

# EE16B

# Designing Information Devices and Systems II

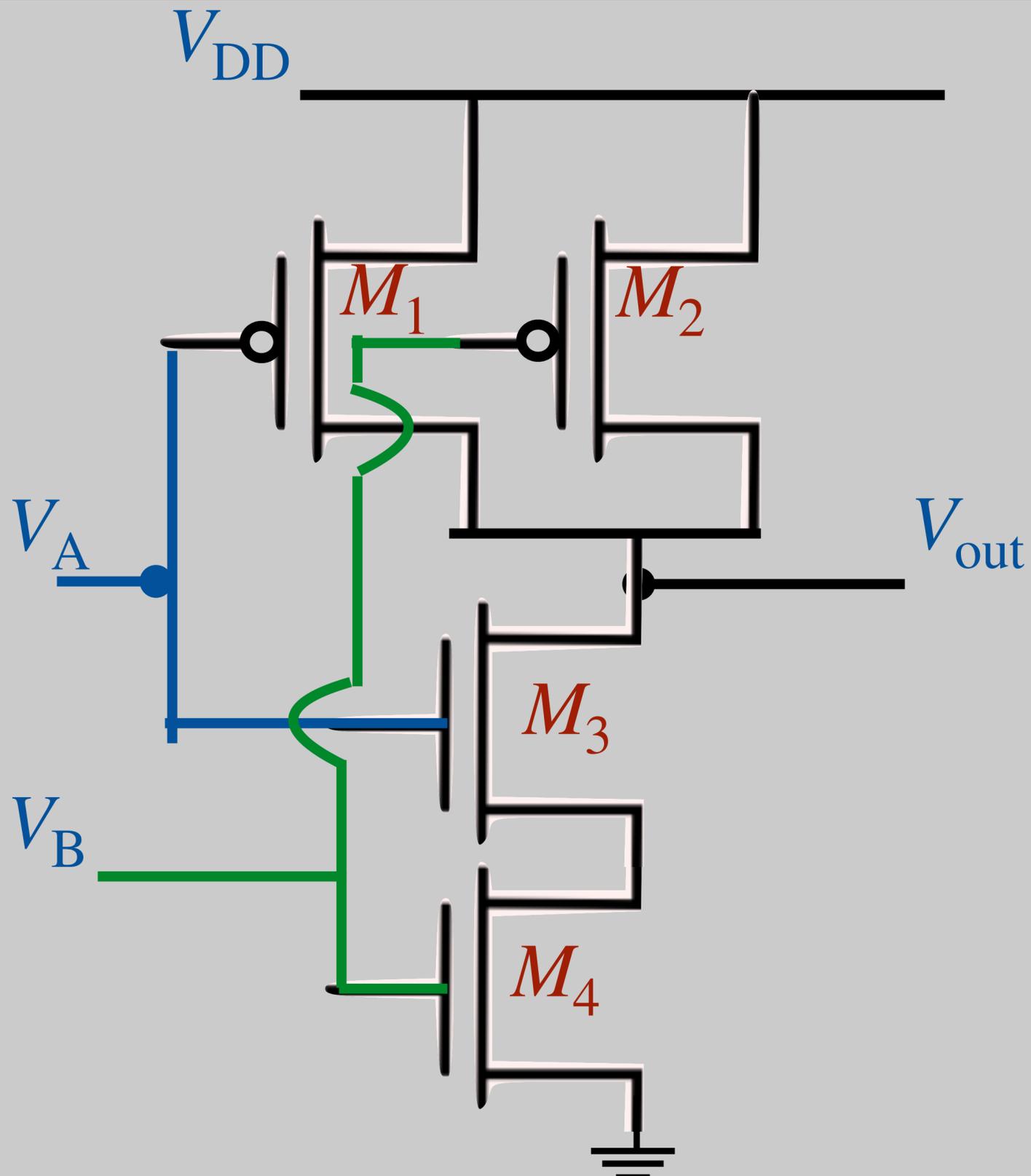
## Lecture 3A Transistors II

# Announcements

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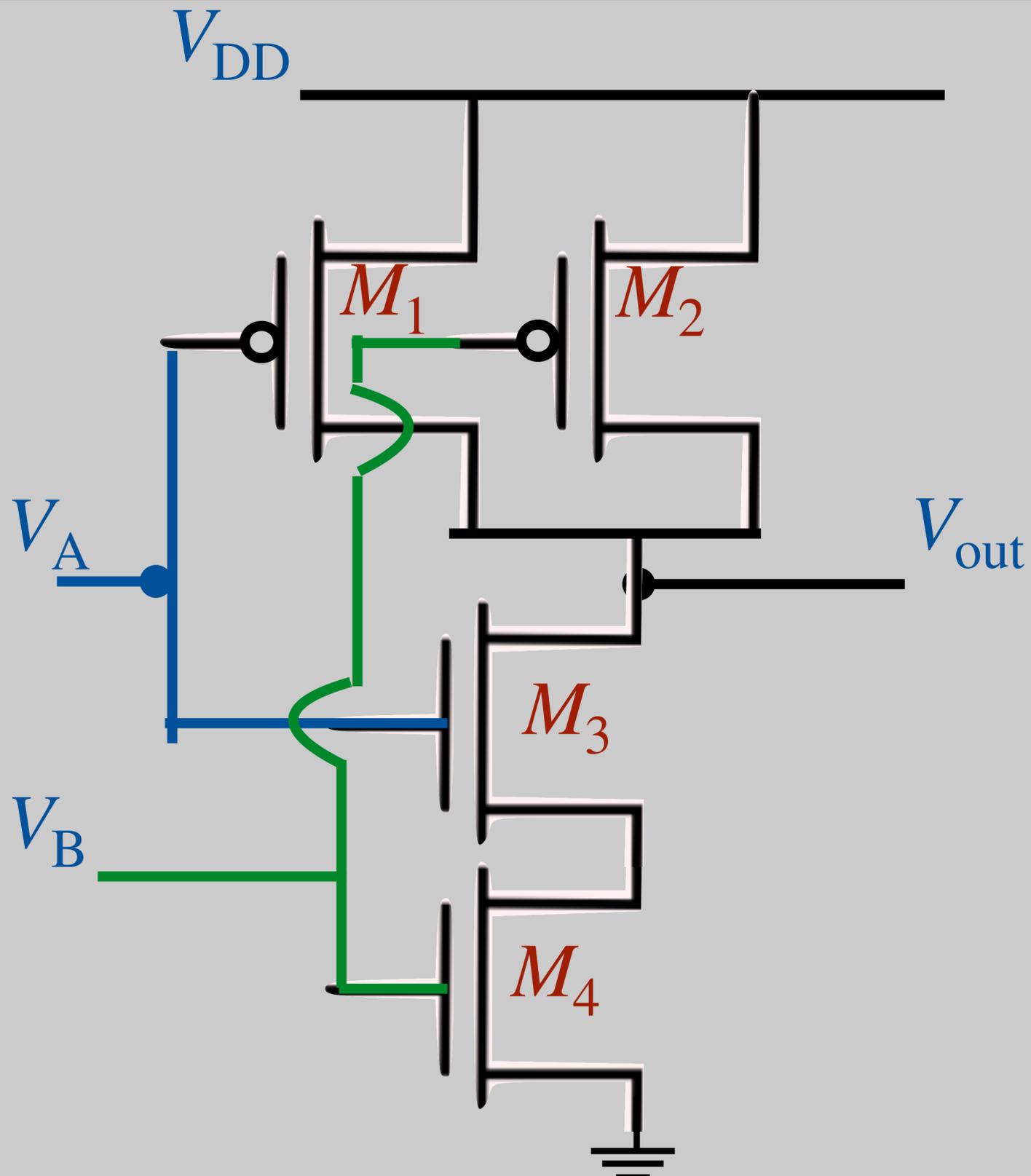
- Last time:
  - Lab 2
  - Transistors as a switch
- Today:
  - Transistors Switch RC model
  - Finish review of complex numbers
    - Euler Formula
  - Solution for sinusoidal input

# CMOS (Complementary MOS)



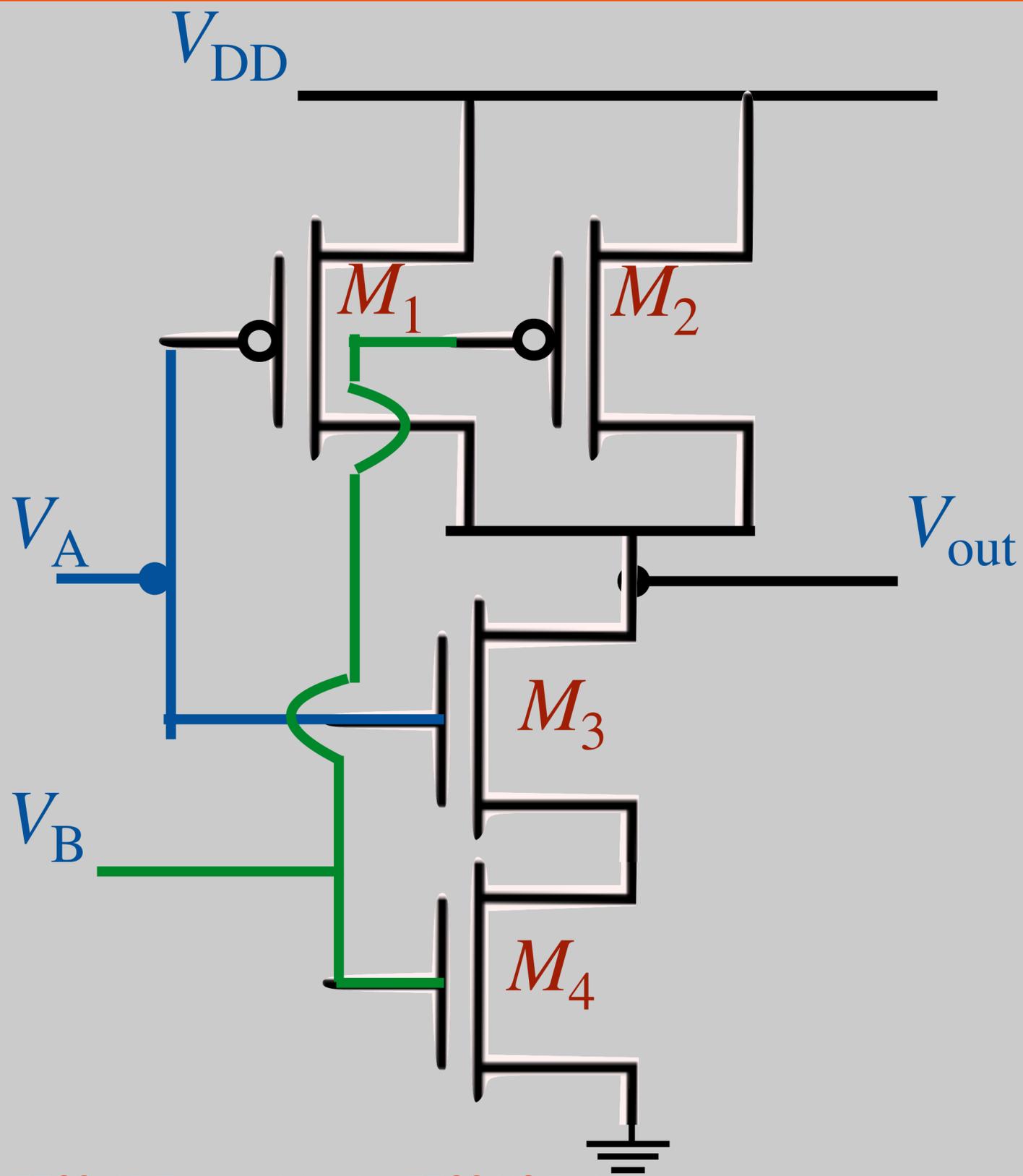
$V_A$	$V_B$	$M_1$	$M_2$	$M_3$	$M_4$	$V_{out}$
0	0					
0	1					
1	0					
1	1					

# CMOS (Complementary MOS)



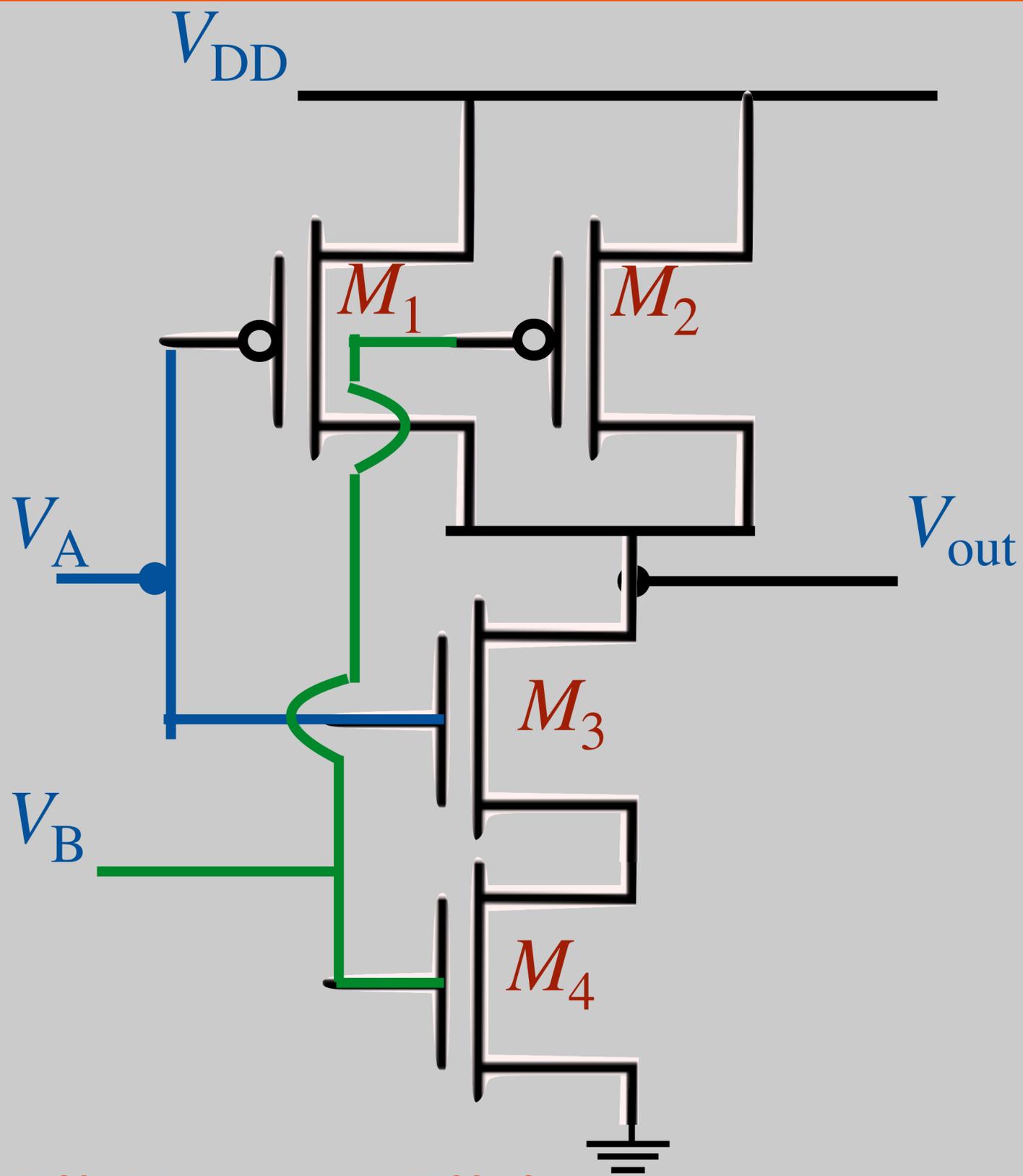
$V_A$	$V_B$	$M_1$	$M_2$	$M_3$	$M_4$	$V_{out}$
0	0	On	On	Off	Off	$V_{DD}$
0	1					
1	0					
1	1					

# CMOS (Complementary MOS)



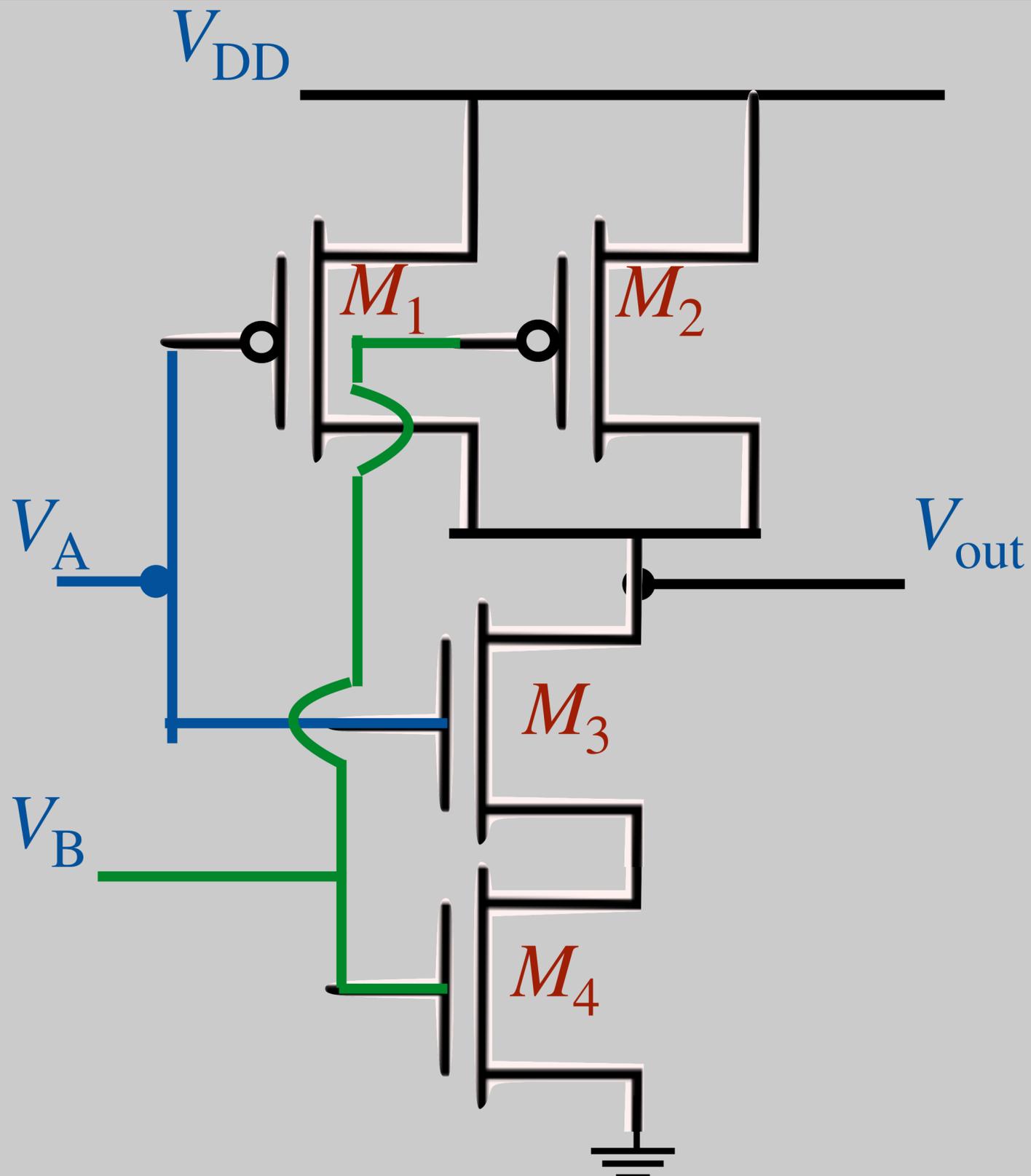
$V_A$	$V_B$	$M_1$	$M_2$	$M_3$	$M_4$	$V_{out}$
0	0	On	On	Off	Off	$V_{DD}$
0	1	On	Off	Off	On	$V_{DD}$
1	0	Off	On	On	Off	$V_{DD}$
1	1	Off	Off	Off	On	0

# CMOS (Complementary MOS)



$V_A$	$V_B$	$M_1$	$M_2$	$M_3$	$M_4$	$V_{out}$
0	0	On	On	Off	Off	$V_{DD}$
0	1	On	Off	Off	On	$V_{DD}$
1	0	Off	On	On	Off	$V_{DD}$
1	1					

# CMOS (Complementary MOS)

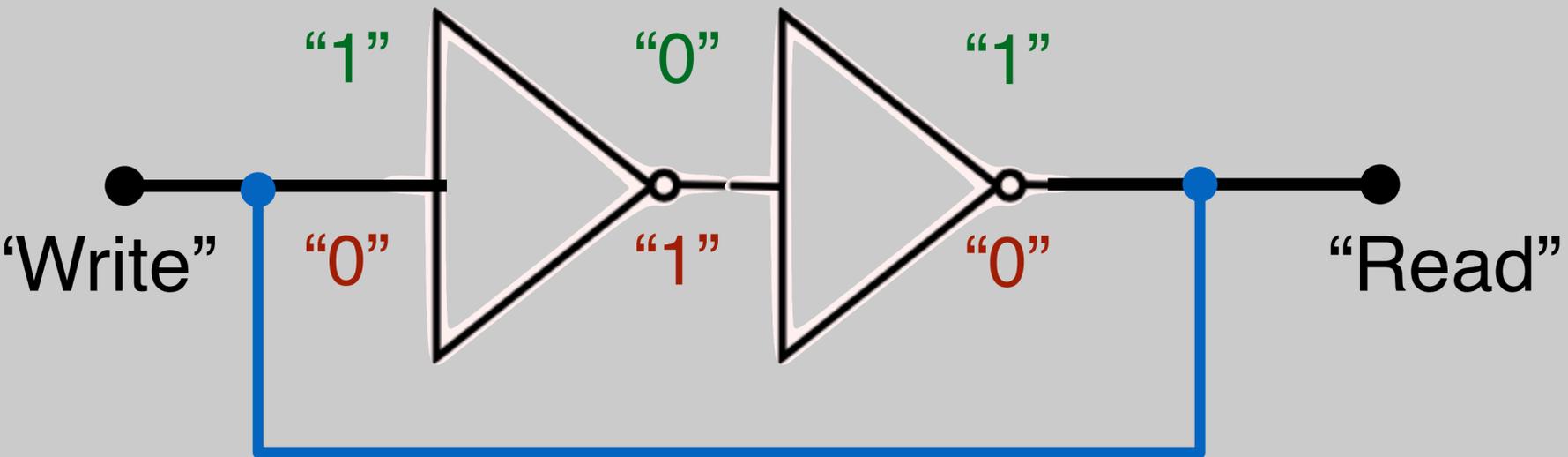


$V_A$	$V_B$	$M_1$	$M_2$	$M_3$	$M_4$	$V_{out}$
0	0	On	On	Off	Off	$V_{DD}$
0	1	On	Off	Off	On	$V_{DD}$
1	0	Off	On	On	Off	$V_{DD}$
1	1	Off	Off	On	On	0

NAND gate!

