore $+j$ or $-j$ signifies signals having 90° phase lead or lag respectively.

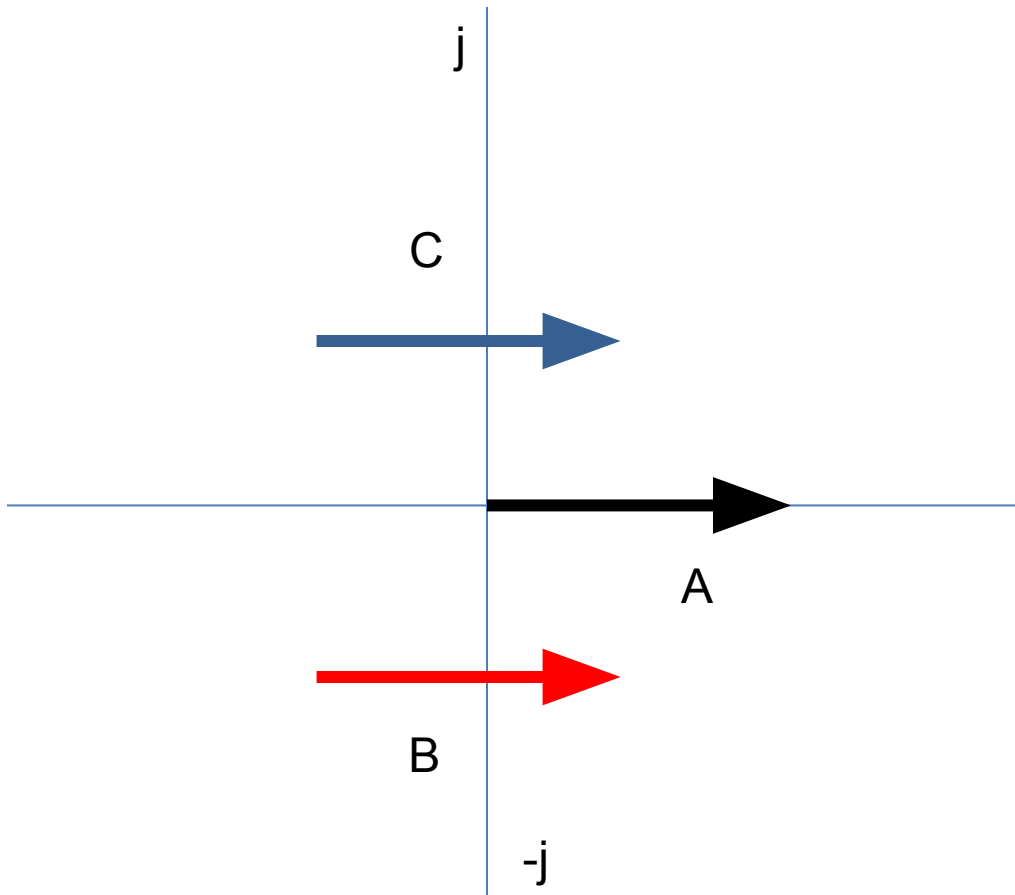
$$v(t) = V\cos(\omega t - 90^\circ)$$



$$v(t) = V\cos(\omega t + 90^\circ)$$

Some Observations

$+j$ or $-j$ signifies signals having 90° phase lead or lag respectively.



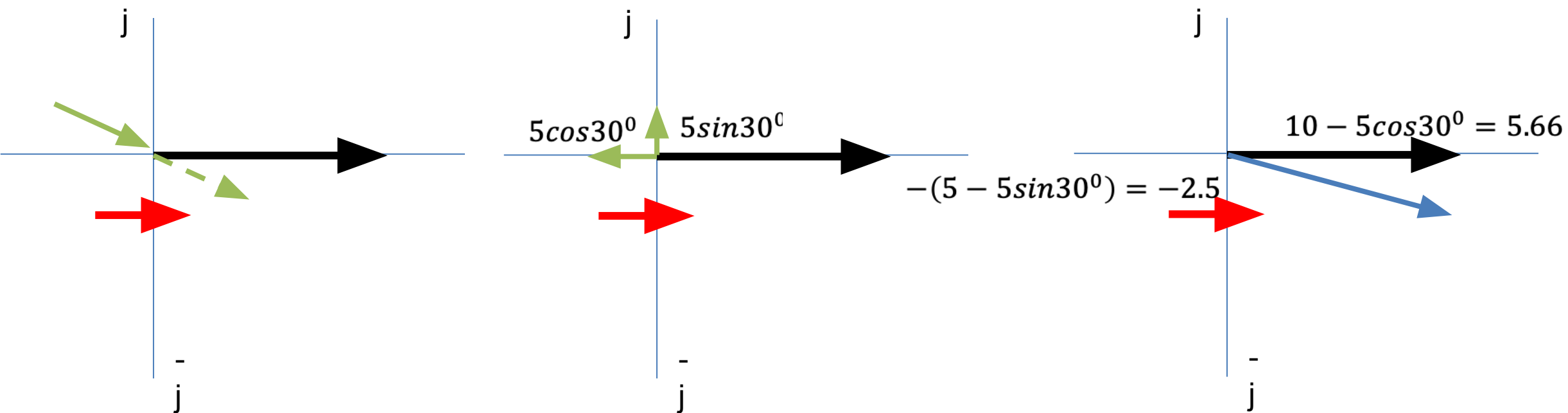
$$A(t) = 5\cos(\omega t) = \text{Real}[5e^{j(\omega t)}]$$