# Static RAM (SRAM)

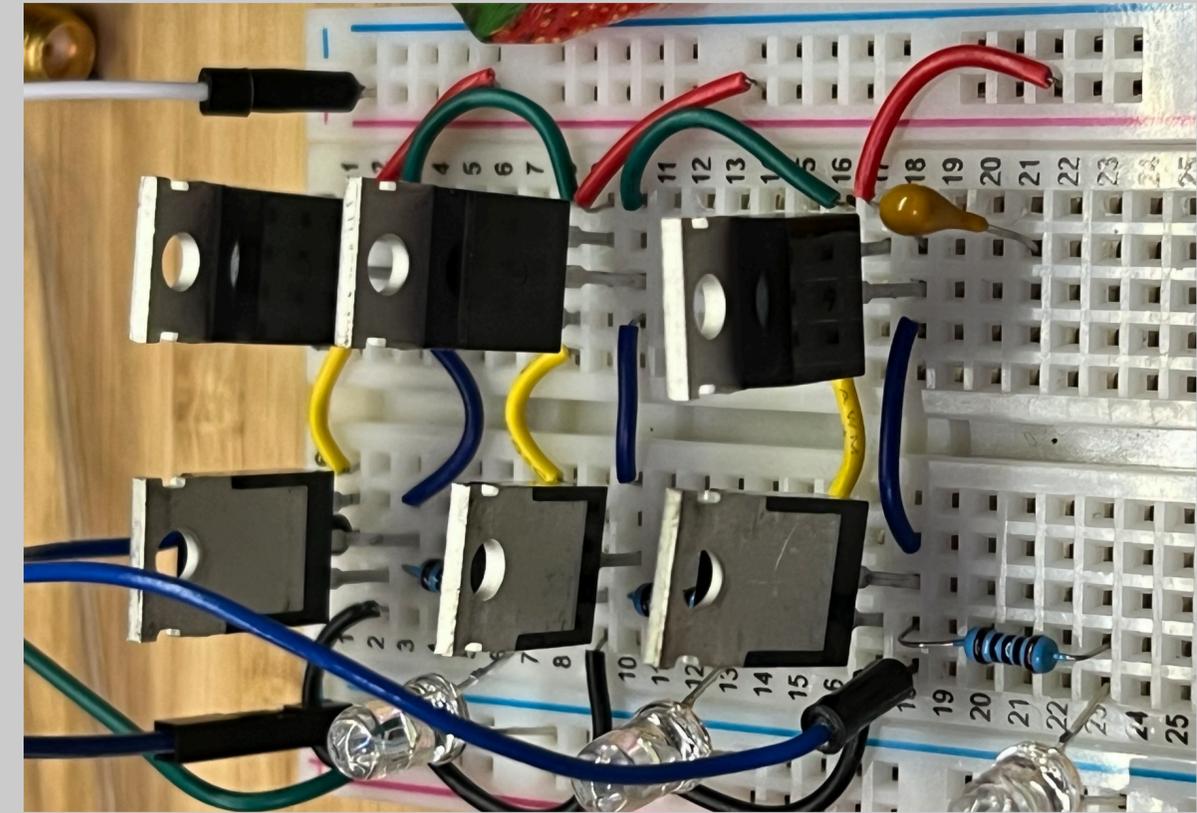
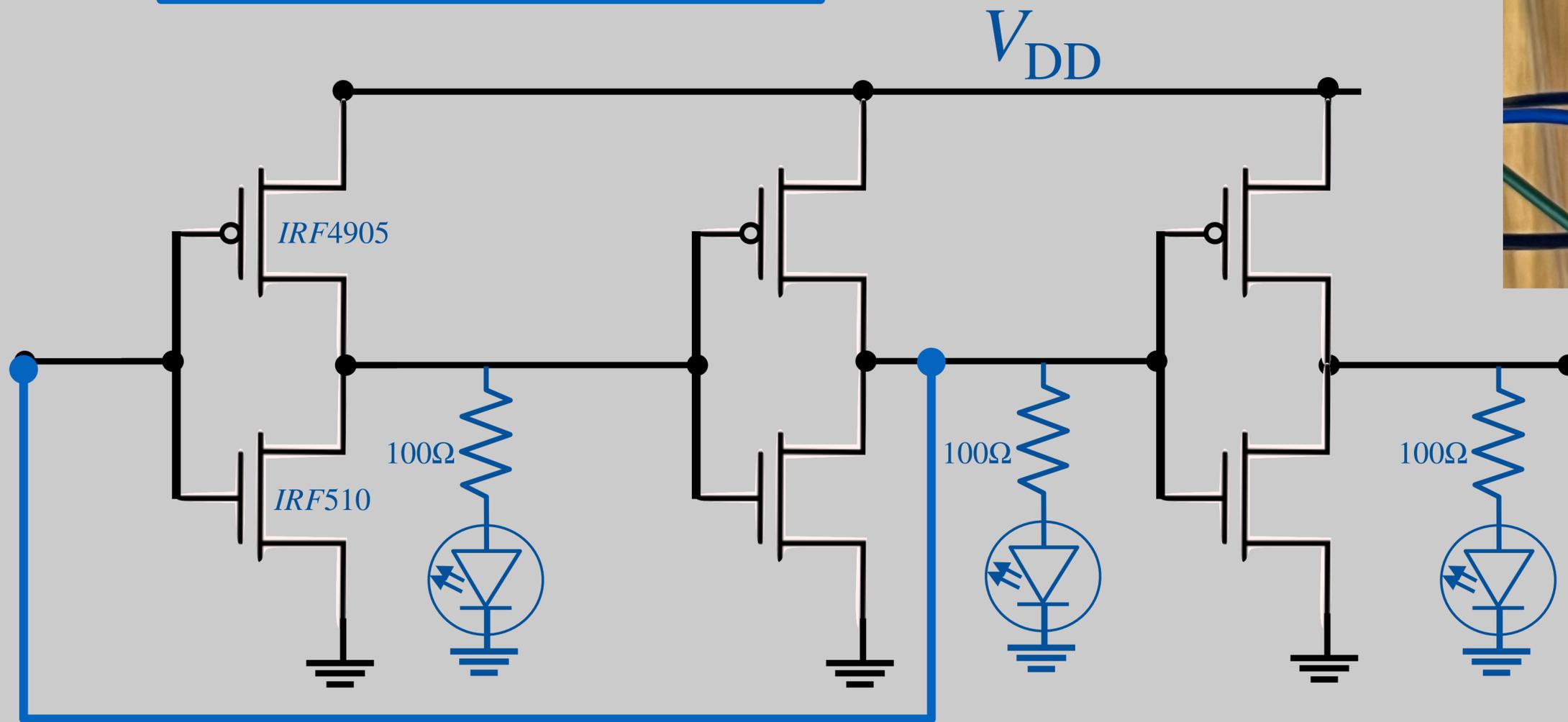
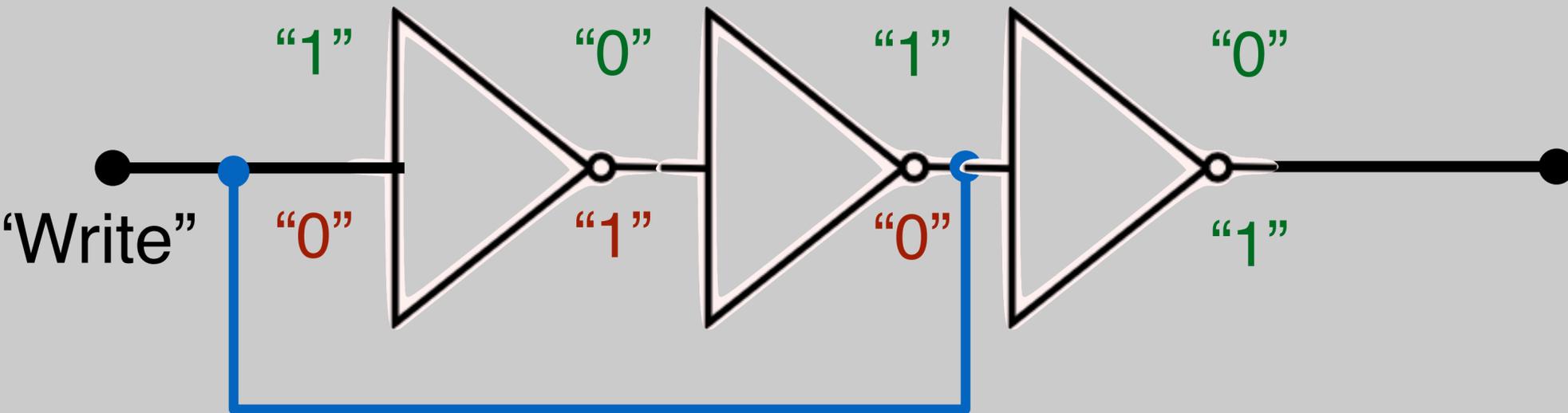
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2 stable states (bistable)

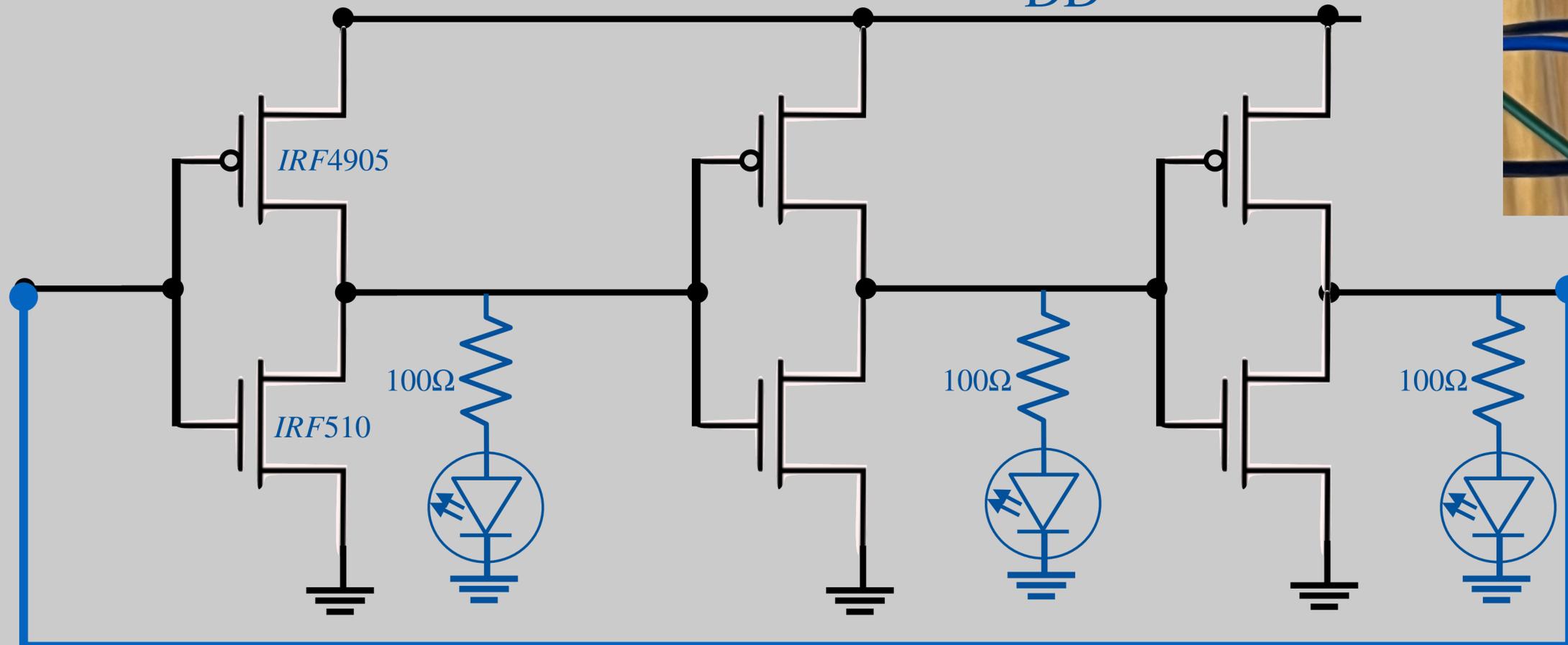
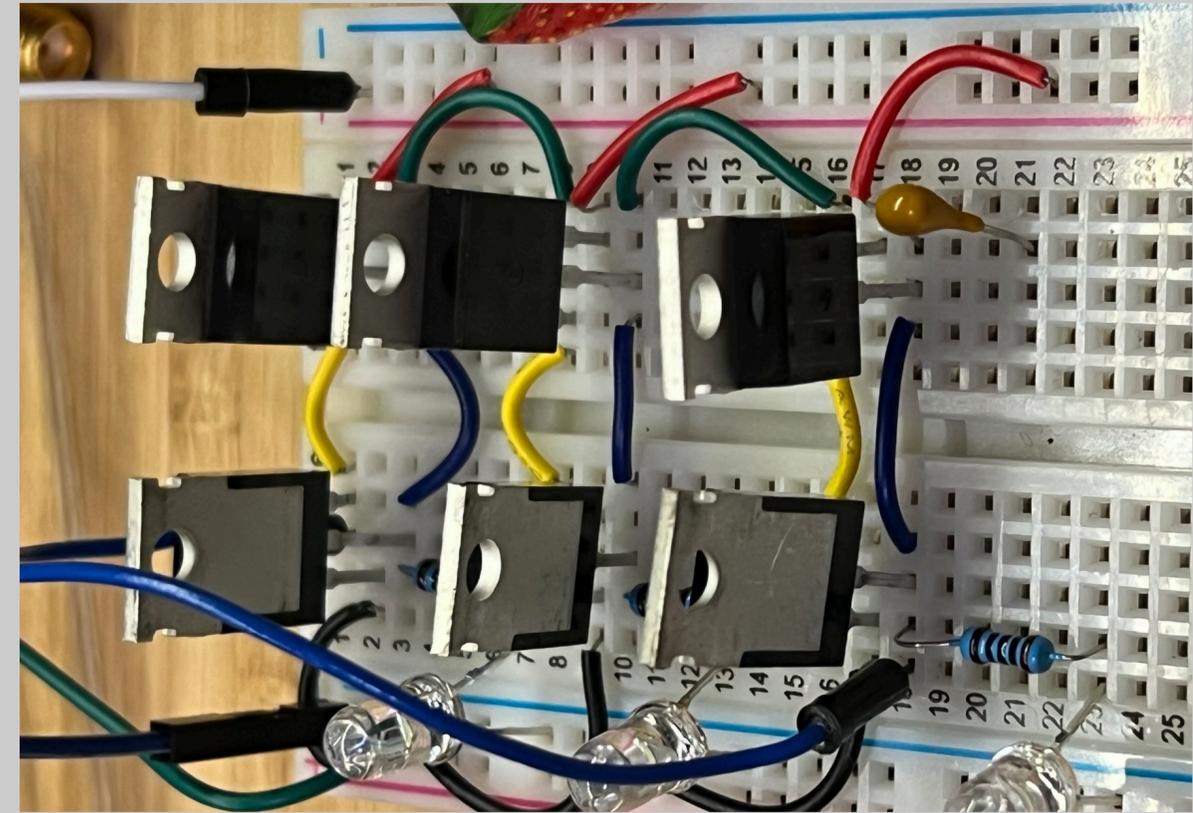
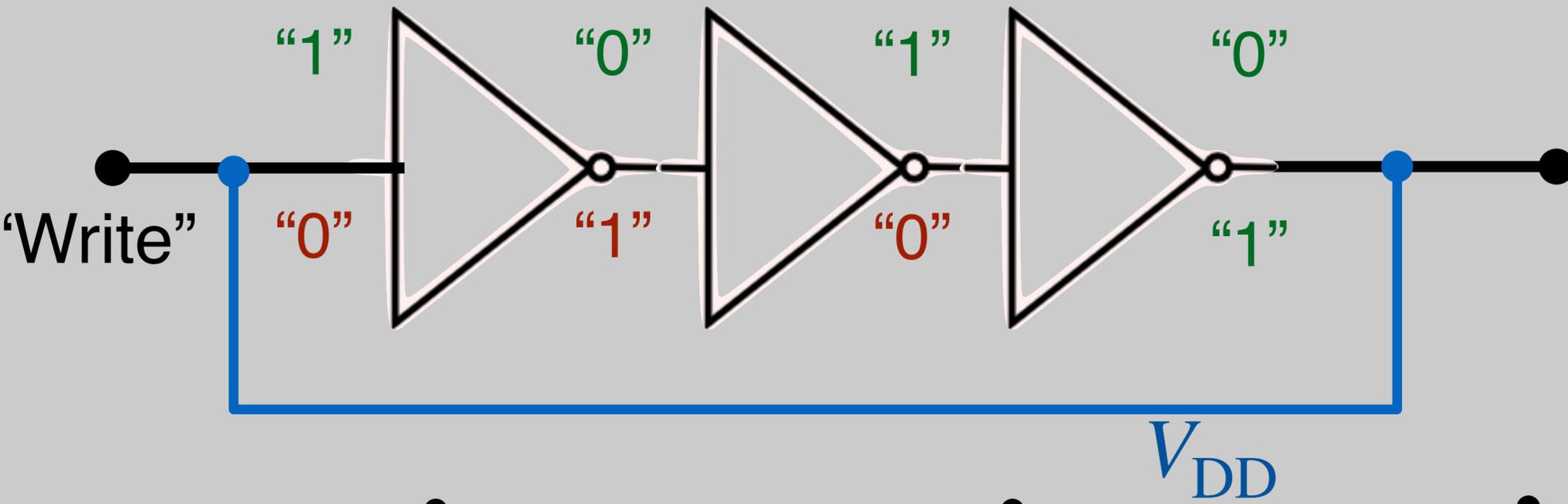
“Write” signal forces circuit to change states.... Memory!

# Demo

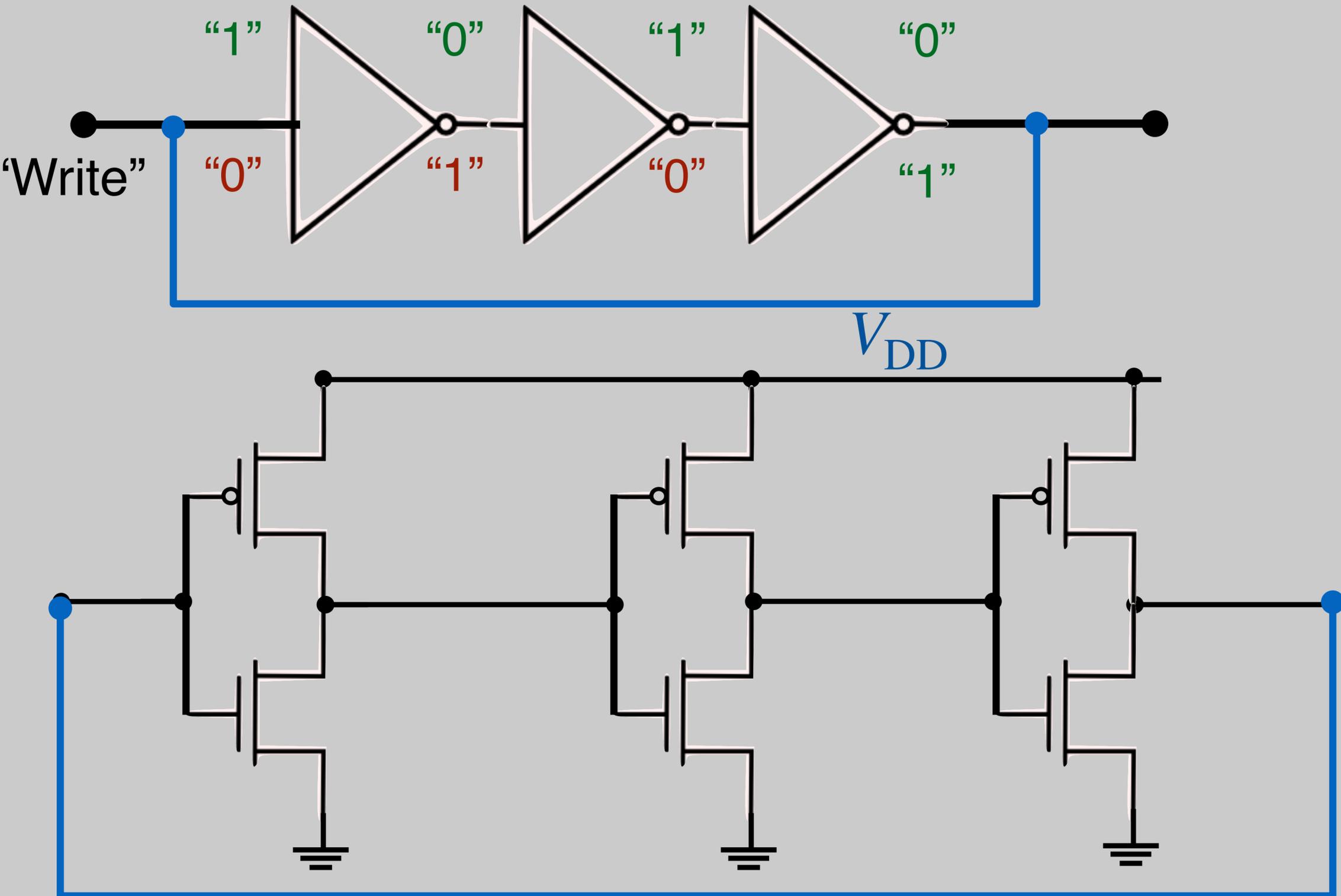


# Demo

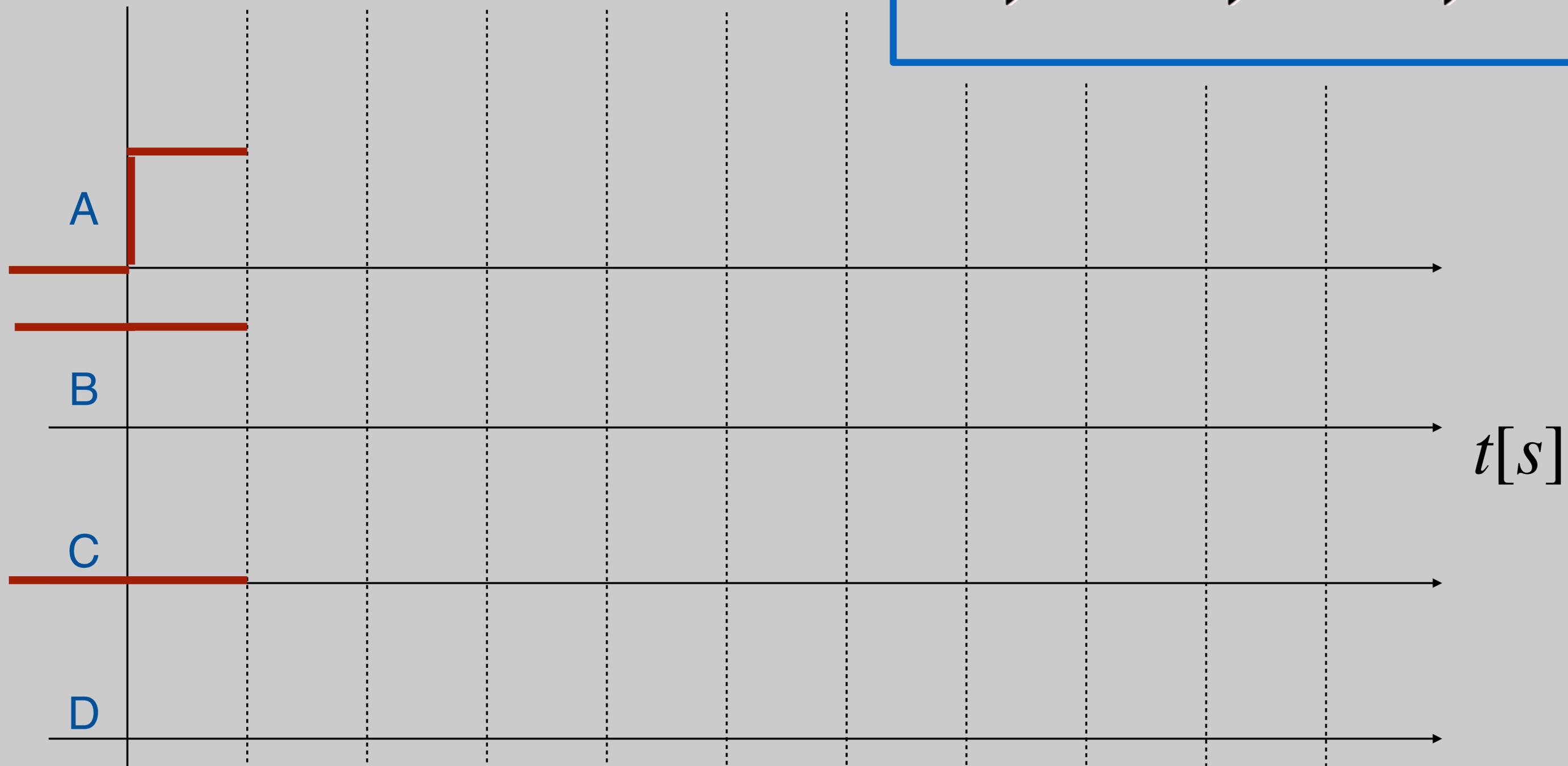
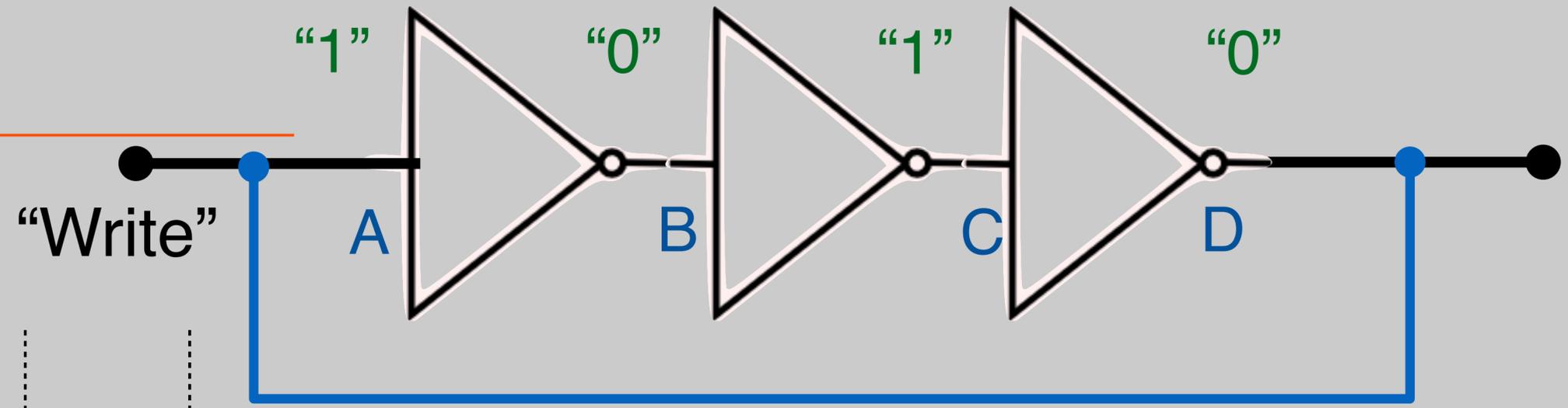
Inconsistent!



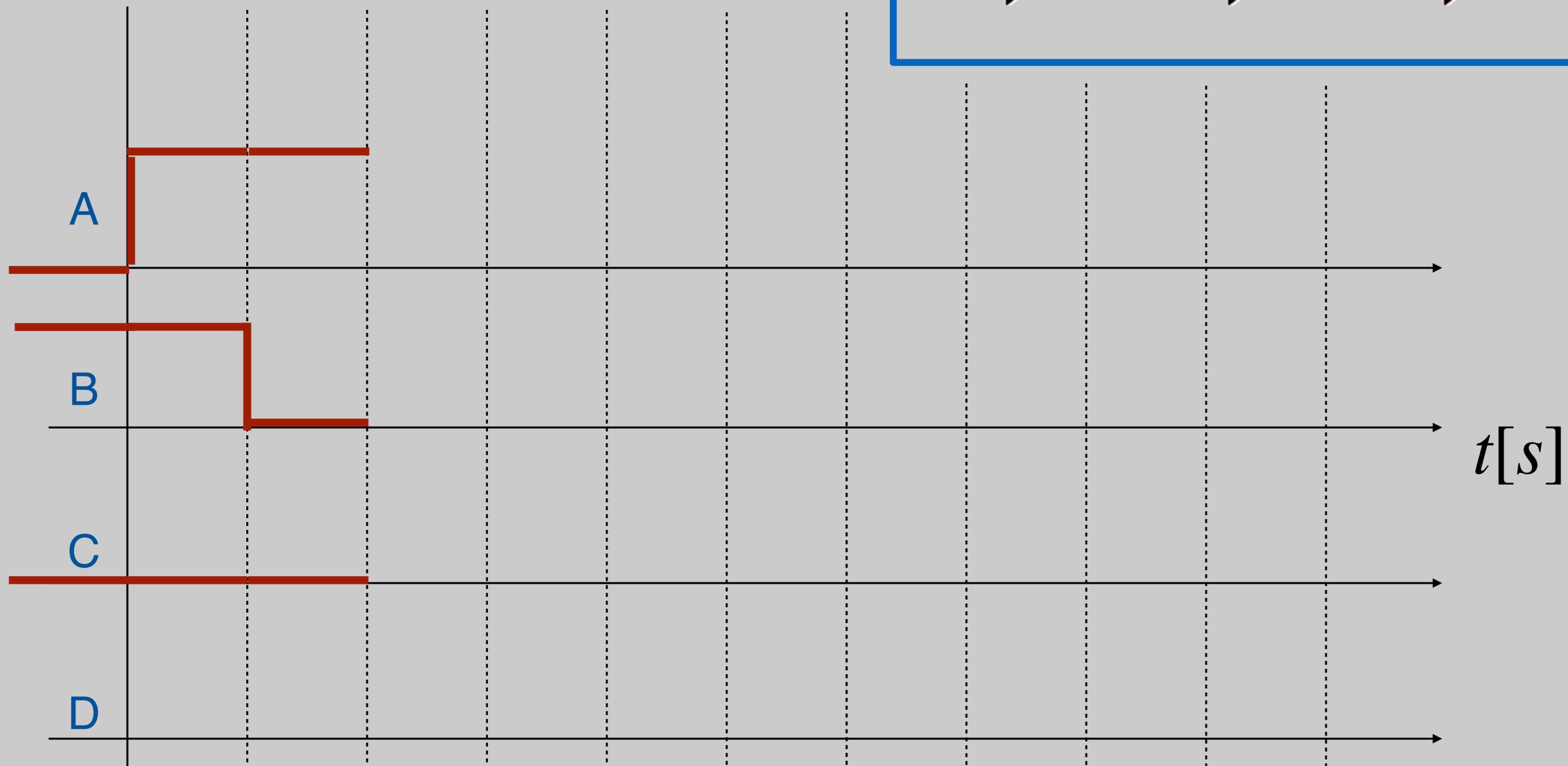
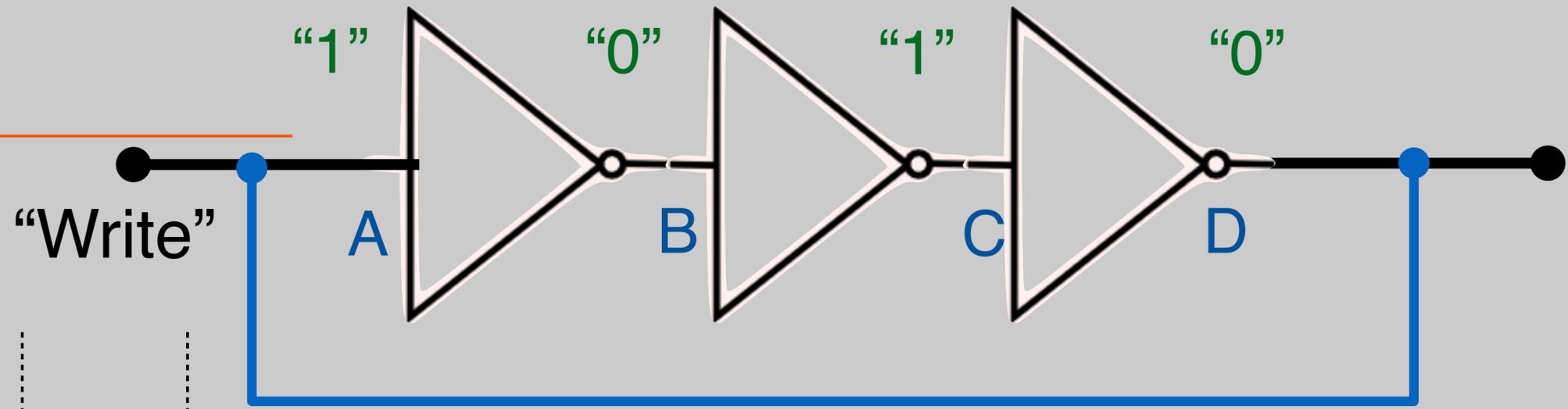
# Ring Oscillator



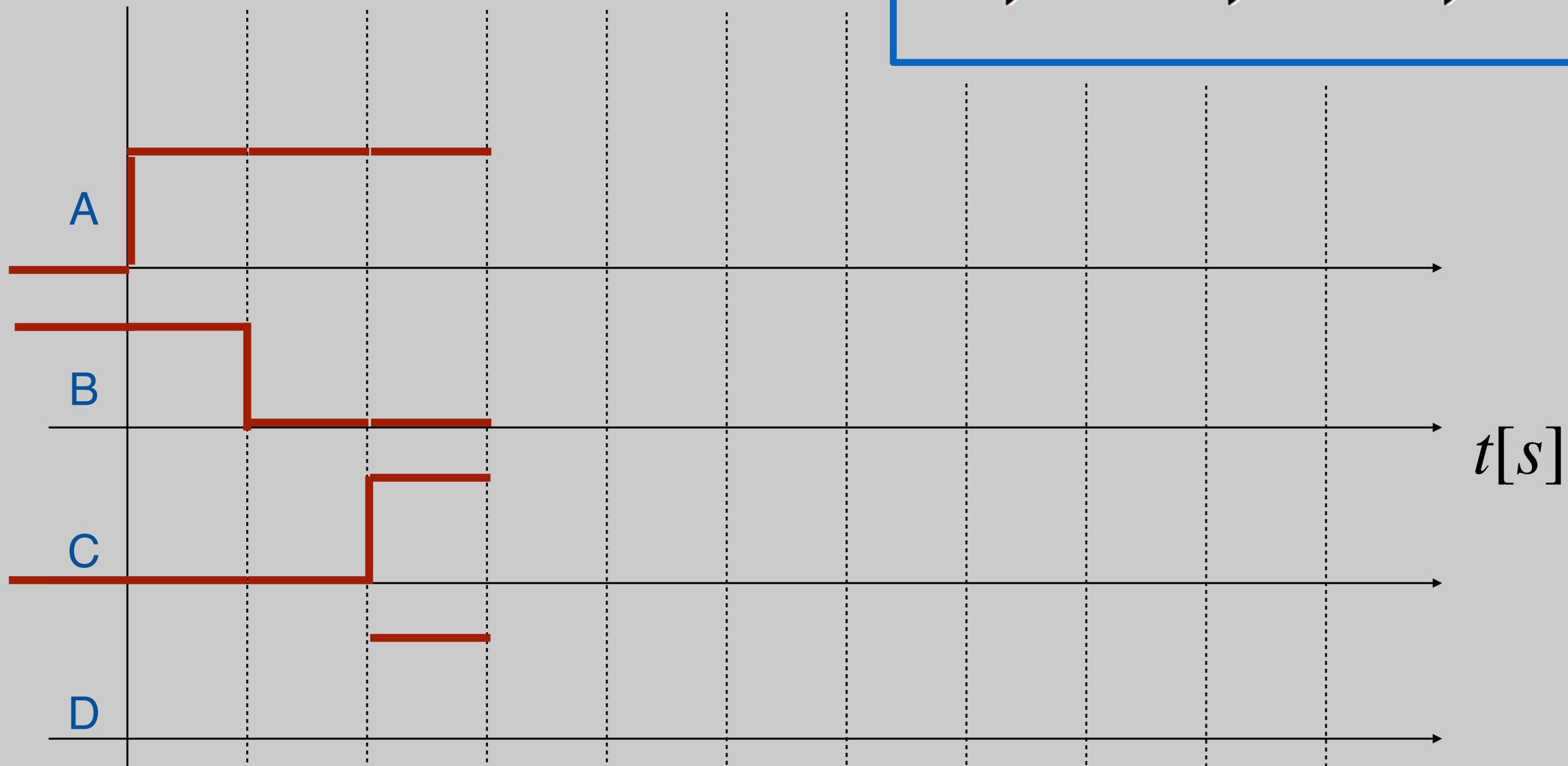
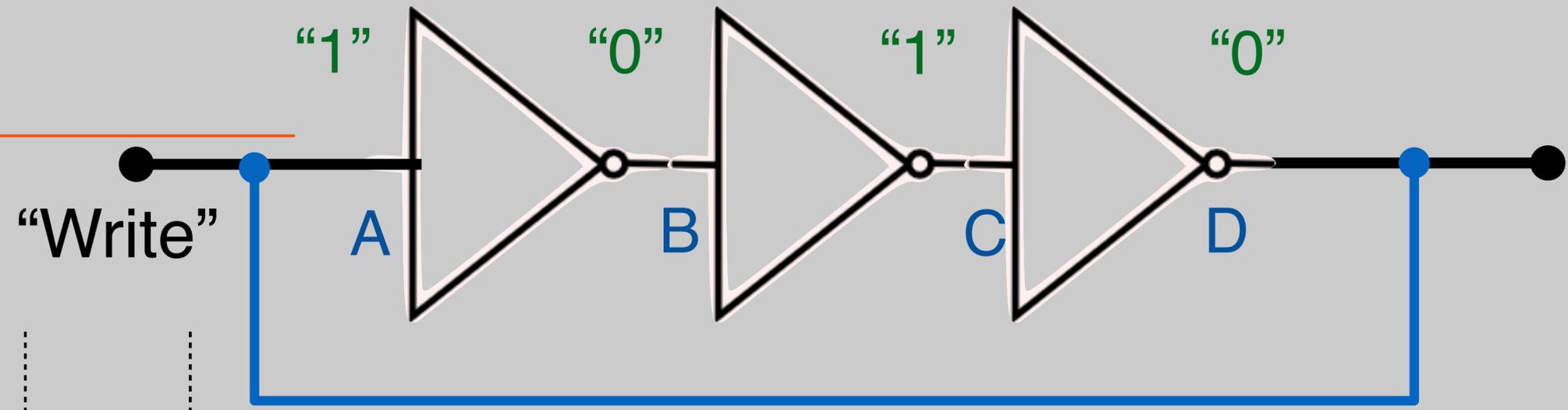
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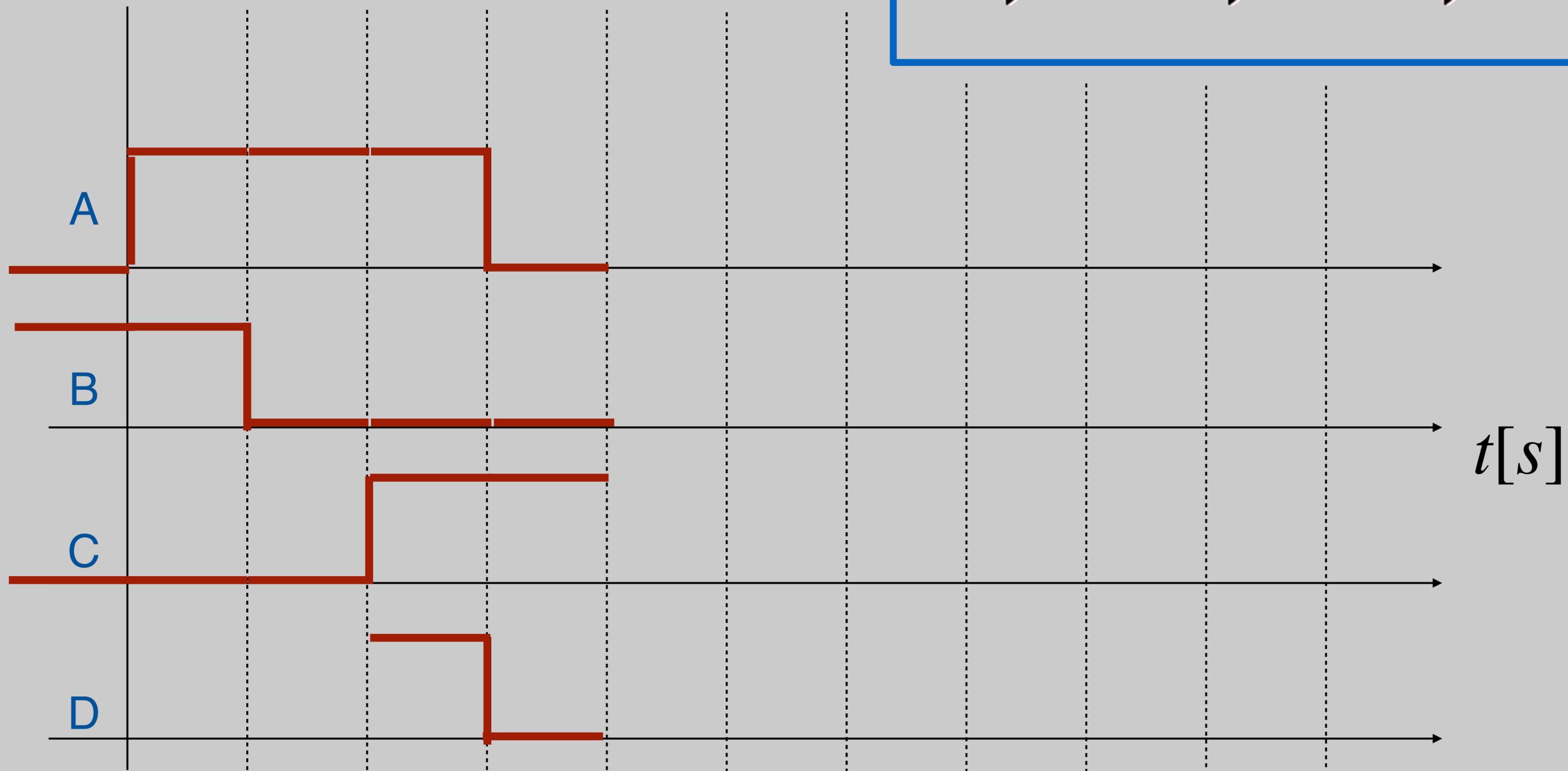
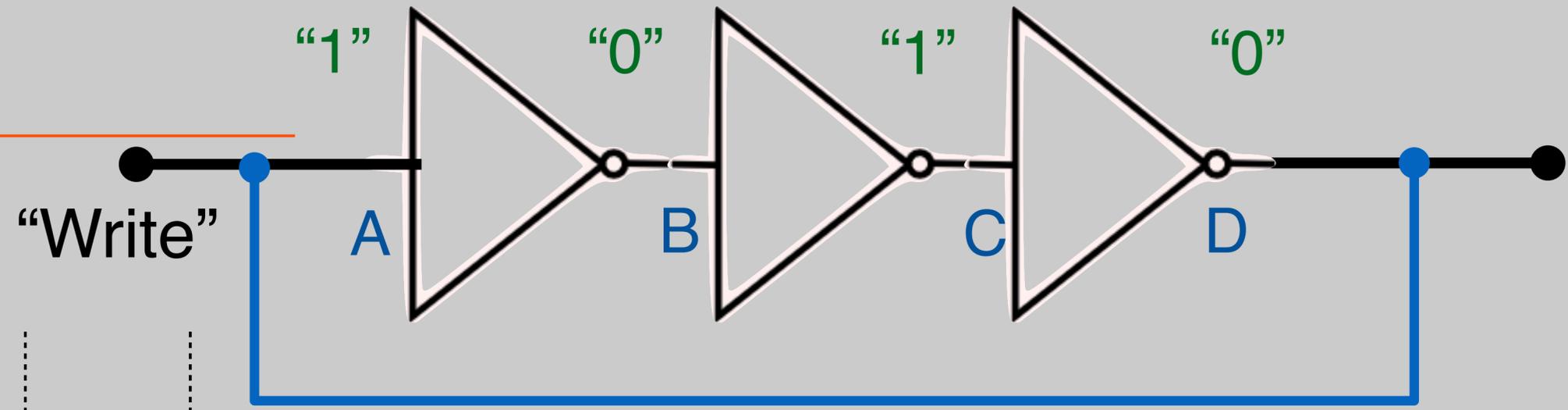
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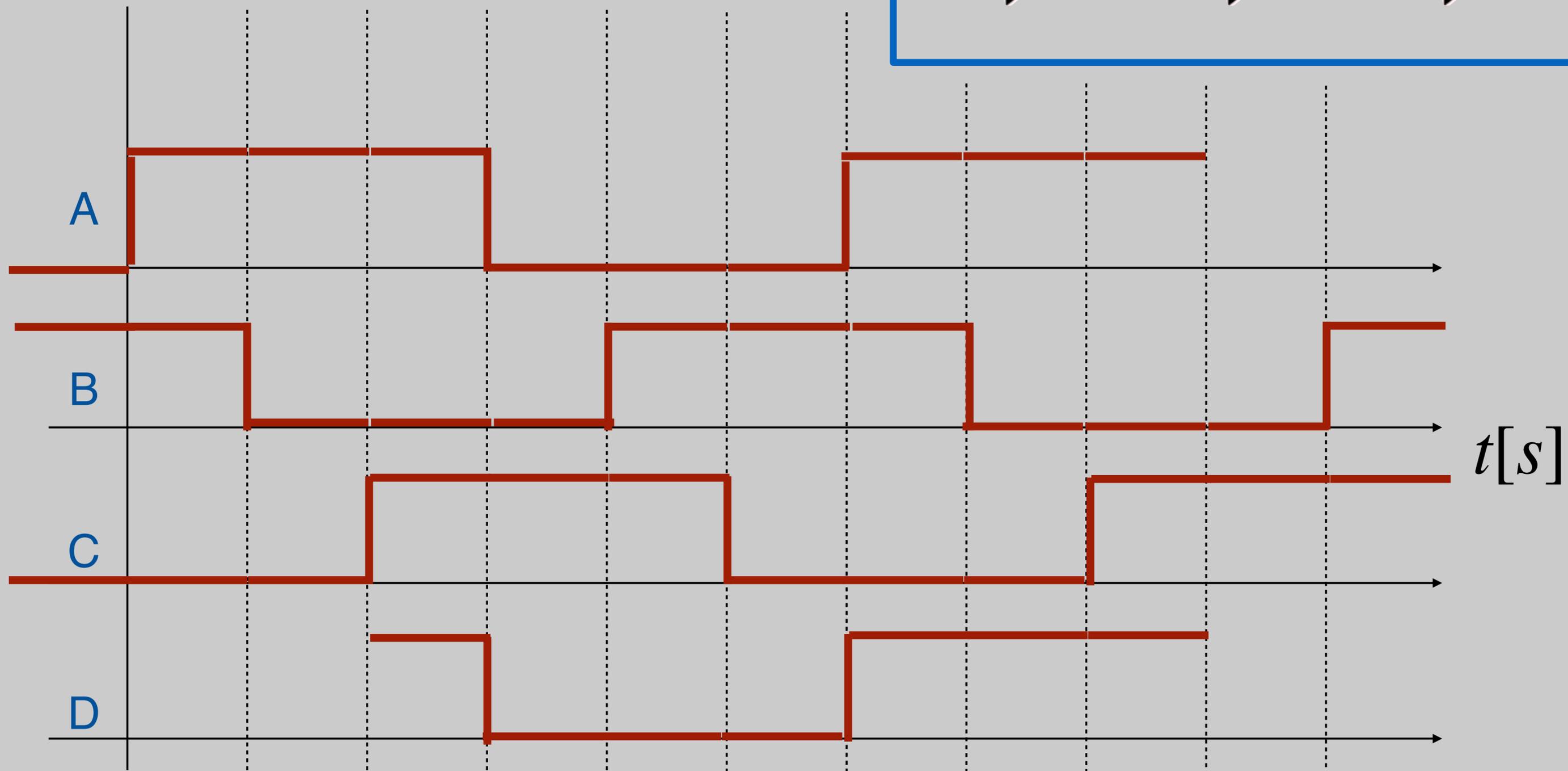
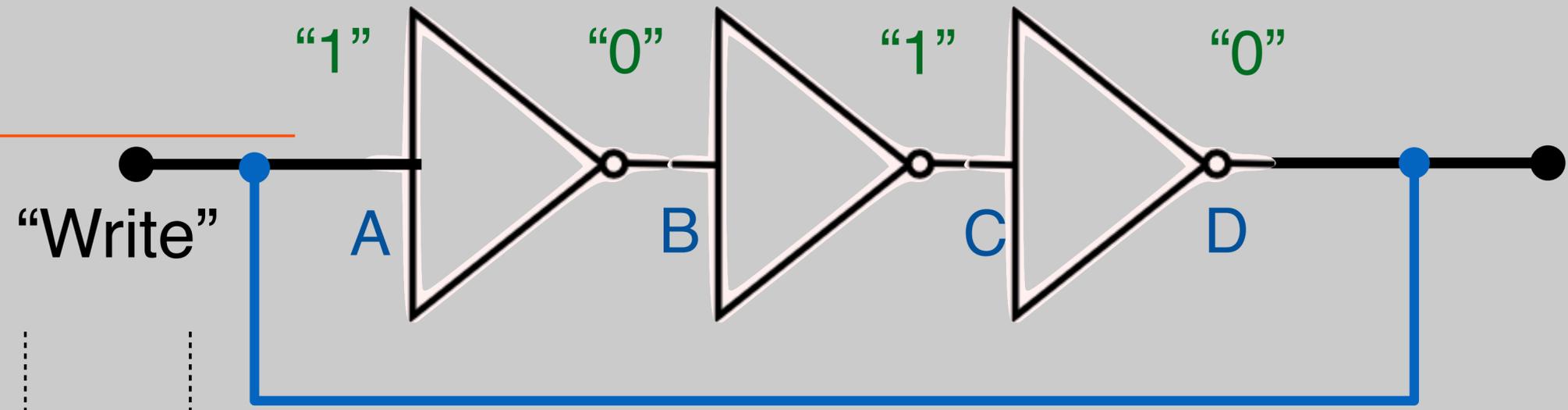
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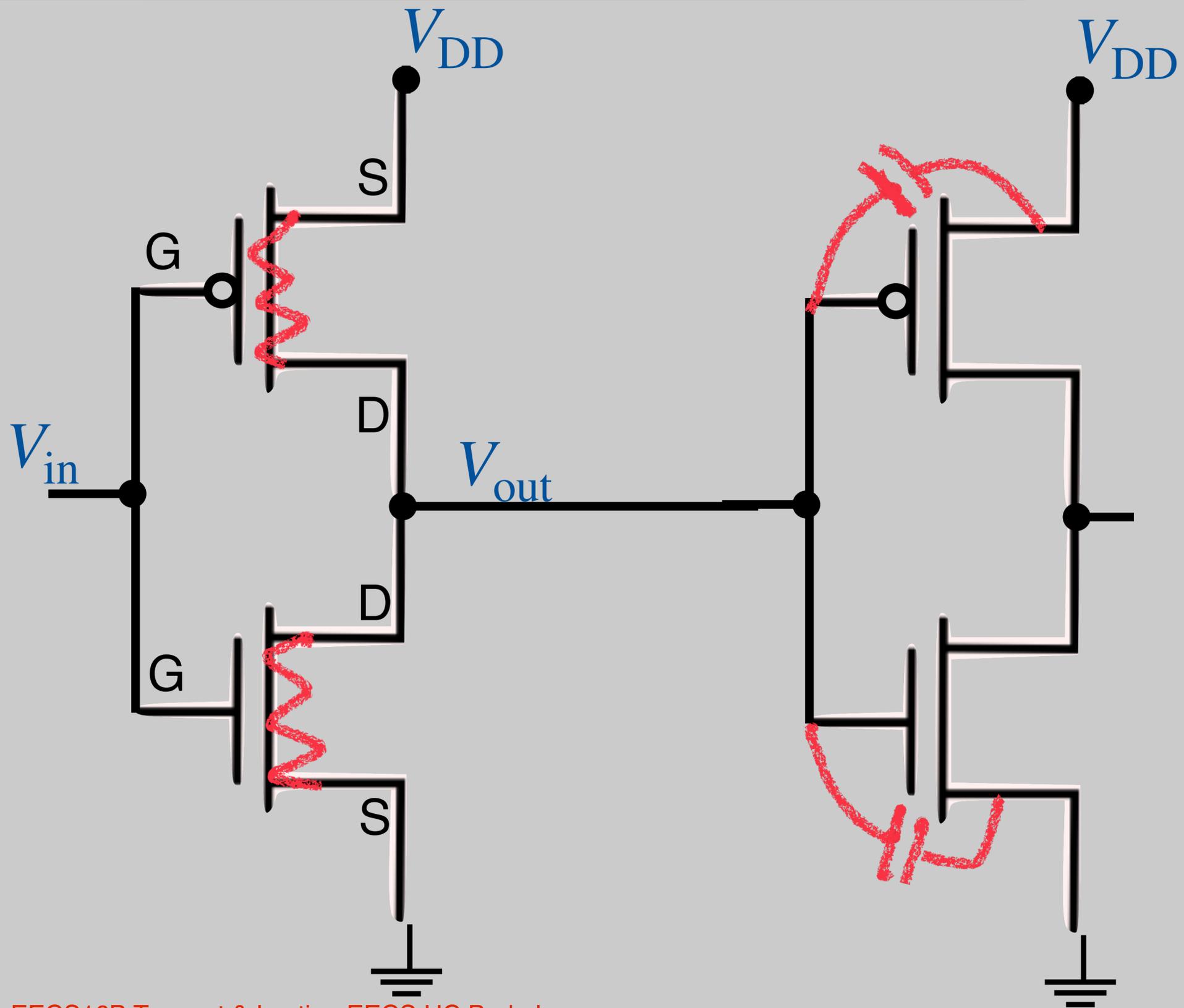
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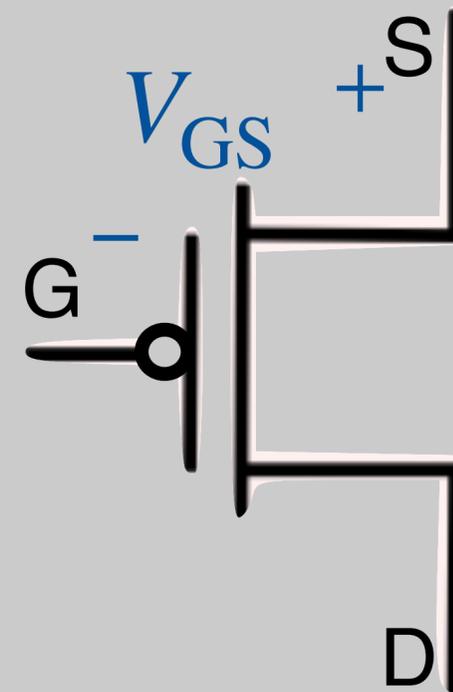
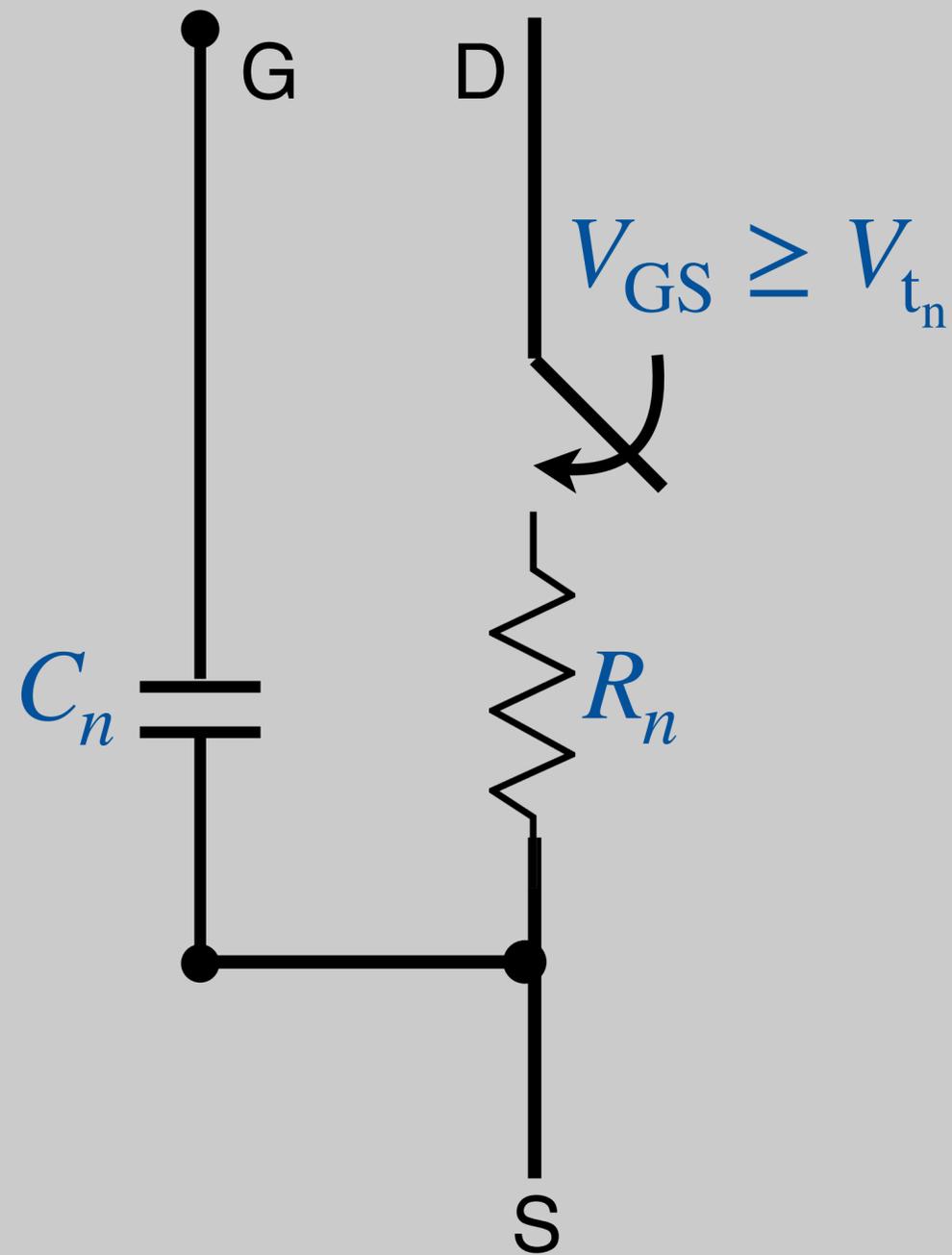
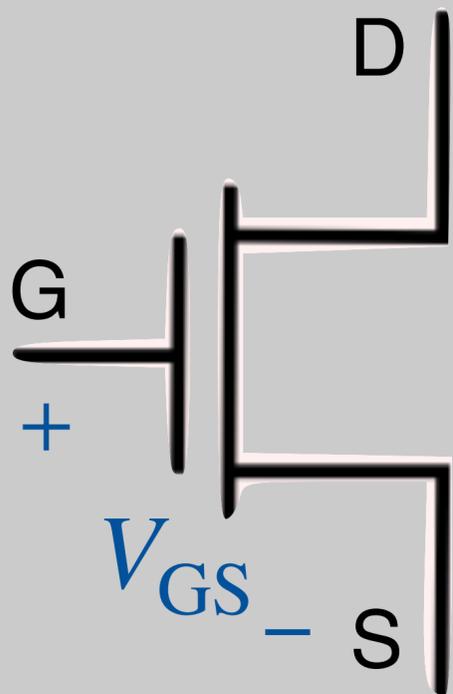
# Ring Oscillator