$$B(t) = 5\cos(\omega t - 90^\circ) = \text{Real}[5e^{j(\omega t - 90^\circ)}]$$

$$C(t) = 5\cos(\omega t + 90^\circ) = \text{Real}[5e^{j(\omega t + 90^\circ)}]$$

Some Observations

$$\begin{aligned}
 v(t) &= 10\cos\omega t + 5\cos(\omega t - 90^\circ) - 5\cos(\omega t - 30^\circ) \\
 &= 10 + 5\angle -90^\circ - 5\angle -30^\circ \quad \text{In phasor notation} \\
 &= 6.18\angle -23^\circ
 \end{aligned}$$



Phasors are like vectors where the phase angle denotes the angle between coordinate axes with j representing 90°

Complex Impedances

Inductance:

Say a sinusoidal current is flowing in a circuit with inductance

$$i(t) = I_0 \sin(\omega t) = I_0 \cos(\omega t - 90^\circ)$$

$$v_L(t) = L \frac{di}{dt} = \omega L I_0 \cos \omega t$$

Therefore, the current in an inductor lags the voltage by 90°

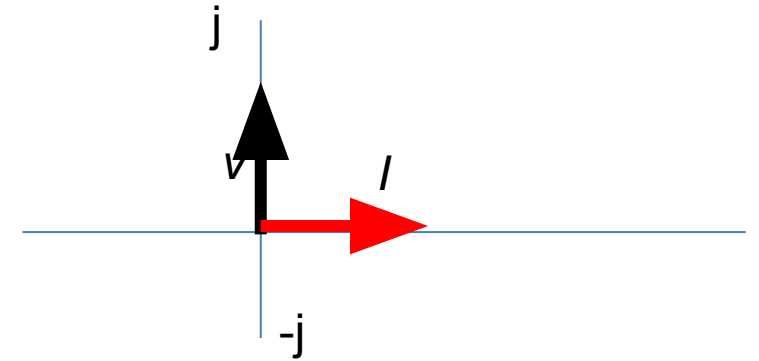
In the phasor notation

$$\mathbf{V} = \omega L I_0$$

$$\mathbf{I} = I_0 \angle -90^\circ$$

Then, inductive impedance

$$\mathbf{Z}_L = \frac{\mathbf{V}}{\mathbf{I}} = \frac{\omega L}{\angle -90^\circ} = \frac{\omega L}{-j} = j\omega L$$



We could have obtained the same result working directly with exponentials

$$v_L(t) = L \frac{d}{dt} \text{Real}[I_0 e^{j(\omega t - 90^\circ)}]$$

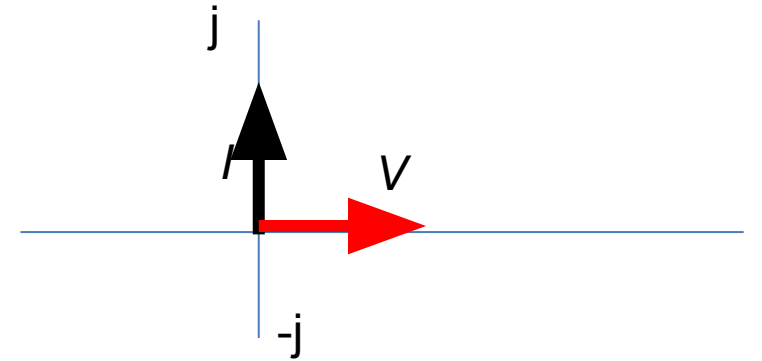
$$v_L(t) = \text{Real}[I_0 L j \omega e^{j(\omega t - 90^\circ)}]$$

$$v_L(t) = j\omega L I_0 \angle -90^\circ$$

$$\mathbf{V} = j\omega L \mathbf{I}$$

Complex Impedances

Capacitance:



Complex Impedances

- AC steady-state analysis using phasors allows us to express the relationship between current and voltage using a formula that looks like Ohm's law:

$$\mathbf{V} = \mathbf{IZ}$$

- Impedance depends on the frequency ω .
- Impedance is a complex number.
- Impedance allows us to use the same solution techniques for AC steady state as