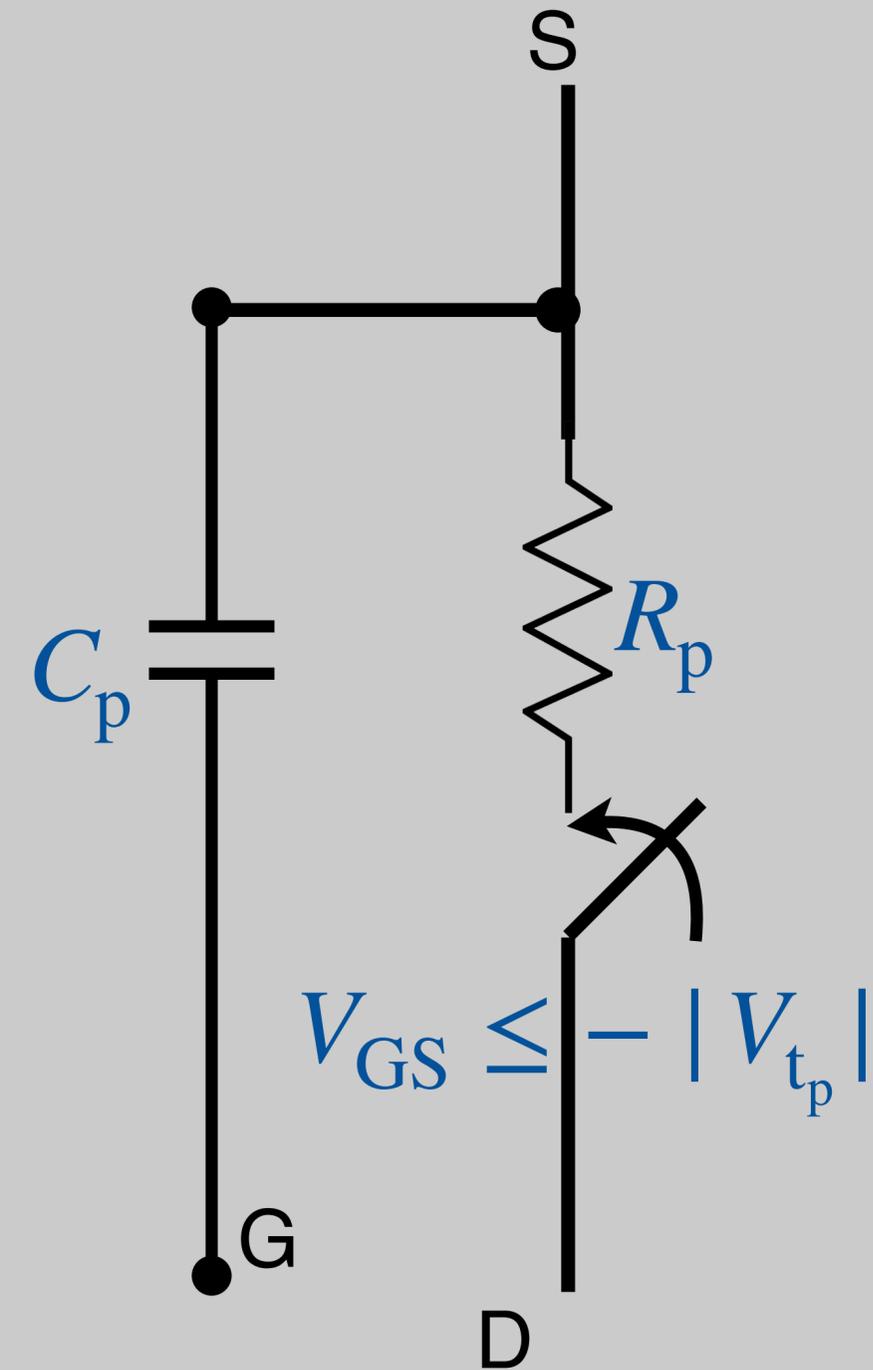
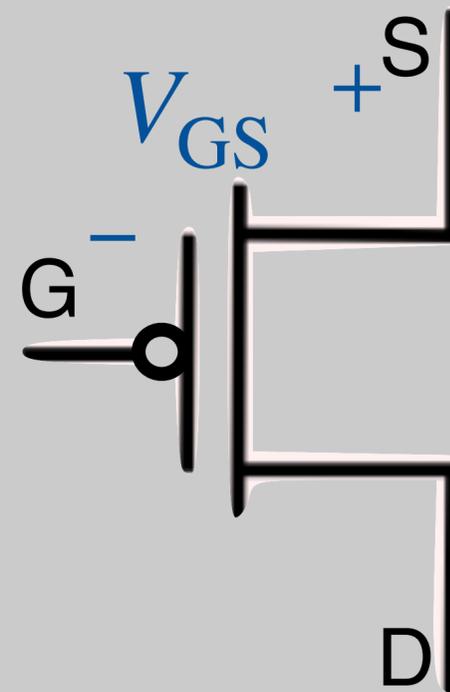
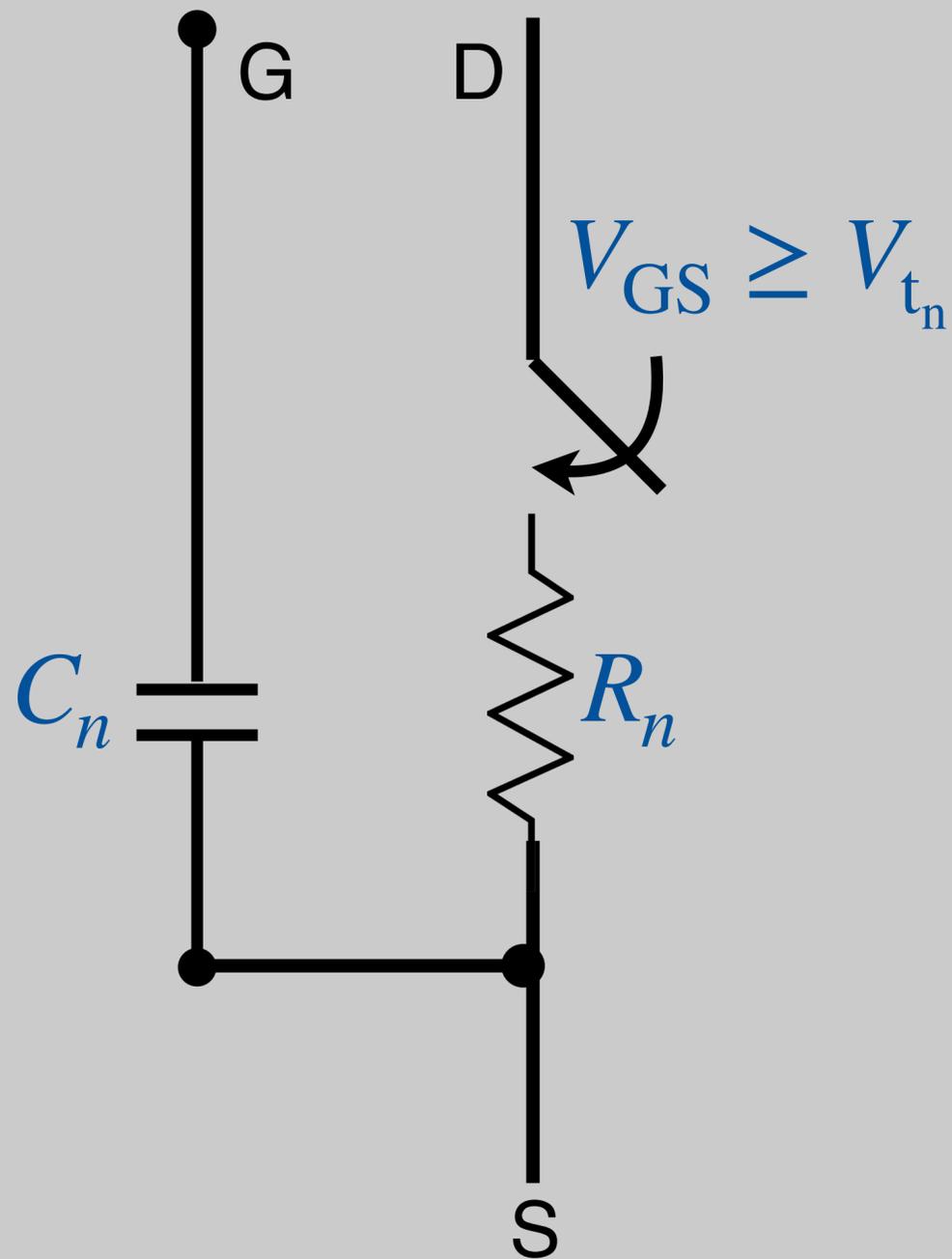
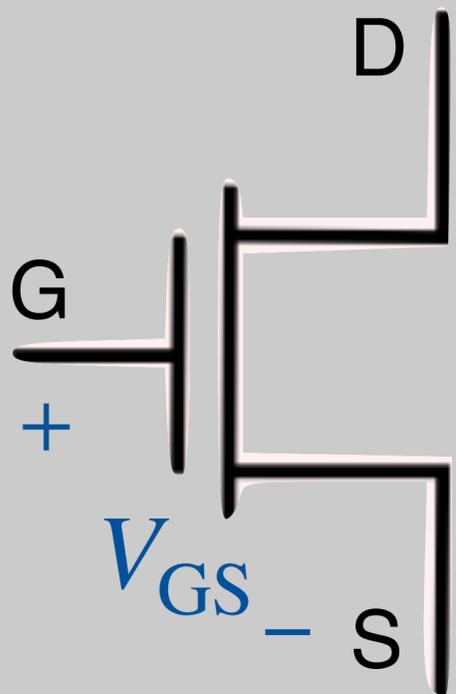
# RC Delay in MOS Circuits



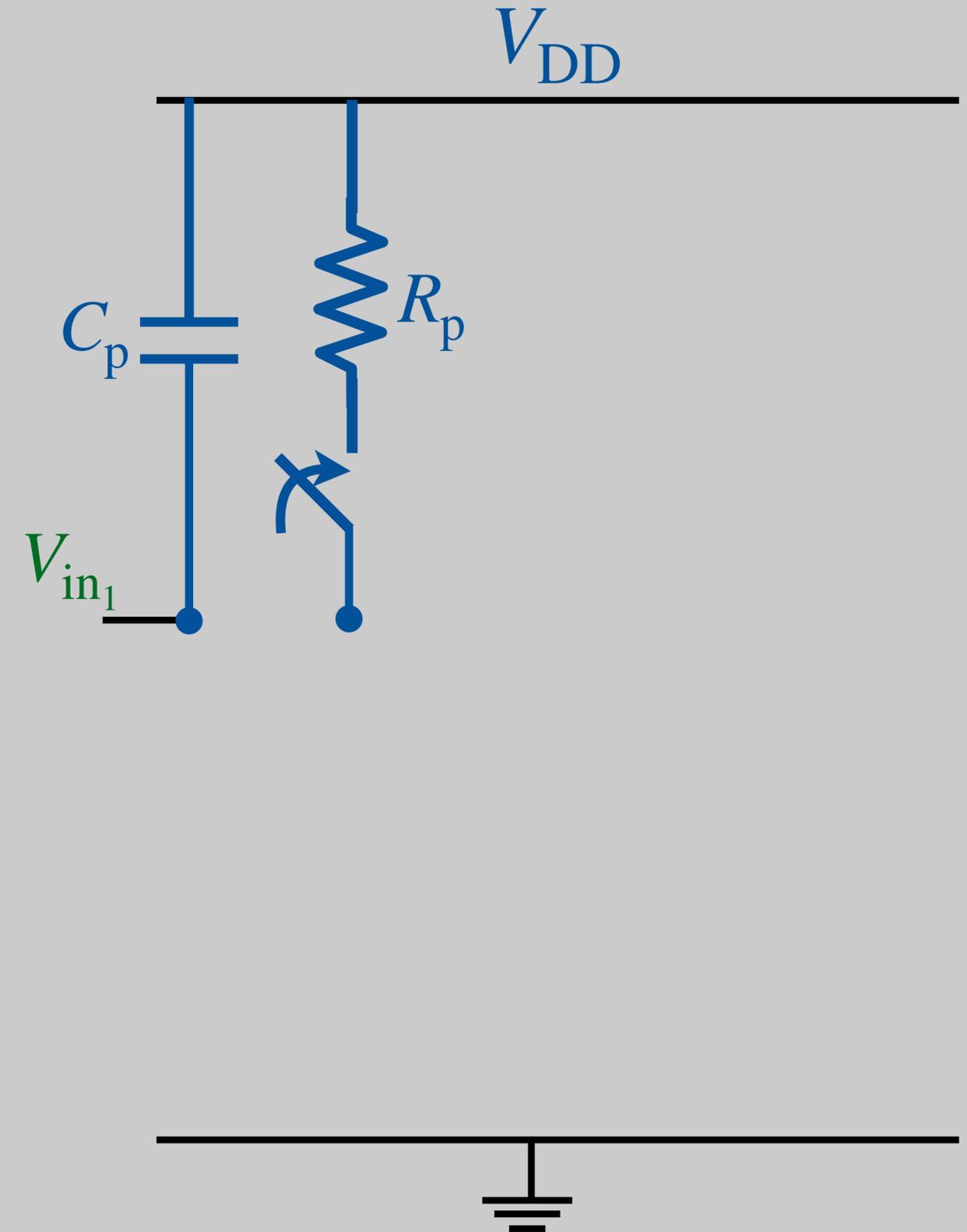
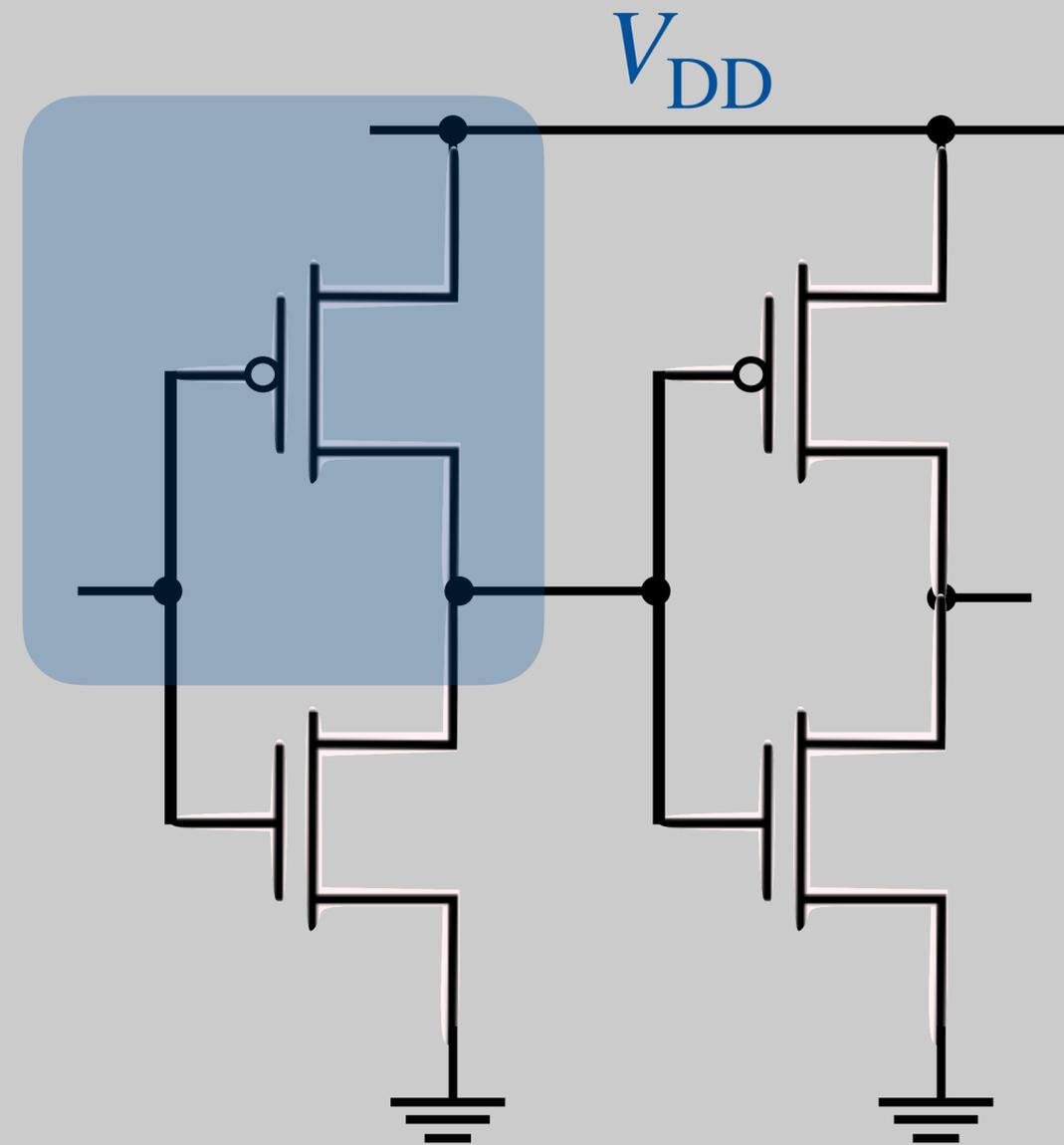
# RC Model of NMOS and PMOS



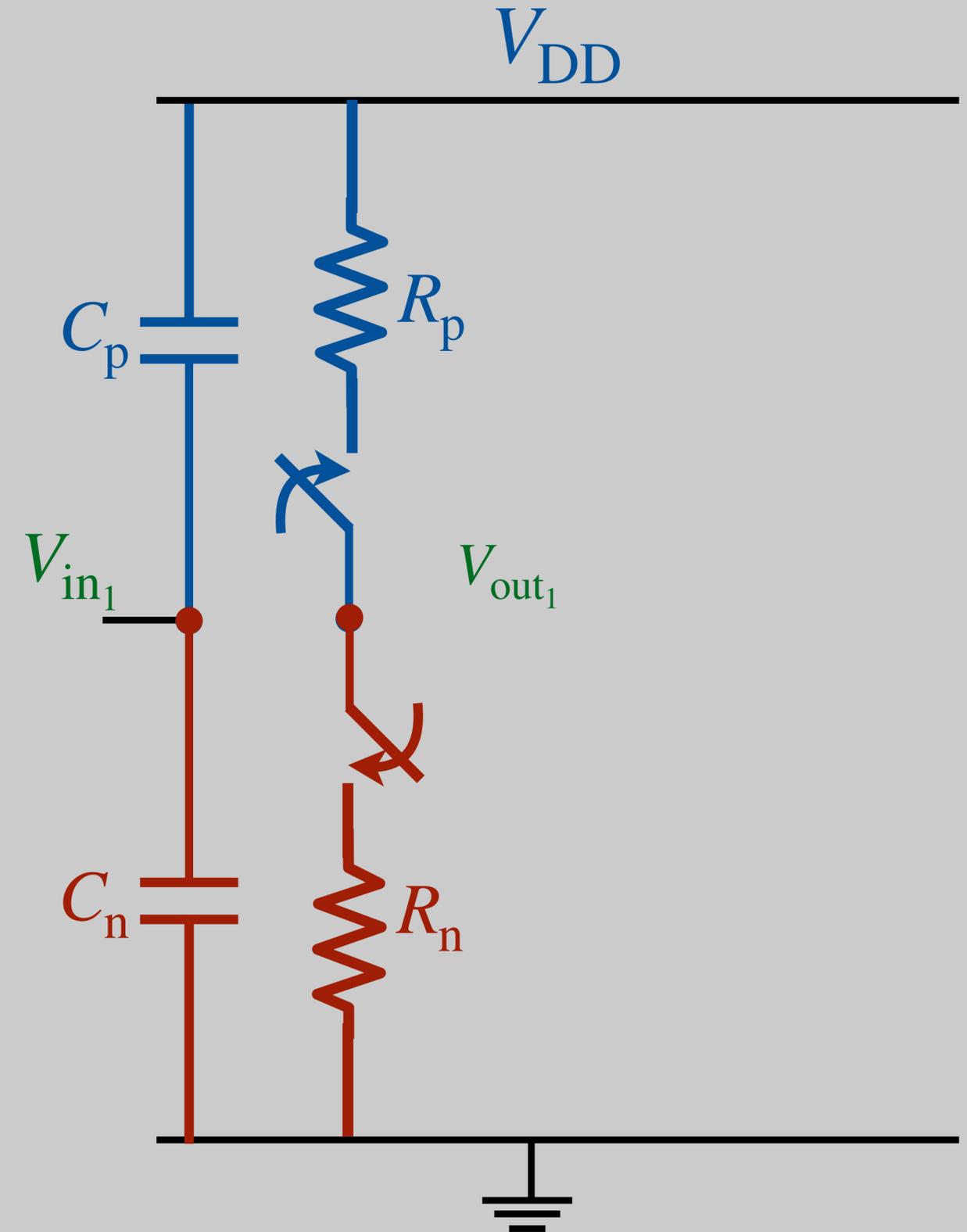
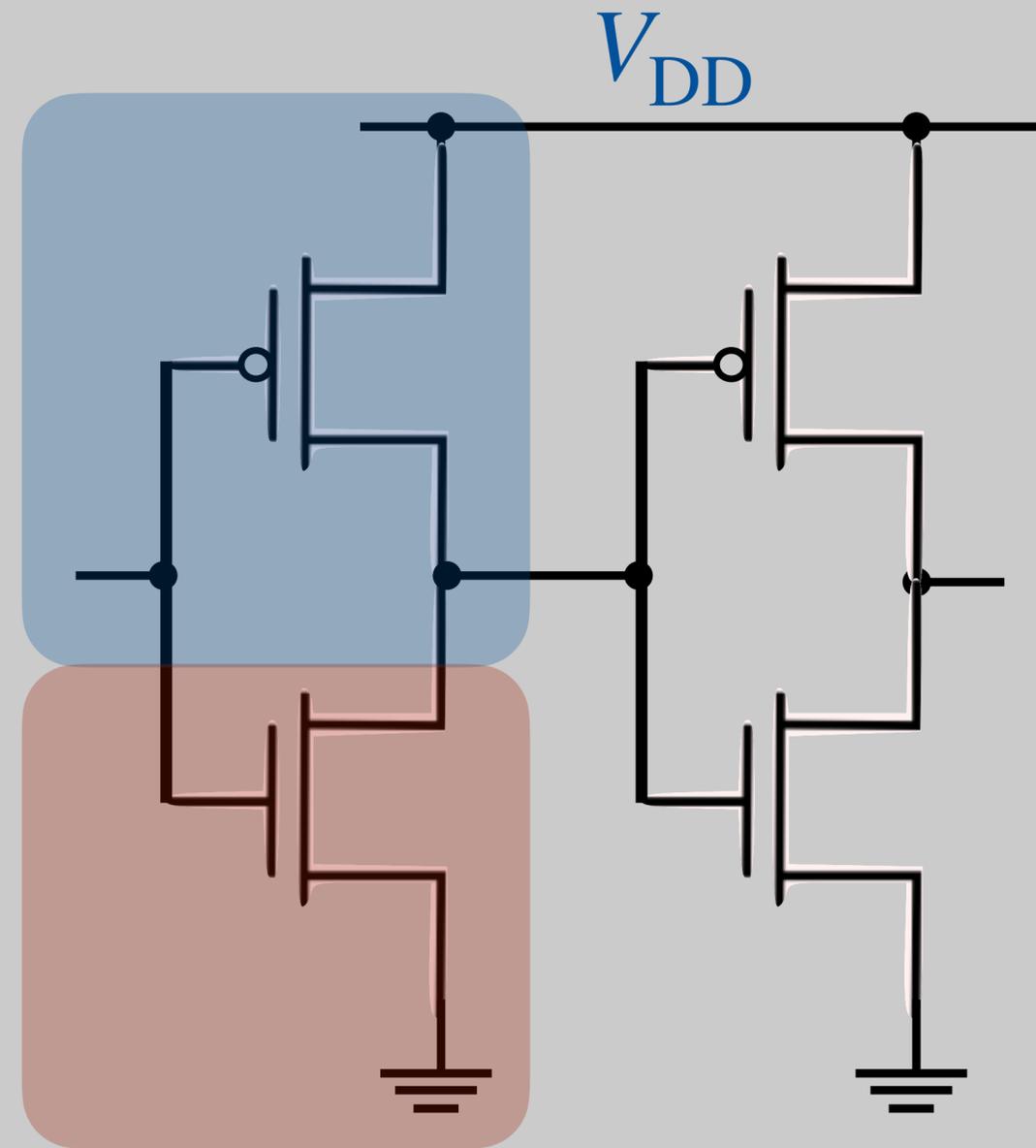
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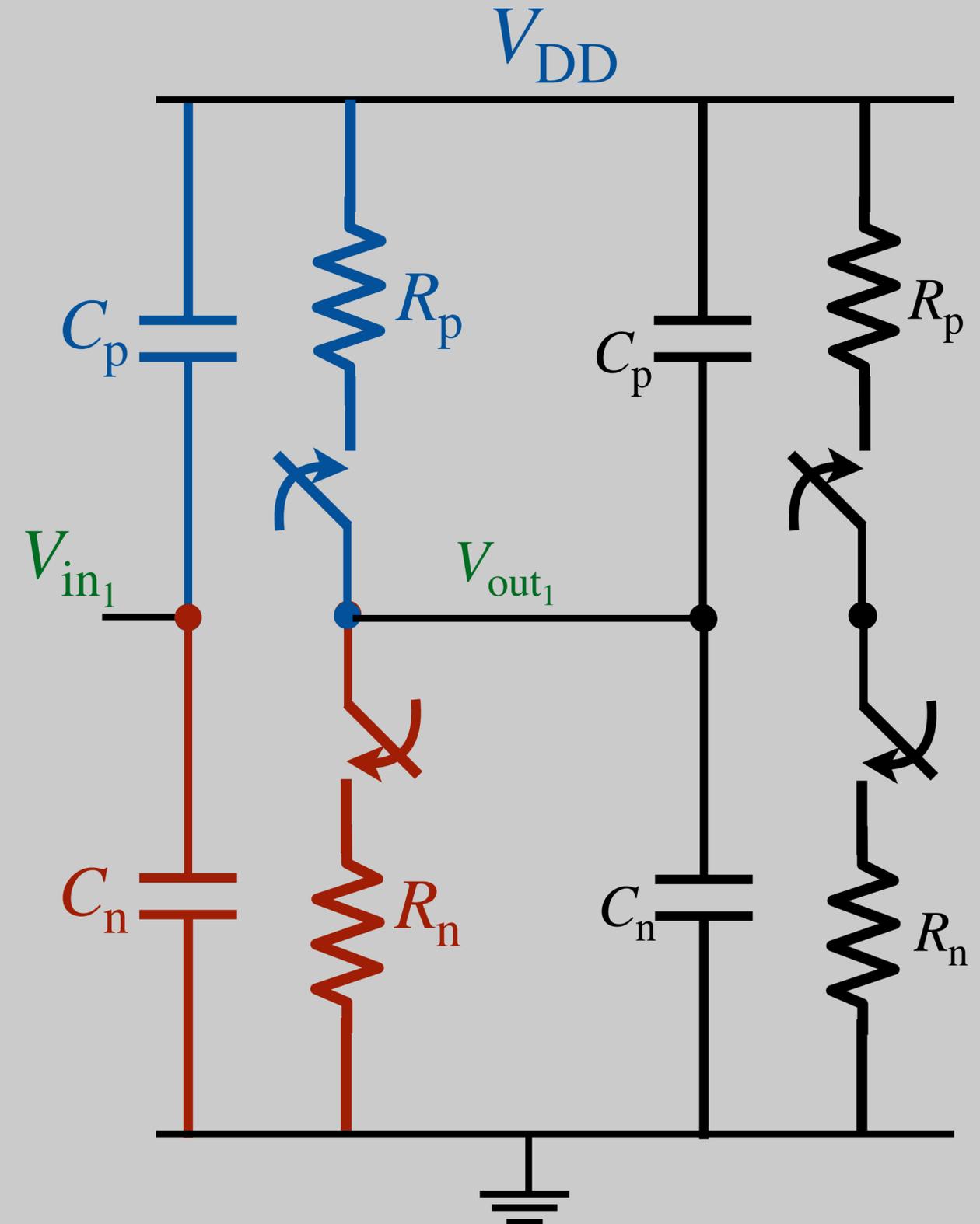
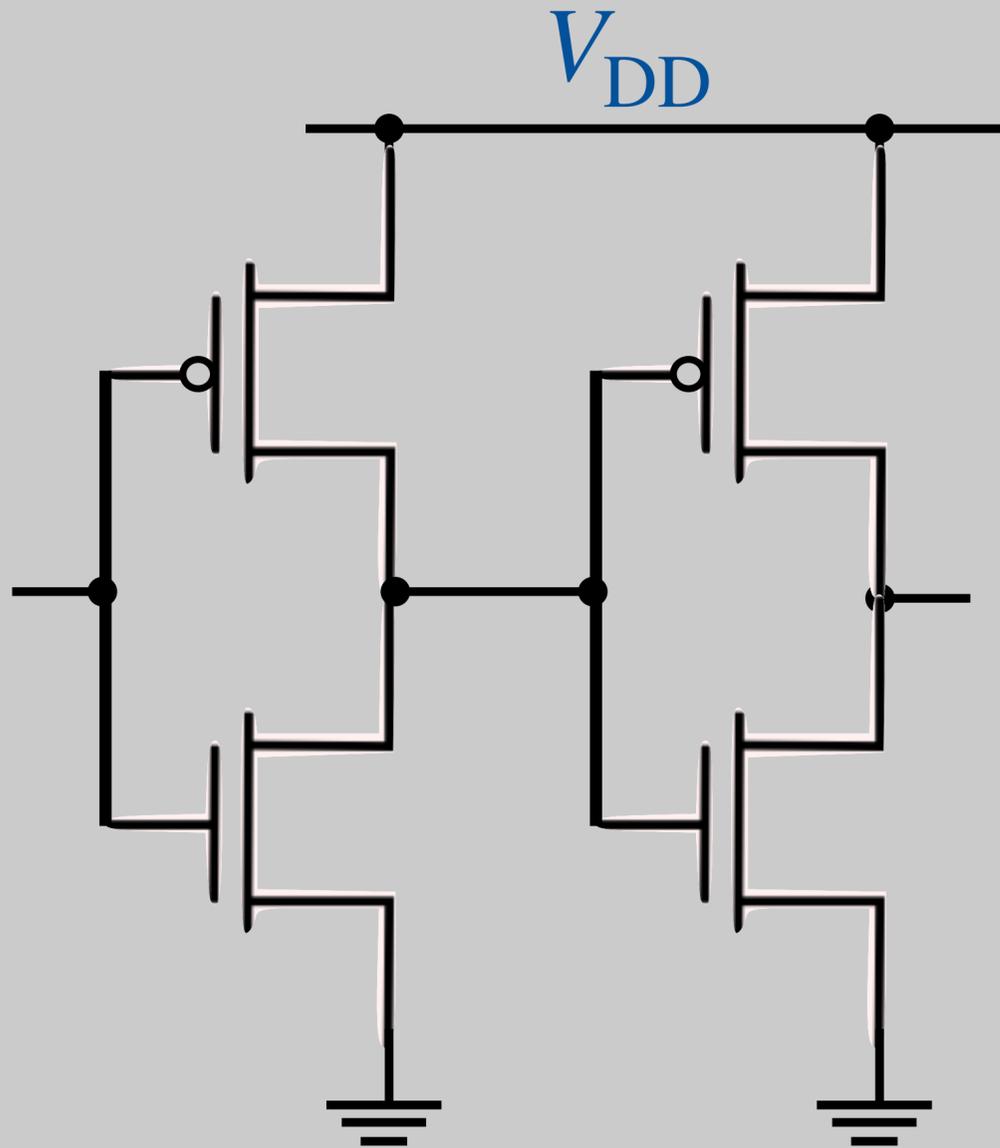
# RC Model of CMOS



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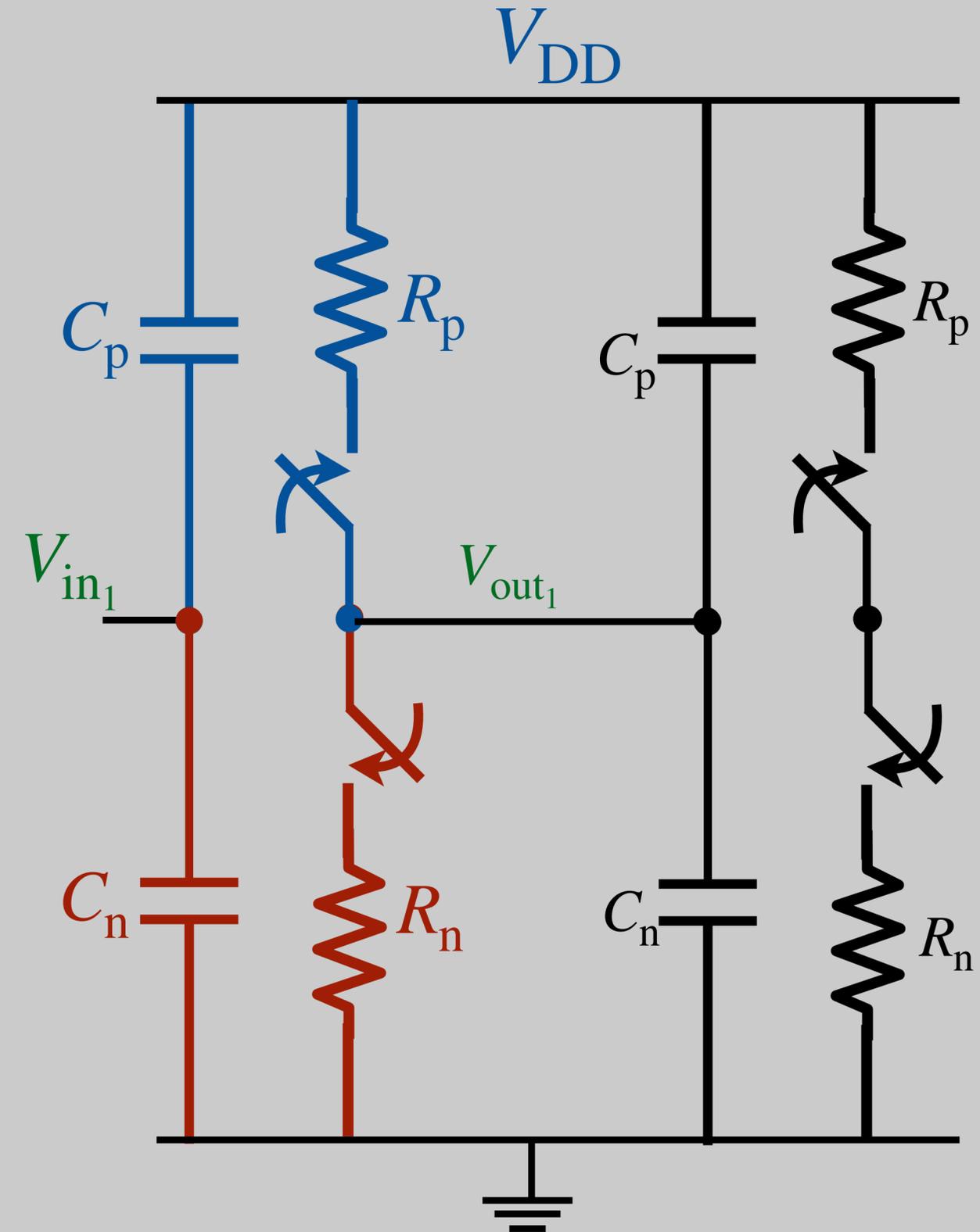
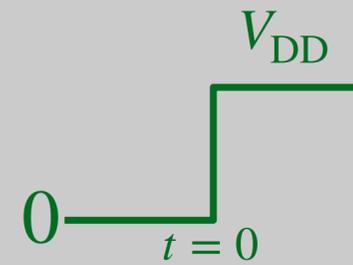
# RC Model of CMOS



# RC Model of CMOS

Example 1: switches from  $V_{in_1} = 0 \Rightarrow V_{in_1} = V_{DD}$

$t < 0$

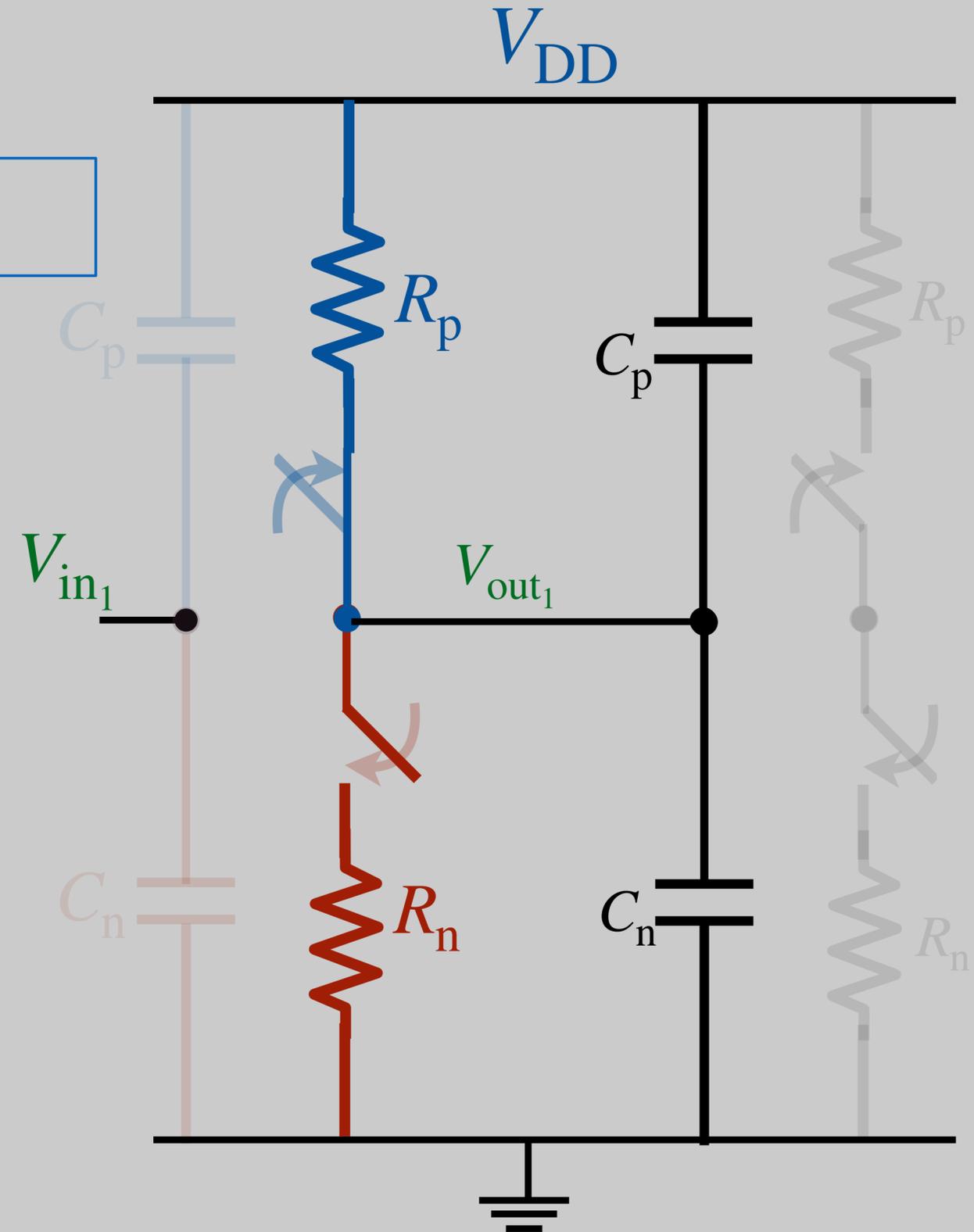
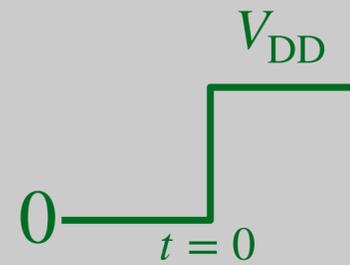


# RC Model of CMOS

Example 1: switches from  $V_{in_1} = 0 \Rightarrow V_{in_1} = V_{DD}$

$$t < 0 \Rightarrow V_{in_1} = \quad , V_{out_1} = \quad , V_{C_p} = \quad , V_{C_n} = \quad$$

$$t \geq 0$$

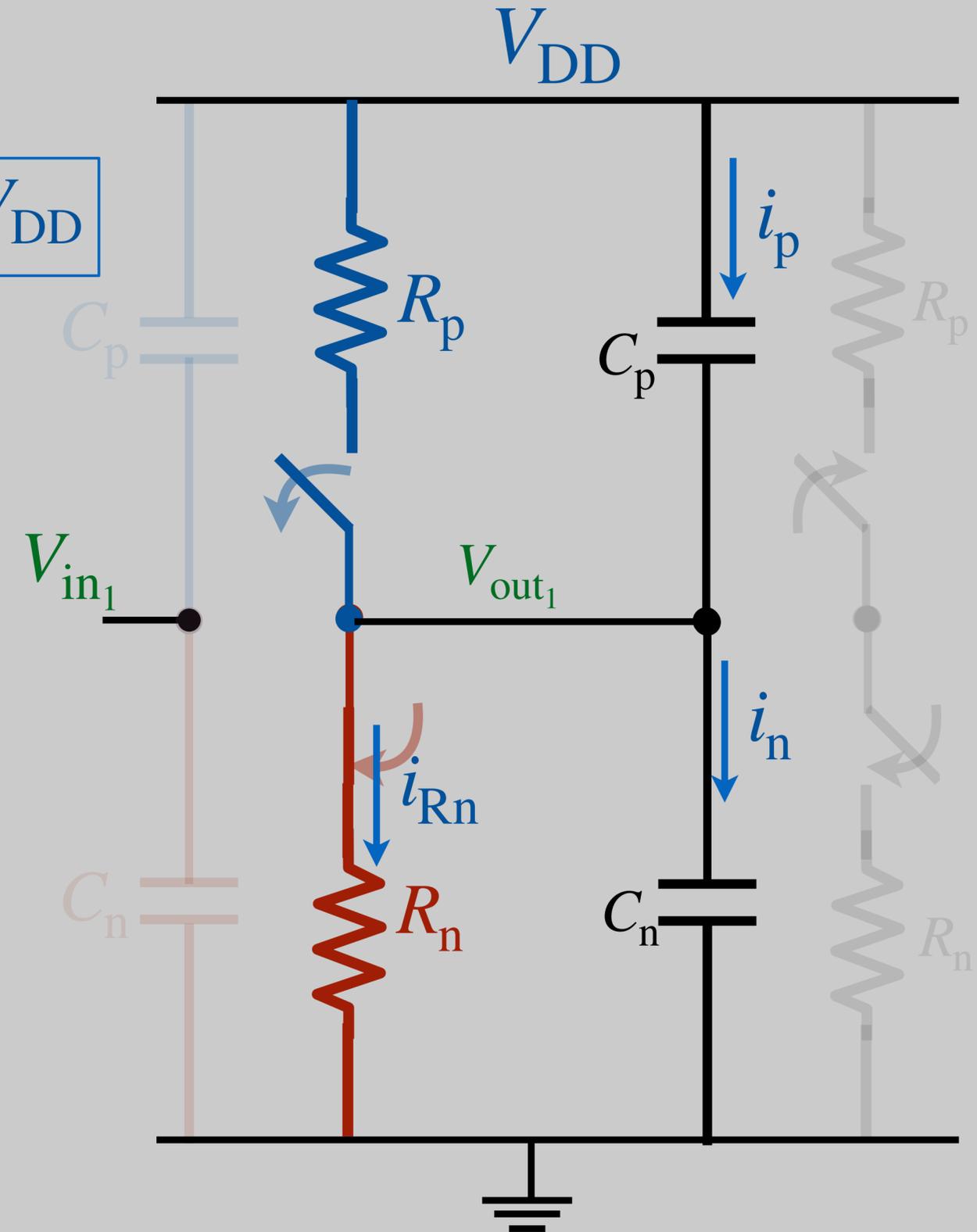
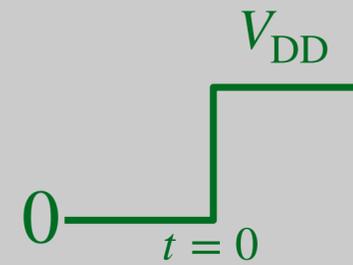


# RC Model of CMOS

Example 1: switches from  $V_{in_1} = 0 \Rightarrow V_{in_1} = V_{DD}$

$$t < 0 \Rightarrow V_{in_1} = 0, V_{out_1} = V_{DD}, V_{C_p} = 0, V_{C_n} = V_{DD}$$

$$t \geq 0$$



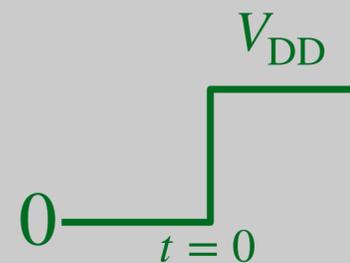
# RC Model of CMOS

Example 1: switches from  $V_{in_1} = 0 \Rightarrow V_{in_1} = V_{DD}$

$$t < 0 \Rightarrow V_{in_1} = 0, V_{out_1} = V_{DD}, V_{C_p} = 0, V_{C_n} = V_{DD}$$

$$t \geq 0 \quad i_{R_n} + i_n = i_p$$

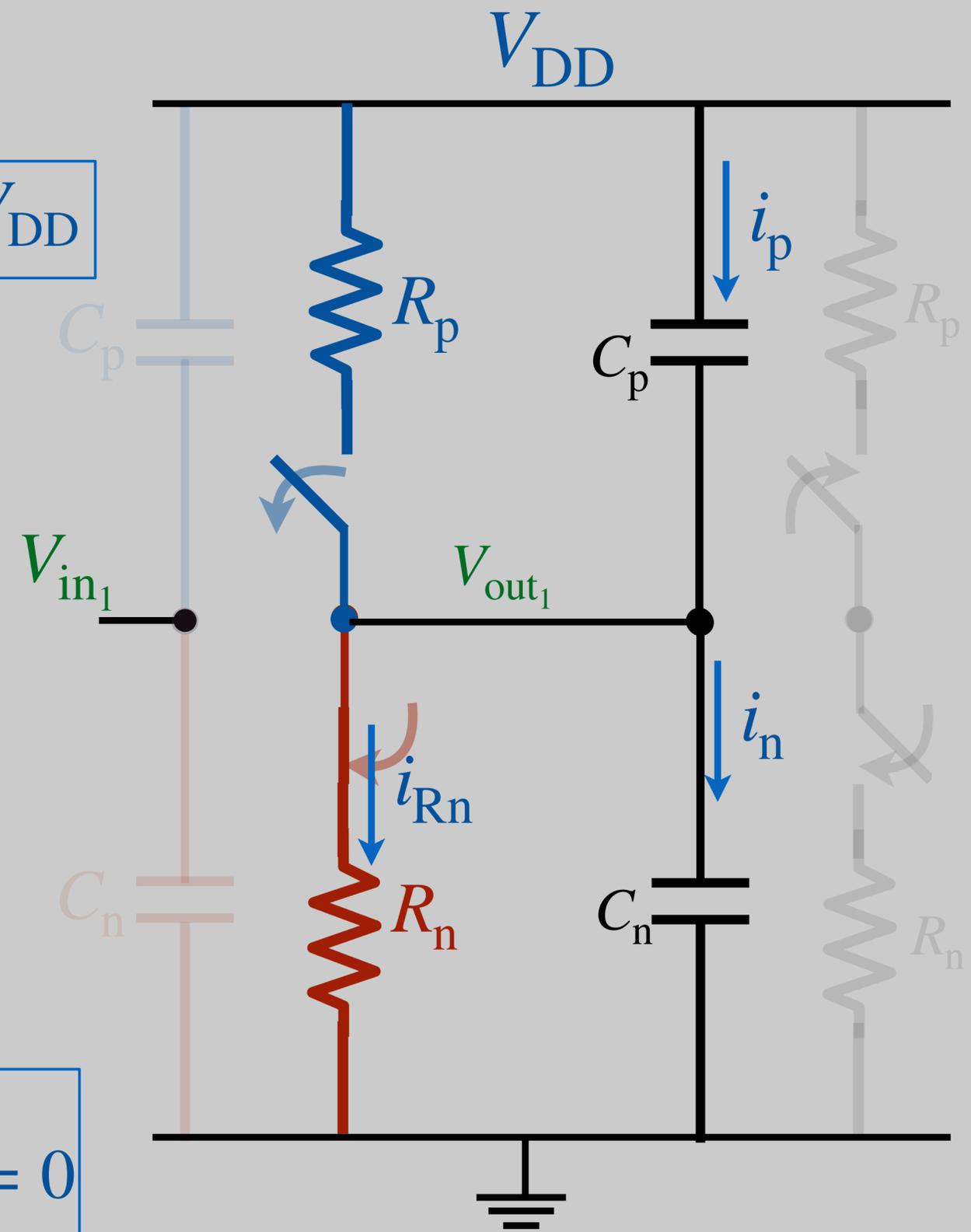
$$\frac{V_{out_1}}{R_n} + C_n \frac{dV_{out_1}}{dt} = C_p \frac{d(\cancel{V_{DD}} - V_{out_1})}{dt}$$



$$C_n \frac{dV_{out_1}}{dt} + C_p \frac{dV_{out_1}}{dt} + \frac{V_{out_1}}{R_n} = 0$$

$$(C_n + C_p) \frac{dV_{out_1}}{dt} + \frac{V_{out_1}}{R_n} = 0$$

$$\frac{dV_{out_1}}{dt} + \frac{V_{out_1}}{R_n(C_n + C_p)} = 0$$



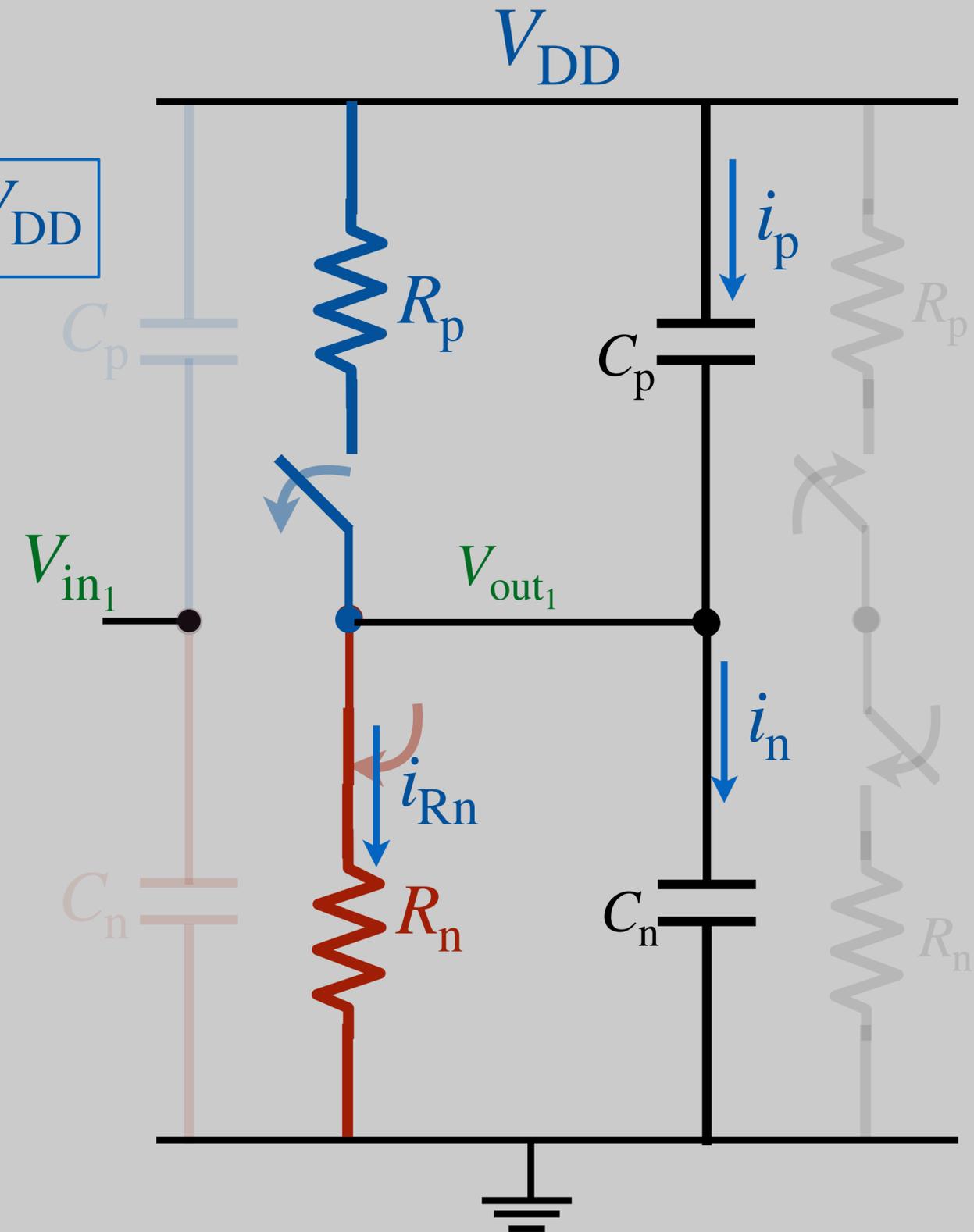
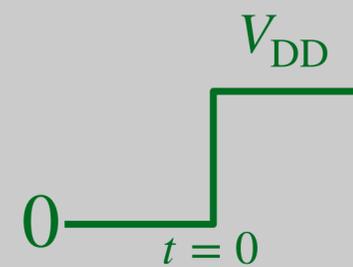
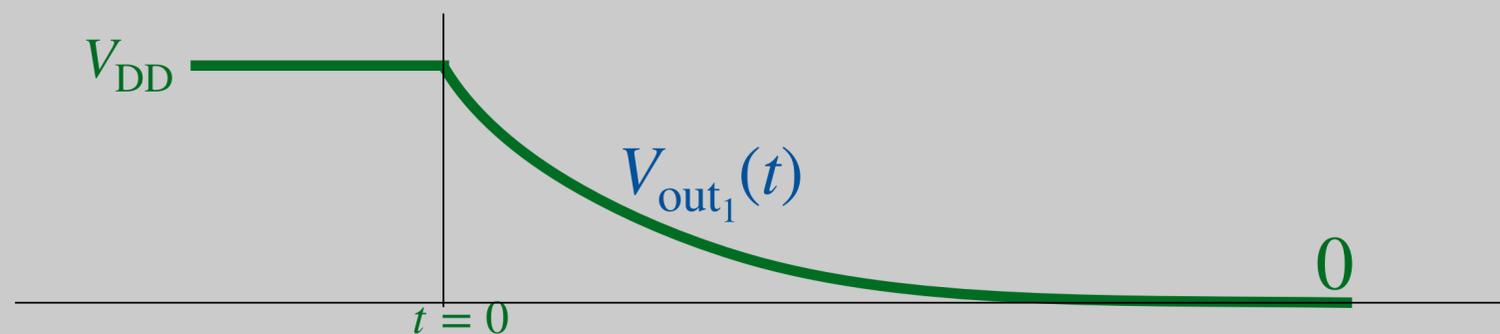
# RC Model of CMOS

Example 1: switches from  $V_{in_1} = 0 \Rightarrow V_{in_1} = V_{DD}$

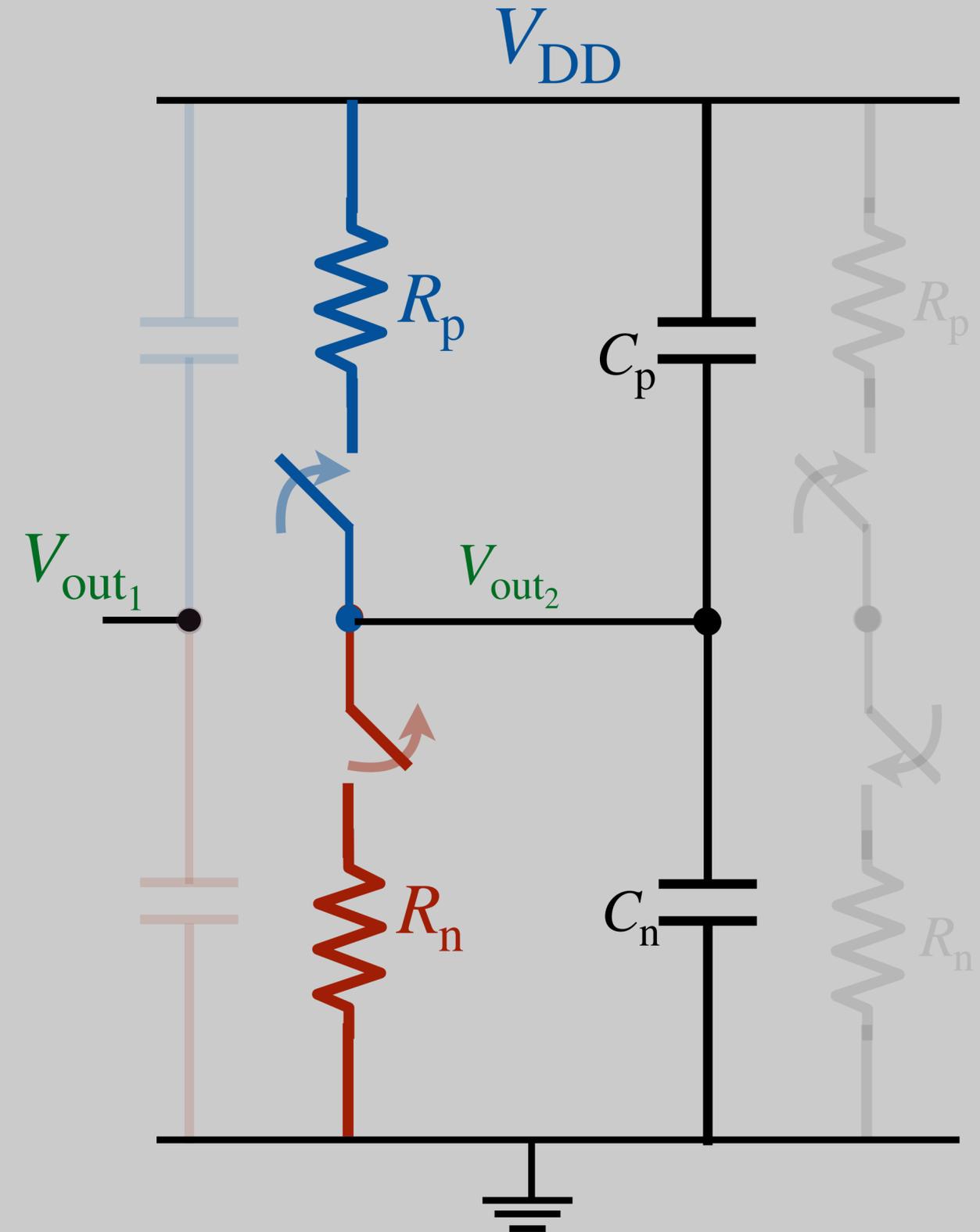
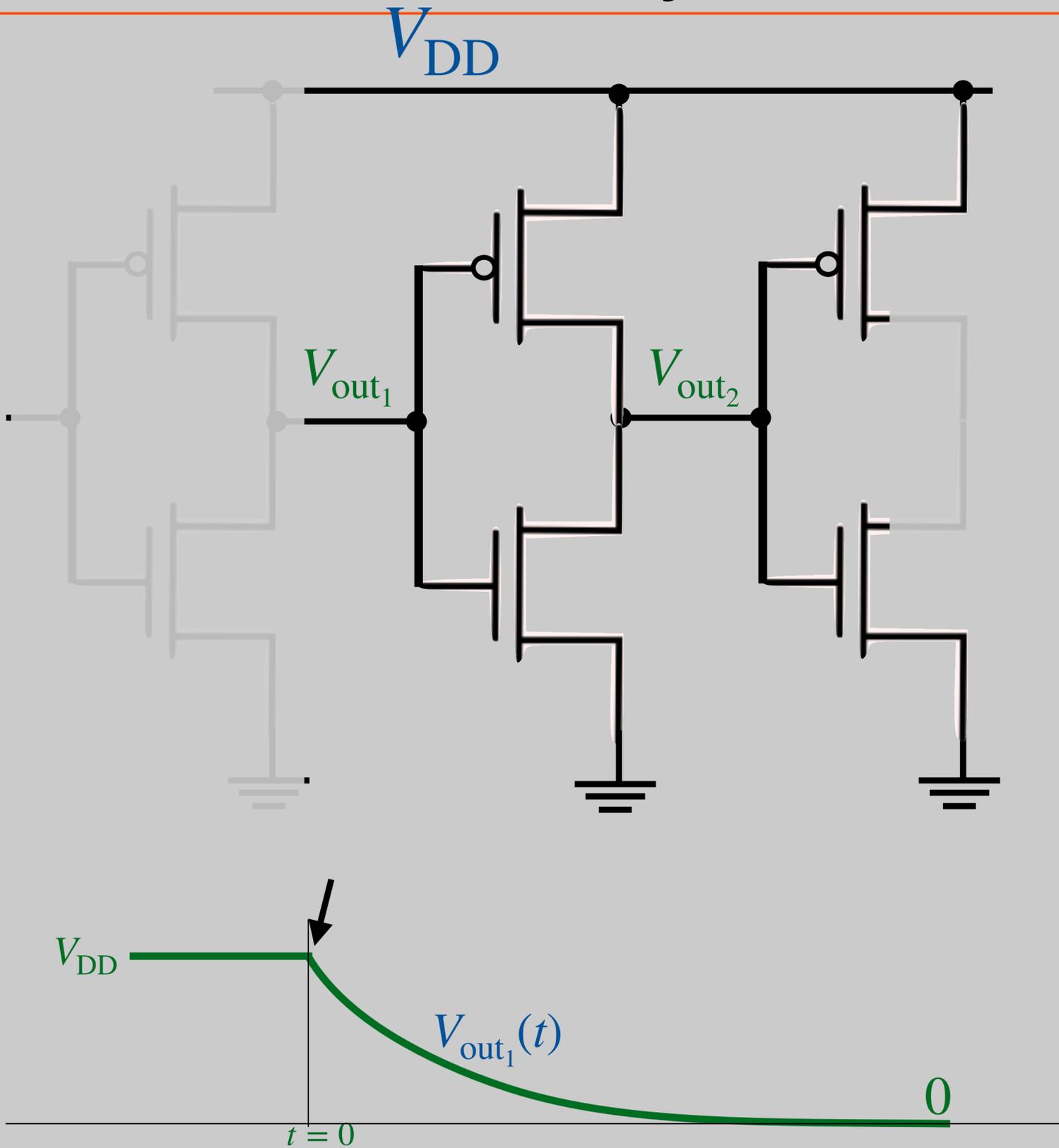
$$t < 0 \Rightarrow V_{in_1} = 0, V_{out_1} = V_{DD}, V_{C_p} = 0, V_{C_n} = V_{DD}$$

$$t \geq 0 \Rightarrow \frac{dV_{out_1}}{dt} + \frac{V_{out_1}}{R_n(C_n + C_p)} = 0$$

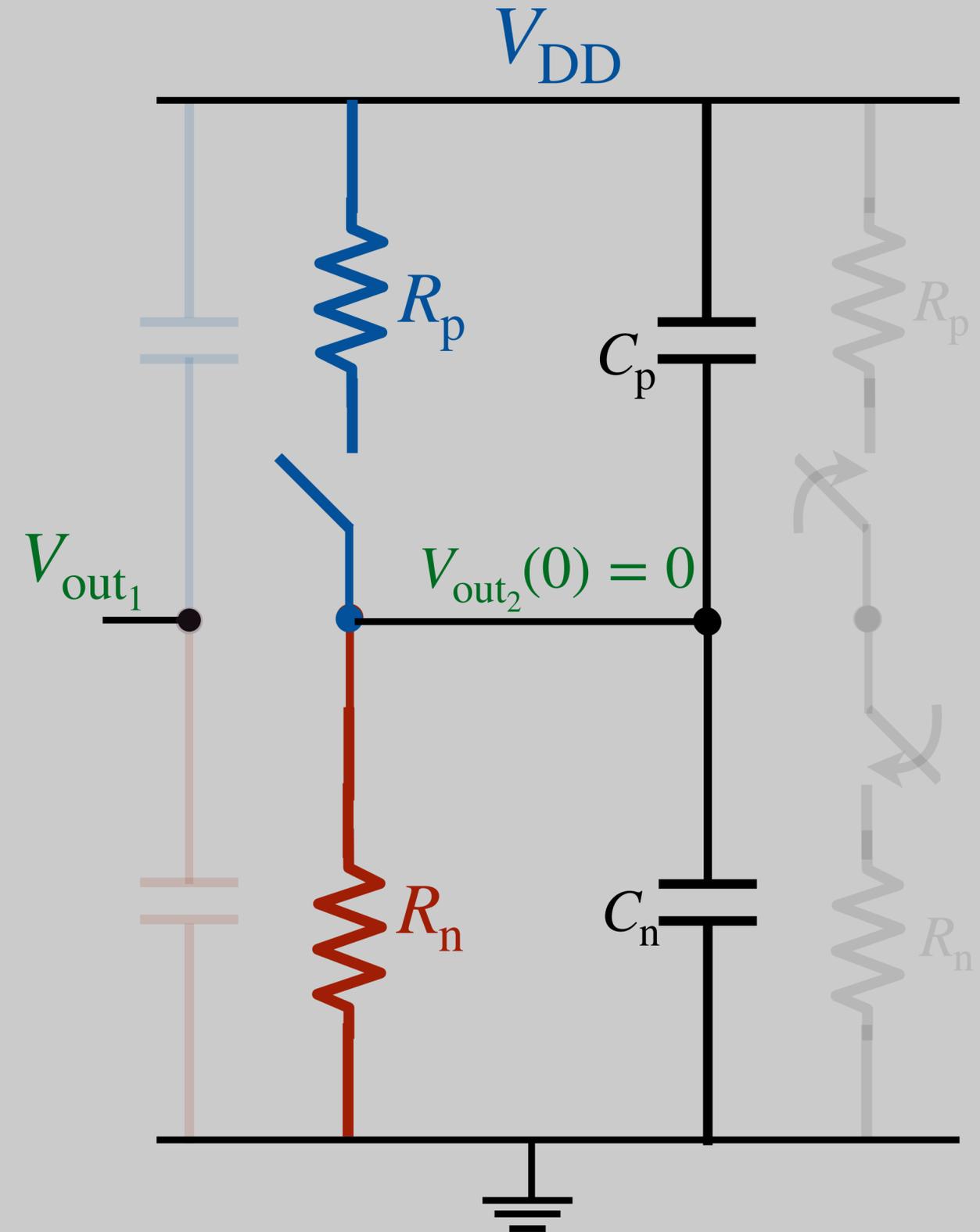
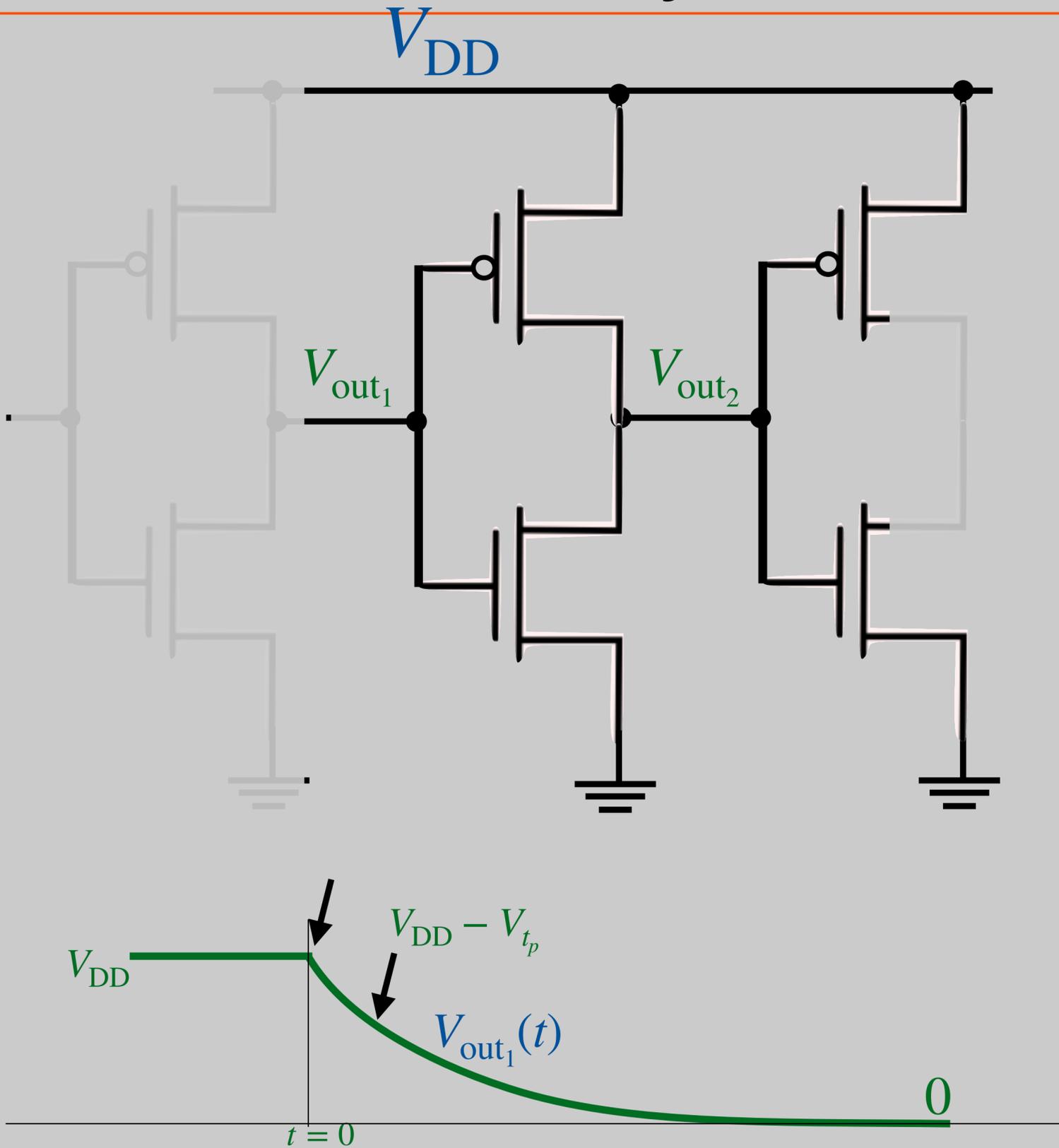
$$V_{out_1} = V_{DD} e^{-\frac{t}{R_n(C_p + C_n)}}$$



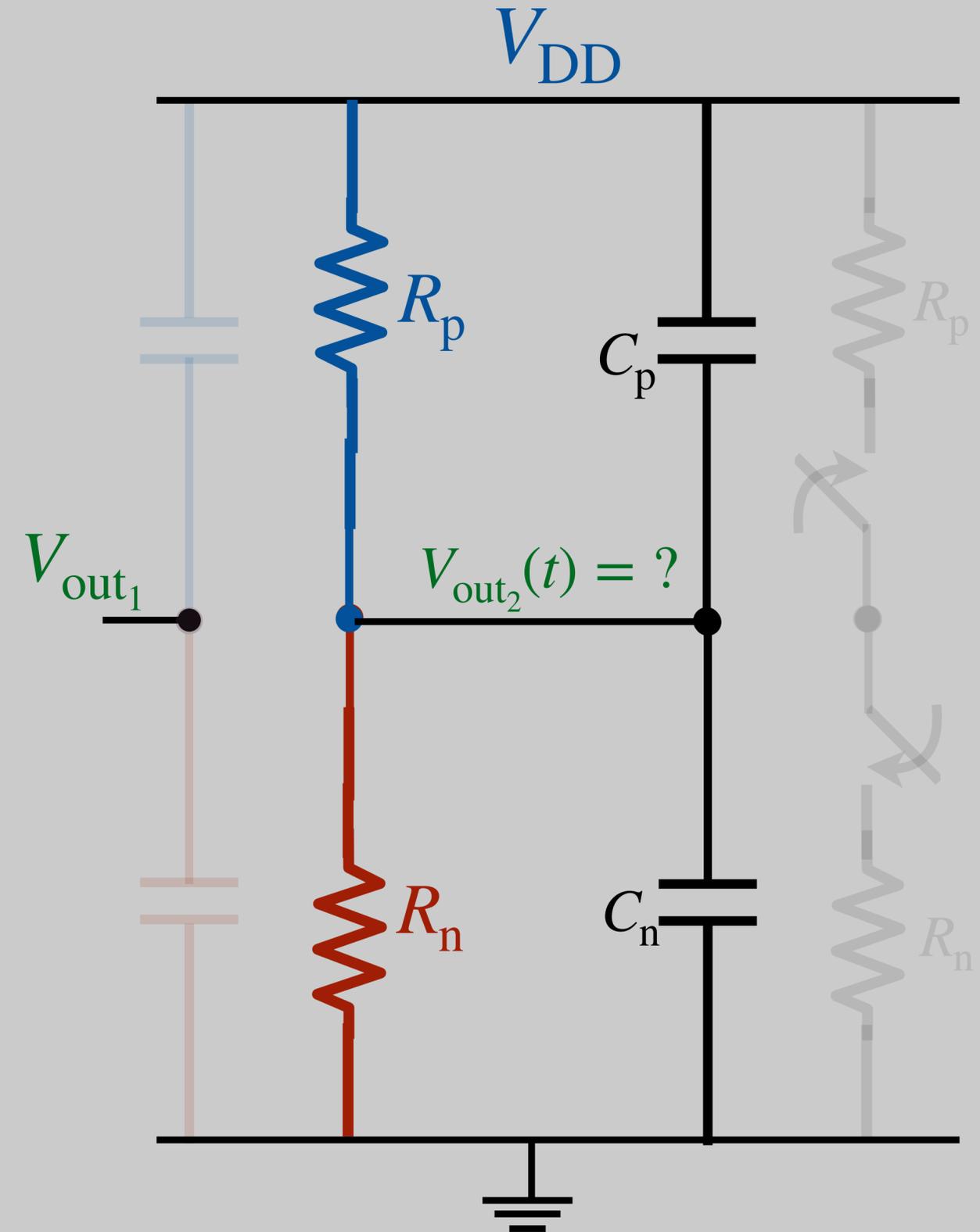
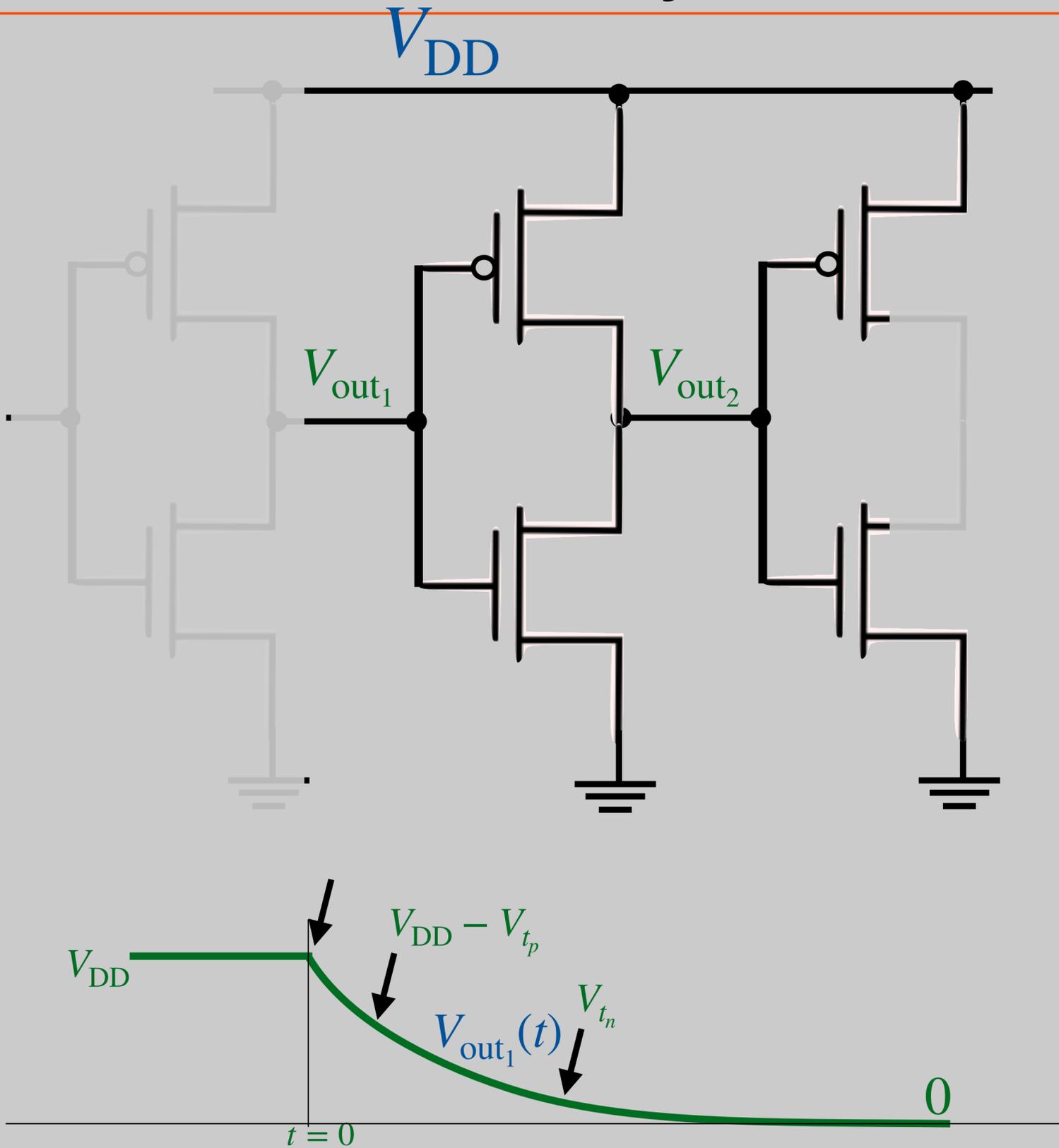
# Realistic RC Delay



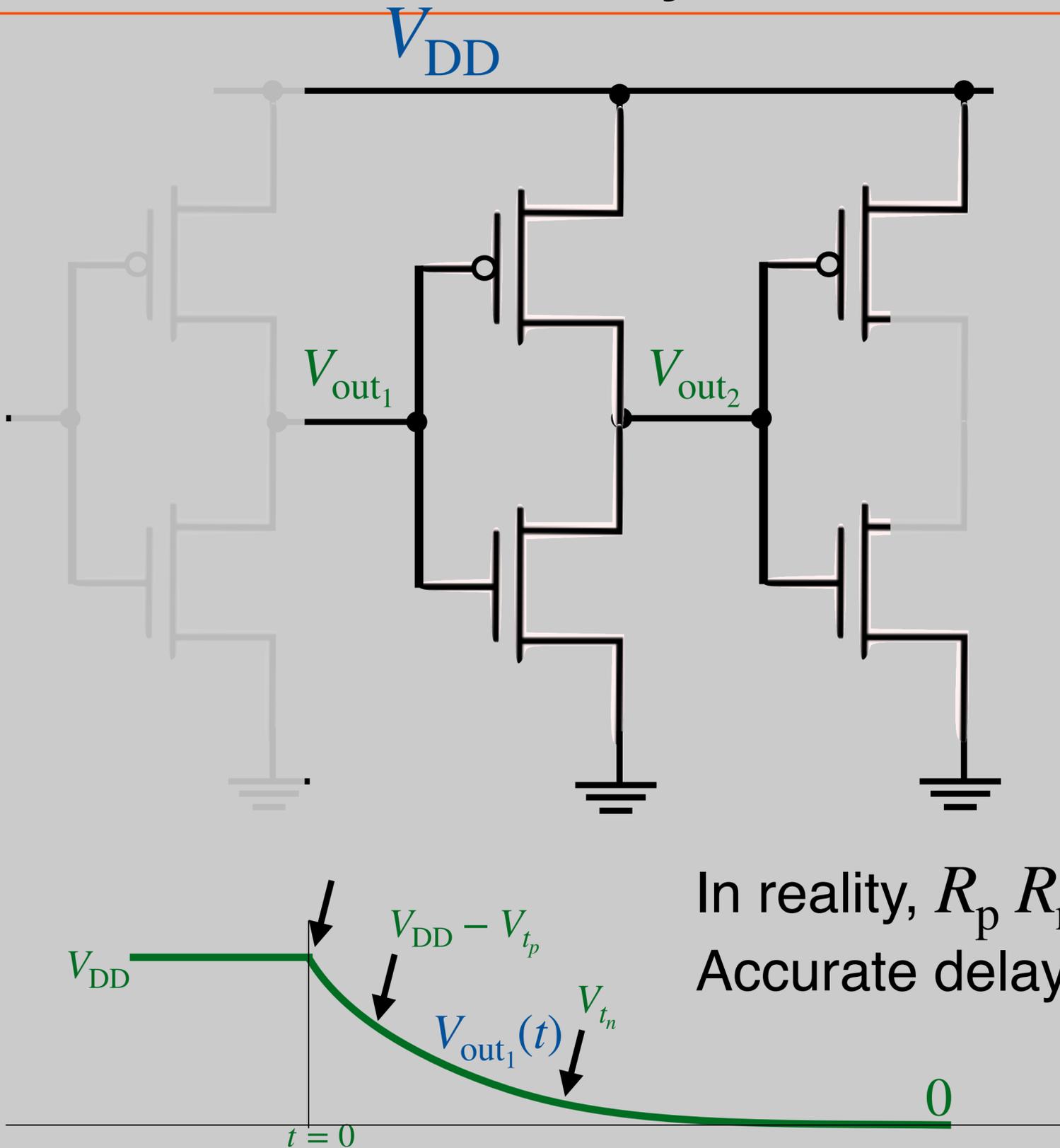
# Realistic RC Delay



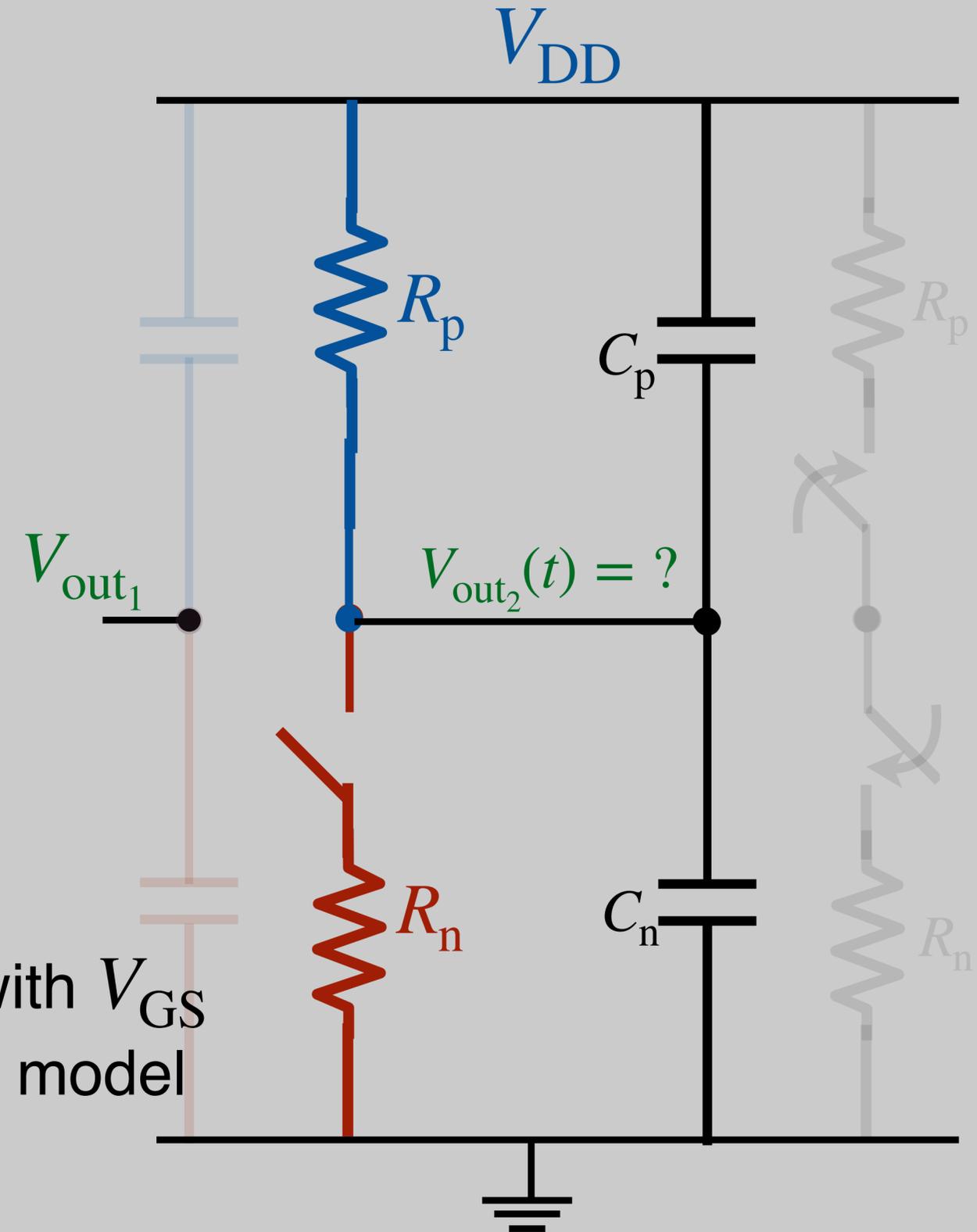
# Realistic RC Delay



# Realistic RC Delay

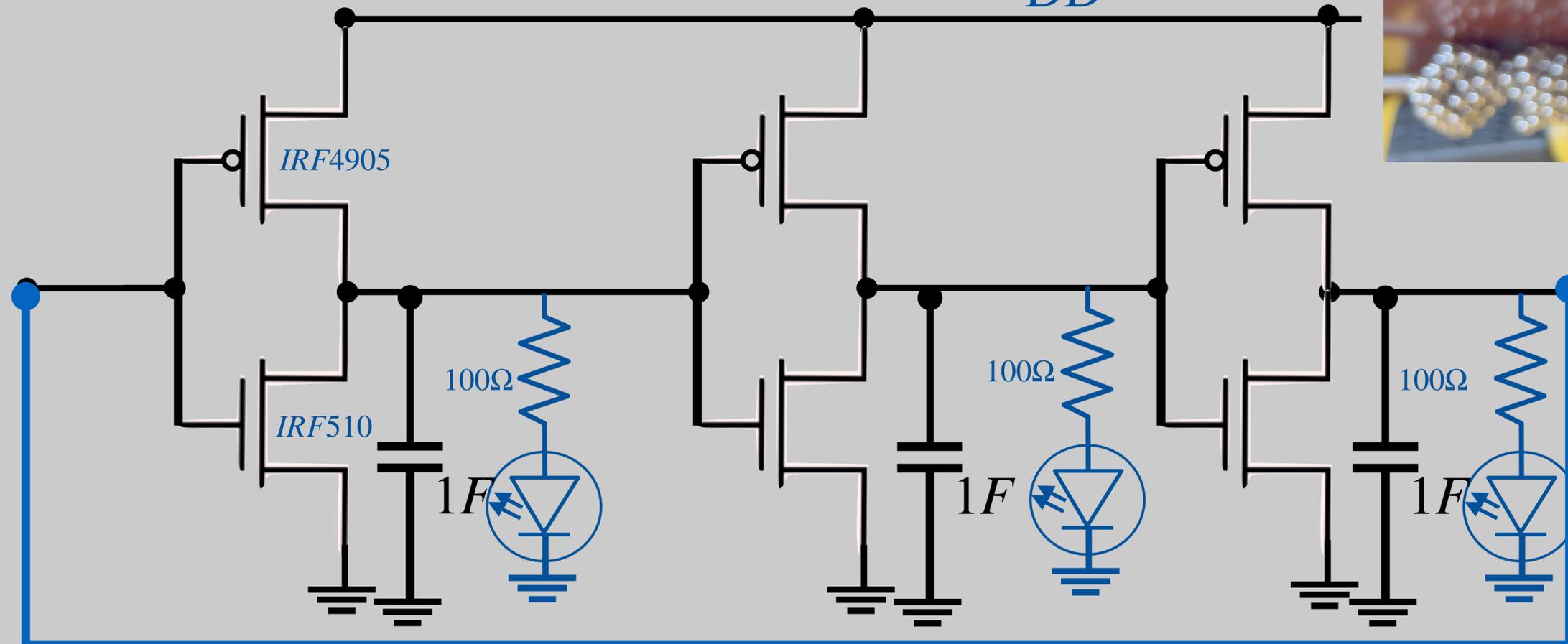
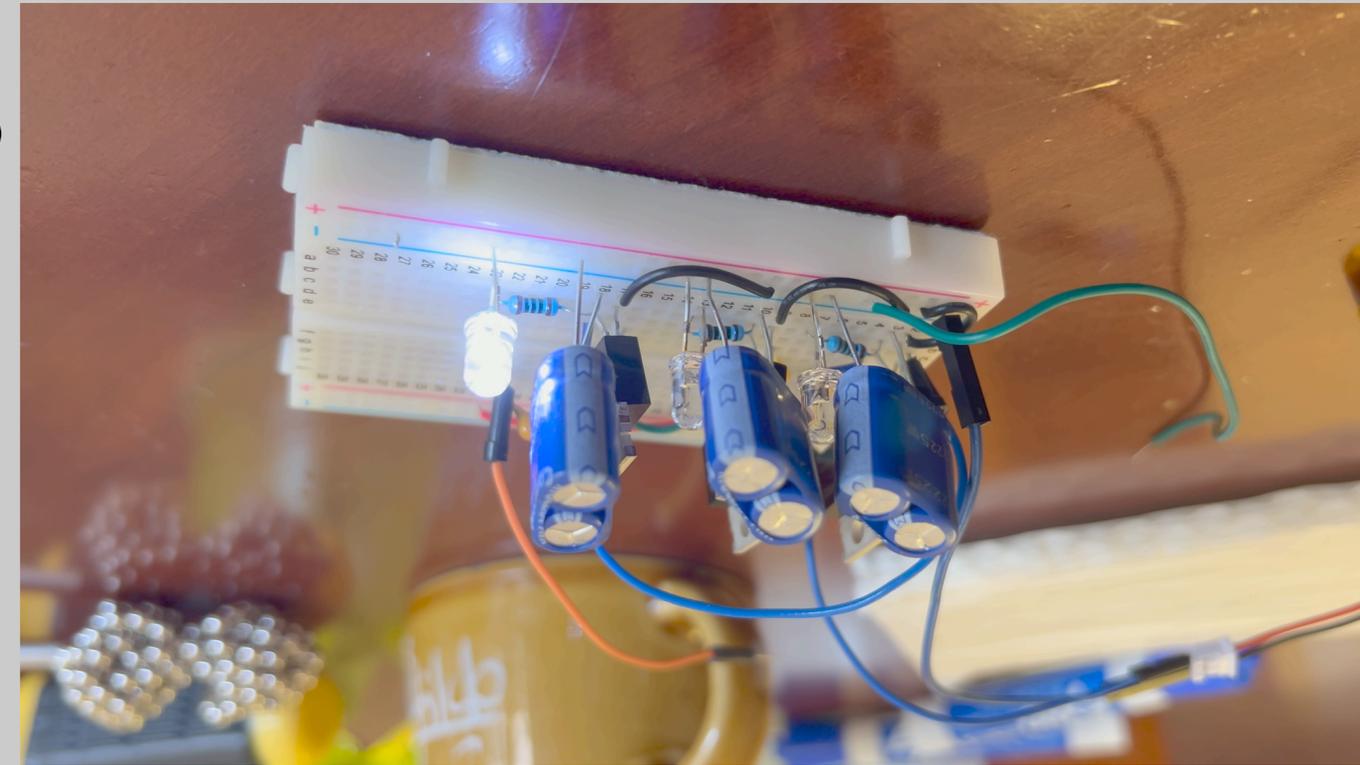
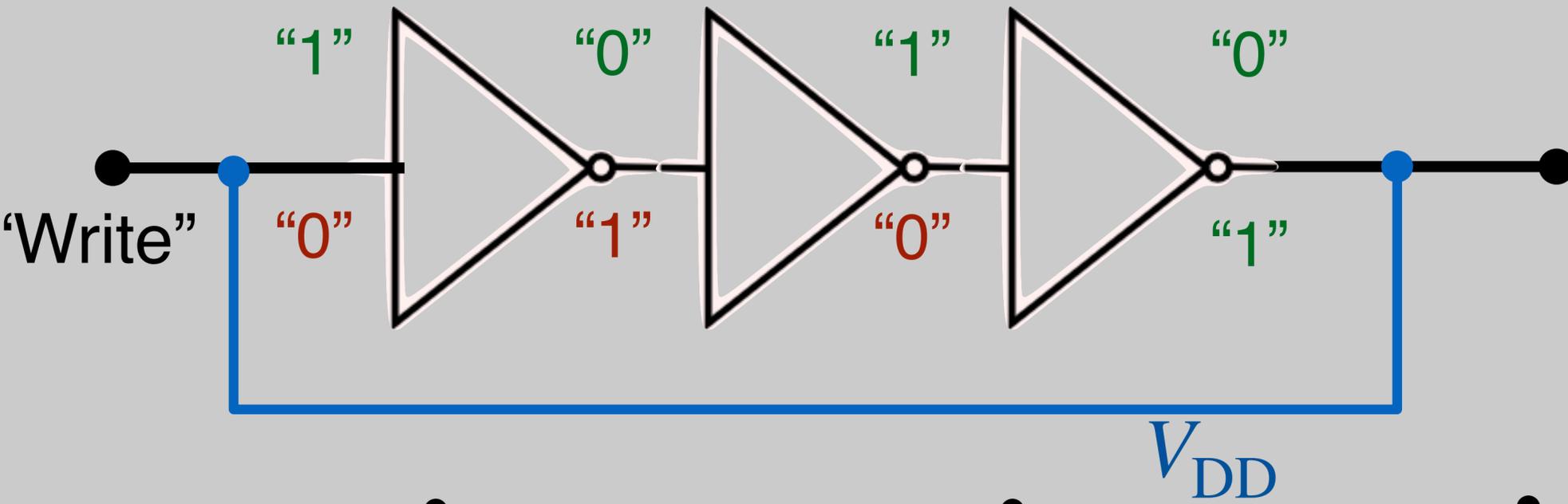


In reality,  $R_p$   $R_n$  also change with  $V_{GS}$   
Accurate delay needs a better model



# Demo

Inconsistent!



# Complex Numbers Review

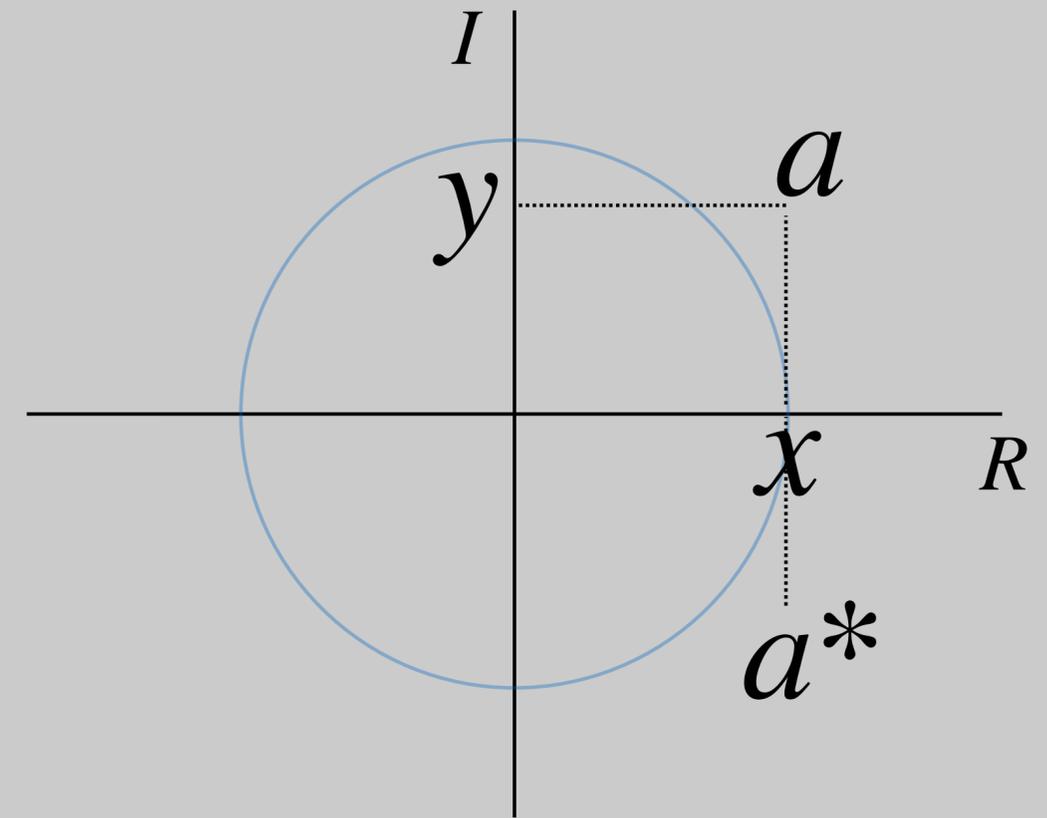
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$$a = x + jy \quad |j = \sqrt{-1}$$

$$a^* = x - jy$$

$$\mathcal{R}e\{a\} = \frac{1}{2}(a + a^*) = x$$

$$\mathcal{I}m\{a\} = \frac{1}{2j}(a - a^*) = y$$



# Euler Formula

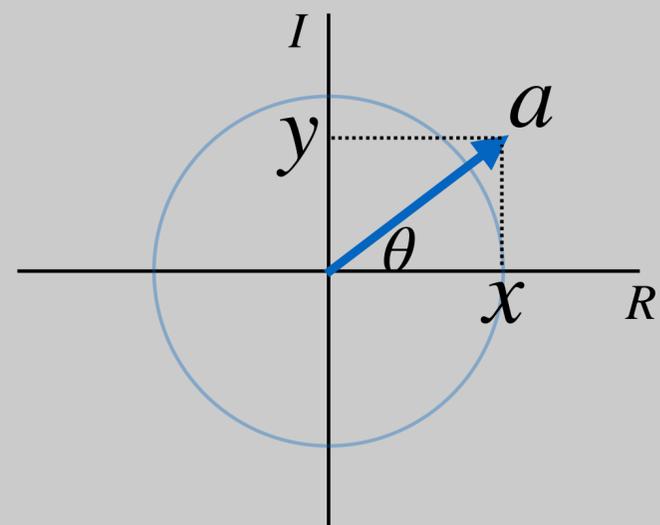
$$a = x + jy$$

$$aa^* = |a|^2 = |x|^2 + |y|^2$$

$$\theta = \angle(a) = \text{atan}\left(\frac{y}{x}\right)$$

Euler:

$$\begin{aligned} a &= |a| e^{j\theta} \\ &= |a| (\cos(\theta) + j \sin(\theta)) \end{aligned}$$



NUMBERS OF THE FORM  
 $n\sqrt{-1}$  ARE "IMAGINARY,"  
BUT CAN STILL BE USED  
IN EQUATIONS.

OKAY.  
AND  $e^{\pi\sqrt{-1}} = -1.$

NOW YOU'RE JUST  
FUCKING WITH ME.



# Euler Formula

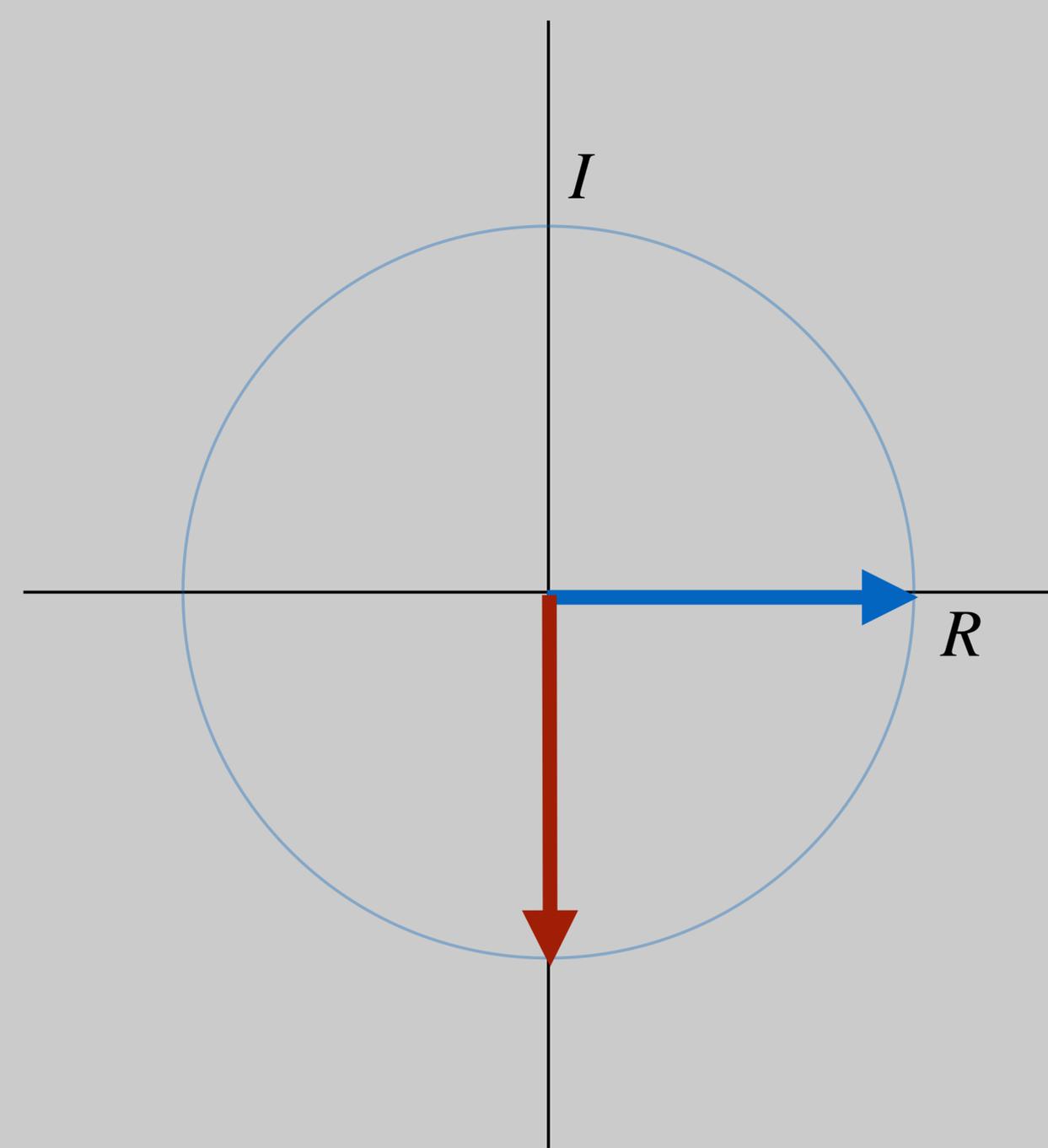
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$$\begin{aligned}\text{Euler: } a &= |a| e^{j\theta} \\ &= |a| (\cos(\theta) + j \sin(\theta))\end{aligned}$$

$$|e^{j\theta}| = \cos^2(\theta) + \sin^2(\theta) = 1$$

$$\mathcal{Re}\{e^{j\omega t}\} = \cos(\omega t)$$

$$\mathcal{Re}\{e^{j(\omega t - \pi/2)}\} = \sin(\omega t)$$



# Euler Formula

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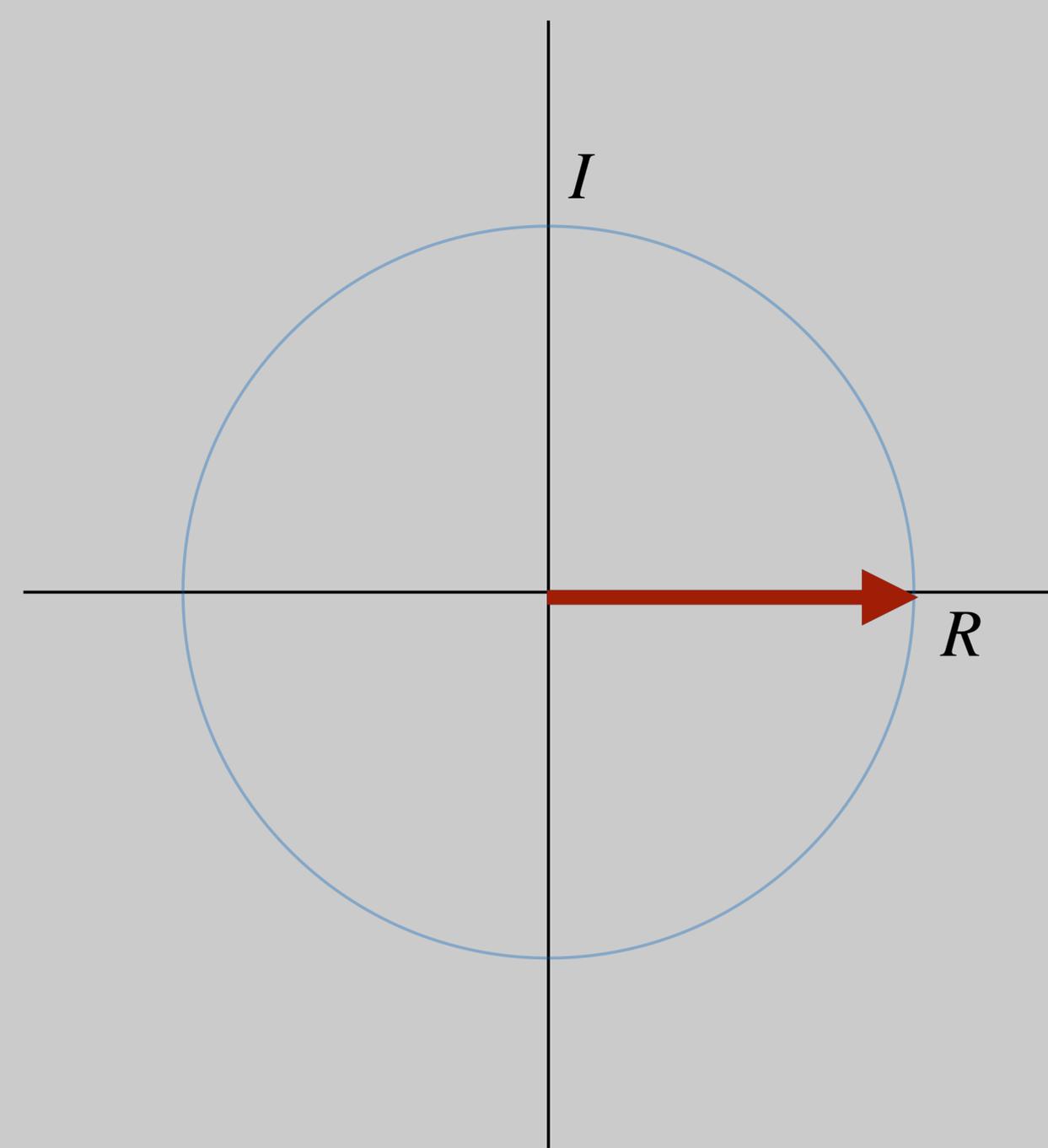
$$\begin{aligned}\text{Euler: } a &= |a| e^{j\theta} \\ &= |a| (\cos(\theta) + j \sin(\theta))\end{aligned}$$

$$|e^{j\theta}| = \cos^2(\theta) + \sin^2(\theta) = 1$$

$$\mathcal{R}e\{e^{j\omega t}\} = \cos(\omega t)$$

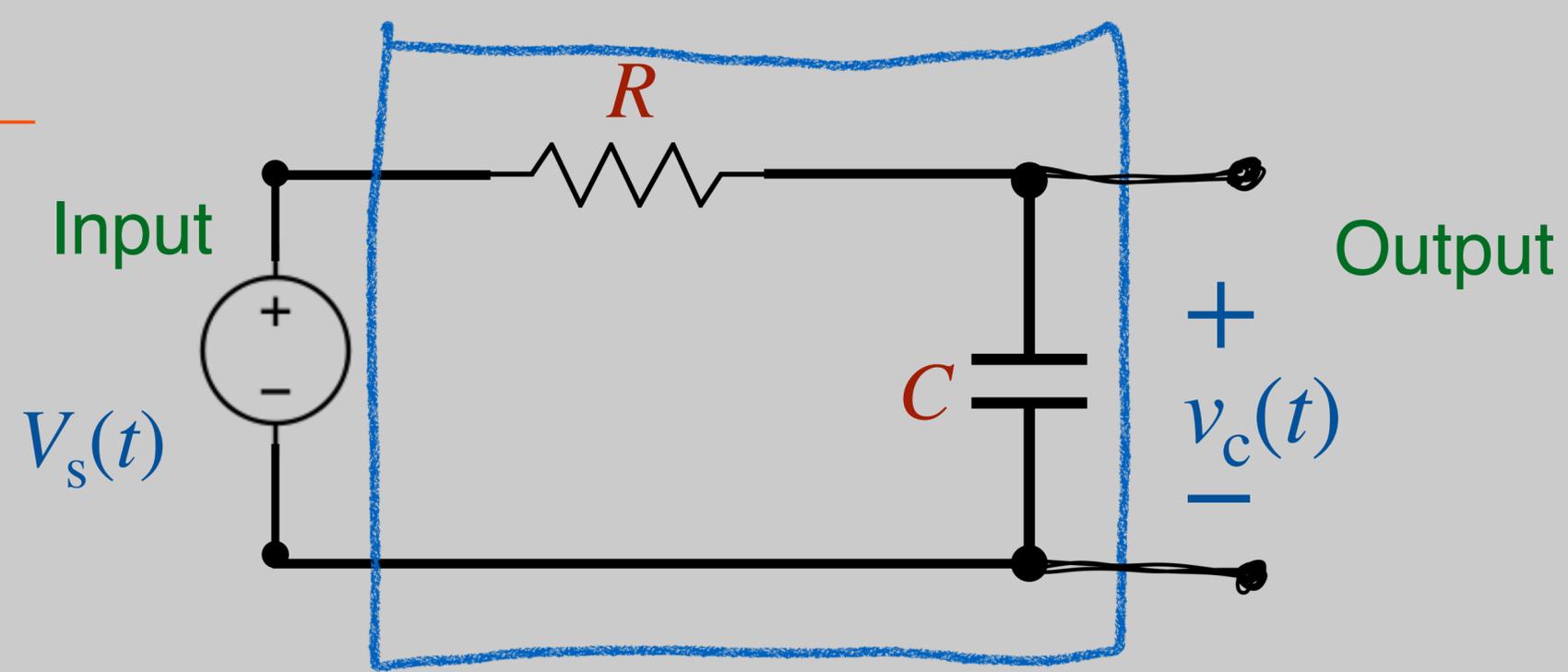
$$\mathcal{R}e\{e^{j(\omega t - \pi/2)}\} = \sin(\omega t)$$

$$\frac{1}{2} (e^{j\omega t} + e^{-j\omega t}) = \cos(\omega t)$$

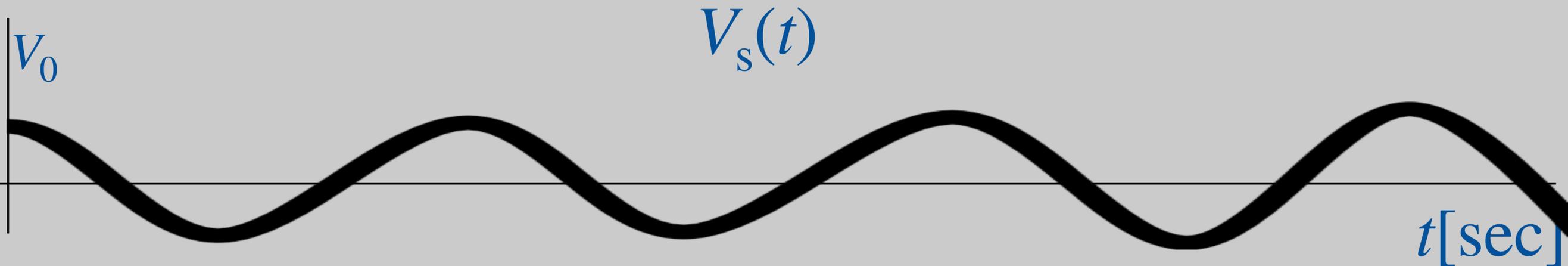


# Sinusoidal Response

$$\frac{dv_c(t)}{dt} + \frac{1}{RC}v_c(t) = \frac{1}{RC}V_s(t)$$



## Example 5: Sinusoidal Input



$$V_s(t) = V_0 \cos(\omega t) \quad | t > 0$$

$$v_c(0) = 0$$

$$\Rightarrow \frac{dv_c(t)}{dt} + \frac{1}{RC}v_c(t) = \frac{1}{RC}V_0 \cos(\omega t)$$

# Sinusoidal Response

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General Solution:

$$v_c(t) = Ke^{-\frac{t}{RC}} + \frac{e^{-\frac{t}{RC}}}{RC} \int_{-\infty}^t V_s(\tau) e^{\frac{\tau}{RC}} d\tau$$

$$V_s(t) = V_0 \cos(\omega t) \quad | t > 0$$

$$v_c(0) = 0$$

Approach:

Set  $V_s(t) = V_0 e^{j\omega t} \quad | t > 0$

Solve for  $v_c(t)$  as usual

Solution for  $V_s(t) = V_0 \cos(\omega t)$  is  $\mathcal{Re}\{v_c(t)\}$

# Sinusoidal Response

General Solution:

$$v_c(t) = Ke^{-\frac{t}{RC}} + \frac{e^{-\frac{t}{RC}}}{RC} \int_{-\infty}^t V_s(\tau) e^{\frac{\tau}{RC}} d\tau$$

$$V_s(t) = V_0 e^{j\omega t} \quad | t > 0$$

$$v_c(0) = 0$$

$$v_c(0) = 0 \Rightarrow K = 0$$

$$v_c(t) = \frac{e^{-\frac{t}{RC}}}{RC} \int_0^t V_0 e^{j\omega\tau} e^{\frac{\tau}{RC}} d\tau$$

$$= V_0 \frac{e^{-\frac{t}{RC}}}{RC} \int_0^t e^{(j\omega + \frac{1}{RC})\tau} d\tau$$

$$= V_0 \frac{e^{-\frac{t}{RC}}}{RC} \frac{1}{j\omega + \frac{1}{RC}} \left[ e^{(j\omega + \frac{1}{RC})\tau} \right] \Big|_0^t$$

$$= V_0 \frac{e^{-\frac{t}{RC}}}{j\omega RC + 1} \left( e^{(j\omega + \frac{1}{RC})t} - 1 \right)$$

$$= \frac{V_0}{j\omega RC + 1} \left( e^{j\omega t} - e^{-\frac{t}{RC}} \right)$$

# Sinusoidal Response

General Solution:

$$v_c(t) = \frac{V_0}{j\omega RC + 1} \left( e^{j\omega t} - e^{-\frac{t}{RC}} \right)$$

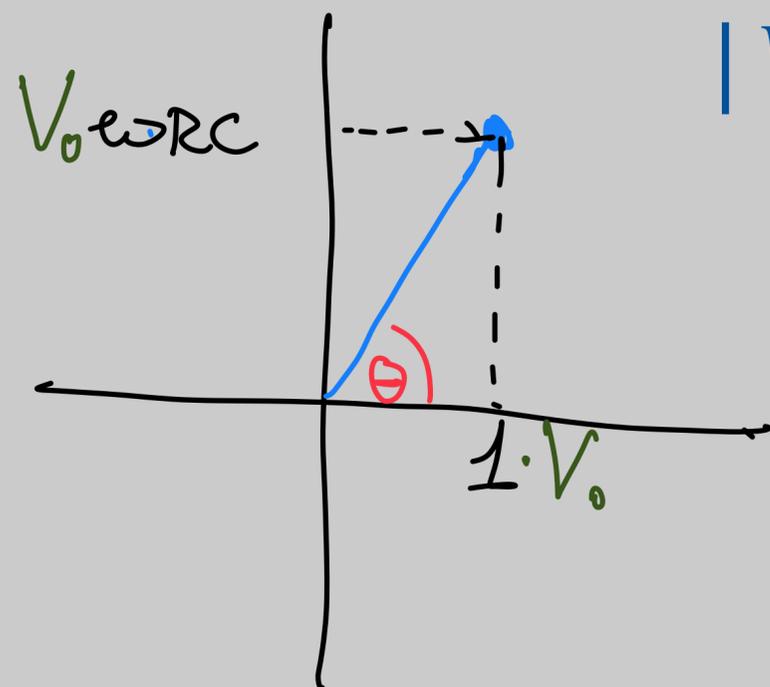
$t \gg RC$

$$v_c(t) = \frac{V_0}{j\omega RC + 1} e^{j\omega t}$$

Transient

$$|V_0(j\omega RC + 1)^{-1}| = |V_0| \left( (\omega RC)^2 + 1 \right)^{-1/2}$$

$$\theta = \angle(j\omega RC + 1)^{-1} = -\text{atan}(\omega RC)$$



## General Solution:

$$v_c(t) = \frac{V_0}{j\omega RC + 1} \left( e^{j\omega t} - e^{-\frac{t}{RC}} \right)$$

$t \gg RC$

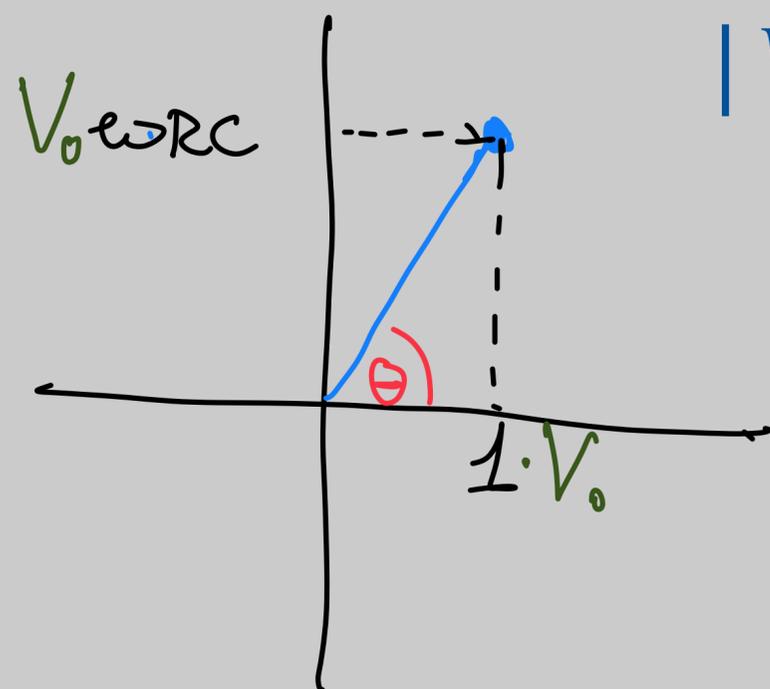
$$v_c(t) = \frac{V_0}{j\omega RC + 1} e^{j\omega t}$$

Transient

$$|V_0(j\omega RC + 1)^{-1}| = |V_0| \left( (\omega RC)^2 + 1 \right)^{-1/2}$$

$$\theta = \angle(j\omega RC + 1)^{-1} = -\text{atan}(\omega RC)$$

$$v_c(t) = \frac{V_0}{\sqrt{(\omega RC)^2 + 1}} e^{j(\omega t + \theta)}$$



# Sinusoidal Response

$$V_s(t) = V_0 e^{\omega t} \quad | t > 0$$

$$v_c(0) = 0$$

$$\Rightarrow v_c(t) = \frac{V_0}{\sqrt{(\omega RC)^2 + 1}} e^{j(\omega t + \theta)}$$

$$V_s(t) = V_0 \cos(\omega t) \quad | t > 0$$

$$v_c(0) = 0$$

$$\Rightarrow v_c(t) = \frac{V_0}{\sqrt{(\omega RC)^2 + 1}} \cos(\omega t + \theta)$$

$$\omega = 0$$

$$\Rightarrow v_c(t) = V_0$$

$$\omega RC = 1$$

$$\Rightarrow v_c(t) = \frac{1}{\sqrt{2}} V_0 \cos(\omega t + \theta)$$

$$\omega RC \gg 1$$

$$\Rightarrow v_c(t) \approx \frac{V_0}{\omega RC} \cos(\omega t + \theta)$$

Low-pass, attenuating high freq!